## Micro Laser Personal Projector

by

Wilfrido Sierra Hernández

Submitted to the Program in Media Arts and Sciences, School of Architecture and Planning in partial fulfillment of the requirements for the degree of

Master of Science in Media Arts and Sciences

at the

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#### Abstract

The proposed research is a study of the technology and applications of *personal projectors*, small and inexpensive video projection devices intended for use in batterypowered hand-held or wearable products. This research will describe some ways of steering laser beams in one dimension to produce up to a one-meter wide screen. A one dimensional micro laser array will be used to avoid the effect of flickering while projecting an image. The use of a single lens to focus all lasers will save the complexity of collimating each laser independently. At the moment, the Micro Laser Personal Projector is displaying an image of 50 x 480 pixels. A DC brushless motor is used to steer the laser beams vertically. A fixed plano-convex lens focuses all 50 lasers. A piezo-electric device will be used on the future to increase the laser projector's image resolution from 50 x 480 to 200 x 480 pixels. With this image resolution the Micro Laser Personal Projector can be used in communications, entertainment, and medical applications.

Thesis Supervisor: V. Michael Bove Jr. Title: Principal Research Scientist

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# Chapter 1

## Introduction

A generally observable trend in technology is that electronic devices are becoming increasingly smaller. For instance, cell phones were initially brick size and are now smaller than our hands, and their batteries last longer than before. Similarly, common projectors tend to be small, inexpensive, and with low power consumption. Currently a projector costs more than a thousand dollars, along with regularly recurring costs of several hundred dollars for its bulb. These projectors are not that inexpensive in comparison to a new technology that is emerging, the Micro Laser Personal Projector, which is expected to cost twenty to fifty dollars.

Sharing a soft-copy of visual information is increasingly important in business and will likely become equally so for consumers [15]. These consumers may want to display an impromptu slide presentation from a pocket PC or laptop, download a map and display it on the wall to see details, or project video from a cell phone in order to have a video-conference with someone they are calling. When privacy or confidentiality is a concern, the Micro Laser Personal Projector (MLPP) will be an addition to a conventional small display, not a replacement for it. Some approaches to this goal have been considered since the late 1960's [4] when a laser beam substituted for the conventional T.V. screen or CRT. However, this approach is far from being a low power consumption approach, since the use of at least two galvanometers for laser steering on the x and y coordinates is required; which necessitates a more powerful single laser beam. Laser projector designs use solid state and gas lasers [4][22] and the most common way to steer the laser beam is through the use of galvanometers. The majority of the previous prototypes have a similar schematic principle. That is, they have input video, which is then processed. After the process, the signal is electronically processed using Pulse Width Modulation (PWM). Next, the signal enters a power amplifier to make it strong enough to light the lasers. This modulated signal is applied to a laser and its light passes through a lens, and from the lens to the deflection system. This deflection system consists of one galvanometer steering the light on the x axis (i.e., from left to right) and another galvanometer steering the light on the y axis (i.e., from top to bottom). Finally, the image is displayed on a screen. Previous work related to the avoidance of using a horizontal scanner has been done using a vector of 64 semiconductor lasers as a light-producing element instead of one laser [7]. Following the idea of using a one-dimensional array of lasers, the MLPP will use 50 micro lasers to save power, and reduce the effect of flickering from the image.

Chapter 2 will explain the MLPP system characteristics. One of its sections will describe in detail how video signals are gathered and the meaning of them all. Another section of chapter 2 is dedicated to the software specifications. In chapter 3, we will see how the lasers where implemented into the system. Then, we will discuss about the MLPP individual parts and how they work, including the optics concepts involved. Furthermore, we will see the whole system integrated and displaying laser video. The last chapter, which is chapter 4, is a motivation of why the MLPP is important to people and their every day life-style. Some future applications to this prototype are discussed within this chapter.

## Chapter 2

# The System

Micro Laser Projection has been under research for several years. Most of the design principles use a single laser beam to produce the whole image. For example, Symbol Technologies [21] is developing what they call the "Laser Projection Display" using a single laser and both a vertical and horizontal scanners. Another way to steer this laser beam is through Amplitude Modulation using spatial light modulation (SLM) [23]. The SLM has an input polarizer and an output cross polarizer. When voltage is applied to the polarizers the laser gets modulated, but almost 50% of the light is lost, and so it is not efficient when dealing with image brightness. Using one laser is not desirable for the same reason and because the laser has to be powerful enough to display all the image information from left to right and from top to bottom. The laser used might be dangerous regarding eye safety and expensive in terms of energy consumption. The MLPP is a device that saves laser power using a horizontal laser array. The complexity of the prototype is rather simple, and the materials used for its construction are as common as a PC fan, a glass or acrylic plano-convex lens, and simple electronics, which make it a very cheap and portable personal projector.

The Micro Laser Personal Projector (MLPP) is composed of many modules and submodules. Figure 2-1 depicts a block diagram of the system. The main module is the *firmware* that drives the behavior of the fifty lasers. This module is called the **Main Driver**. Within the Main Driver there is a huge set of blocks designed and connected in such a way that they manage all the input and output (I/O) signals.

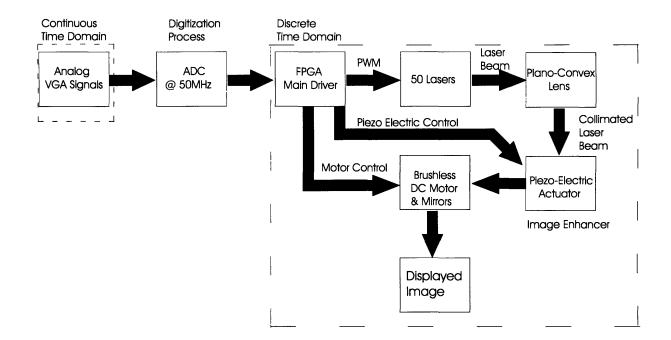


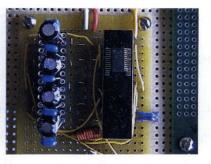
Figure 2-1: Block Diagram of the System

Besides the Main Driver, the MLPP has an Analog to Digital Converter (ADC) that is used mainly to digitize the input analog VGA (Video Graphics Array) signals, so the Main Driver can distribute the digital information correctly and transform these VGA signals into laser pixels or picture elements.

Some other submodules are also important to consider such as a brushless DC motor, a piezo-electric device, a plano-convex lens and the Vertical Cavity Surface Emitting Lasers (VCSELs) [6]. The brushless DC motor is the actuator that has attached to it a six sided mirror prism used to steer the lasers in the vertical (i.e., from top to bottom) direction. The piezo-electric device is used to steer the lasers on the horizontal (i.e., from left to right) direction, to enhance the image resolution, making each laser to depict four different pixels and to cover the gaps that exist between each laser. A plano-convex lens is used to collimate (focus) the lasers in such a way that the divergence is up to ten degrees. This divergence will help us to project an image one meter wide, from a distance of one to two meters from the projected surface. The VCSELs are small dice of 6 mm long, which comes in an array of 25 lasers each, on the horizontal direction. Each laser will be responsible for displaying



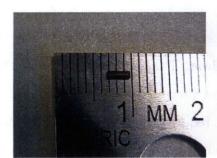




B) ADC08100 ADC

D) A six sided mirror prism attached to a brushless DC motor. Below, all 50 Lasers in their acrylic

protection.



C) A VCSEL Die



E) A 1/3 of a Plano-Convex Lens between the prism and the laser protection.

Figure 2-2: The whole working system

four different pixels from the VGA signal. For example, if we want an image of 200 x 480 pixels we need two VCSEL dice glued together, making a total of 50 lasers. These 50 lasers will display four pixels, and therefore we will get an image of 200 pixels on the horizontal direction times 480 pixels on the vertical direction. Figure 2-2 shows the complete prototype and its parts. A thorough explanation of the system will be described below.

# 2.1 Software Design, FPGA Board and its Peripherals

The most important element of the MLPP design is the software within the Field Programmable Gate Array (FPGA). The FPGA is a firmware device that runs with Xilinx ISE4.1 [10] (or higher) software. Within this software we are able to use digital operators such as *AND*, *OR*, *NOT*, *XOR*, and *NAND* among others. Besides these operators there is another way of doing more complex digital design using the VHDL language. VHDL stands for VHSIC Hardware Description Language. The VHSIC acronym stands for Very High Speed Integrated Circuits. For example, designing a *state machine* out of digital gates, will require many gates, and the connections between them may be difficult to debug whereas using a VHDL pseudo code is just a matter of few lines which are considered to debug.

#### 2.1.1 MLPP Peripherals

#### VGA Signals and Video

The MLPP receives VGA signals from any VGA source. The VGA signals are: **H**-Sync (horizontal synchronization signal), **V-Sync** (vertical synchronization signal), **R** (Red Component), **G** (Green Component), and **B** (Blue Component). The **RGB** components are the ones that can produce a huge variety of colors. Each component has up to 255 (i.e. 8 bit resolution) different levels of intensity, which lead us to a range of 16,581,375 different colors. That is

$$FullColor = R * G * B = 255_{RED} * 255_{GREEN} * 255_{BLUE}$$

Any image displayed on a computer screen is composed of these VGA signals. The image resolution is characterized by the number of pixels (picture elements) that it has, and is measured in M x N pixels. The relationship between M and N follows an *aspect ratio*. The aspect ratio of any conventional VGA image is 4:3, which reads four to three. For example, the image resolution of my screen is 1400 x 1050 pixels. That is

$$M = 1400 \ pixels$$
 and  $N = 1050 \ pixels$ 

and its aspect ratio of 4:3 can be proven as follows:

$$\frac{M}{N} = \frac{1400}{1050} = \frac{4}{3} \tag{2.1}$$

It is important to understand the meaning of **M** and **N**. The value of M implies that there is a certain number of pixels on a horizontal array, where each pixel has an RGB value. The value of N indicates the number of *scan lines*. A scan line is a set of pixels on the horizontal line. In our last example we had a total of 1050 scan lines. Each of these scan lines is a single H-Sync pulse. Figure 2-3 depicts the VGA signal arrangement. The red component starts at the rising edge of the H-Sync pulse, and its value lasts right before the H-Sync pulse goes back to zero. Within a Vertical Synchronization pulse (V-Sync) there are as many as 480 or more H-sync pulses. Each V-Sync pulse forms a field or a "still picture," which refreshes itself every sixtieth of a second or at a rate of 60 Hz (Hertz) or faster.

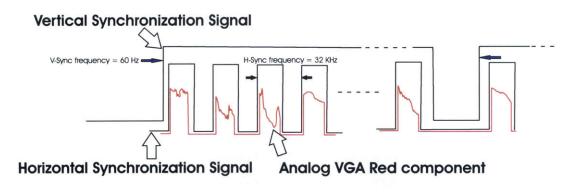


Figure 2-3: VGA Signals

The continuous refreshment of the field will produce a set of displayed still images. This V-Sync refreshment is so fast that the human eye would not notice the difference between the first and the following image. The image production process described above is nothing but a sequence of pictures which we call video.

Summary: Video is decomposed into five major signals which are the R, G, B, H-Sync, and V-Sync signals. The combination of the RGB components produce color, and the color is a function of the RGB intensities. A single H-Sync pulse at its rising edge contain M number of pixels and each pixel contain an RGB value. A set of N number of H-Sync pulses (referred to the V-Sync) produces a "still image". The rate of every still image is of one sixtieth of a second or 60 Hz. This rate is so fast that the human eye cannot distinguish between two different fields. The sequence of images is what we called **video**.

#### Analog to Digital Converter (ADC)

Once we know how video behaves, now we are able to understand how VGA signals will be treated for the MLPP. The signal to be treated in this section is the **R** component. This signal comes from any VGA source. The R component passes through an 8-bit ADC. The digitized signal is then processed by the FPGA and video is displayed after this process.

The ADC08100 used for the MLPP is a device which receives one analog signal which in turn is converted into an 8-bit digital word (refer to Appendix A). The selection of this chip was based on the assumption of its small power consumption which reduces to 1.3 mW (mili-watts) per MHz (Mega Hertz) of *clock frequency*. Its clock frequency is taken from the FPGA clock which is 50 MHz, giving a top power consumption of 65 mW. With this specification the ADC08100 can run on battery power.

The process of digitization is depicted in Figure 2-4. As seen in the figure, the analog signal enters the ADC as a function of time X(t). Our sampling time Ts is selected in such a way to satisfy the Nyquist Theorem, which states that the highest frequency of any signal which can be accurately represented is less than one-half of the sampling rate. That means that if we chose a resolution of 640x480 pixels, we have a total of 307,200 pixels depicted at 1/60 seconds or 16.66 ms. With this information we can assume that each pixel takes about  $\frac{16.66}{307,200} = 54.25$  ns per pixel,

Table 2.1: Red scale based on the two most significant bits of the ADC08100 D7 and D6

D7 (MSB)	D6	Description
1	1	Laser Brightness High
1	0	Laser Brightness Medium
0	1	Laser Brightness Low
0	0	Laser Off

which corresponds to 18.43 MHz. The FPGA clock frequency that feeds the ADC is 50 MHz, which means that the analog signal can be represented accurately into a digital signal X[n].

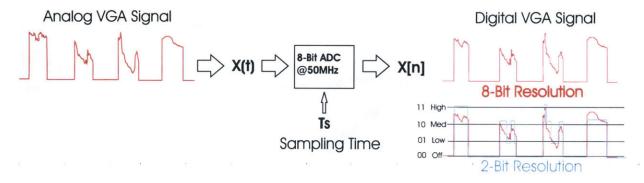


Figure 2-4: Digitizing VGA Signals

The resolution chosen for the MLPP is now based on two bits. The data taken from the ADC08100 is only the two most significant bits (MSB). With this information we will see a 2-bit displayed image. The final output will be a four red shade based video. The criterion of this red scale is shown in Table 2-1.

#### 2.1.2 The Software

The MLPP is designed mainly by software. The VHDL language is a powerful tool used to program any FPGA device. The MLPP design consists of the management of many I/O signals. The main input signals are the R component, H-sync and V-sync signals, including the FPGA clock signal that runs at 50 MHz.

The software design is composed of many independent modules that are connected

in a way that combines digital design as well as software pseudo codes, which makes it easy to debug. The basic idea is depicted in Figure 2-5. The process begins with a data selector where all signals are processed individually. This data selector takes the values of the red components, Red0 and Red1, Red0 being the **MSB** from the ADC. Inside the data selector there is an internal counter (hcnt) directly controlled by the **CLK**(clock) and the H-sync signals. Each clock cycle is counted by the internal counter while the H-sync signal is in "logic one," and it stops and resets when H-sync is in "logic zero."

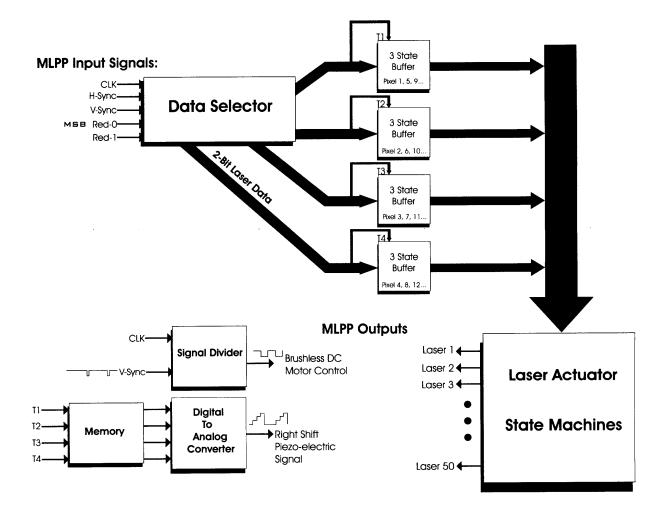


Figure 2-5: Software Block Diagram

The data selector will start to acquire the red component data 2  $\mu$ s after the H-sync signal sets up to logic one, in other words, after the internal clock counts up

to one hundred clock cycles. The FPGA needs to store each pixel's information from this red component signal every two clock cycles or every 4 ns. The FPGA will not get any data in less than this time.

At this point for the MLPP design, we need to remember that our goal is to reach a resolution of 200 pixels on the horizontal line by 150 on the vertical line. This is following the aspect ratio of 4:3 where 200/150 = 4/3. Since we have only 50 micro lasers, we are going to focus first on the horizontal line. The solution to the problem is to store the red component data into a three-state buffer. A three-state buffer is basically a place where digital data can be placed temporarily and its outputs are "logic one," "logic zero," and "high impedance" (Z). Table 2.2 shows how data is stored at each buffer. Each buffer stores data of every other pixel. In this case since

Table 2.2:	Laser	Storage	for	Four-Way	Shifting

Buffer at Time	Internal Counter Range	Trigger Range
T1	100 - 492	98 - 99
T2	102 - 494	415 - 416
T3	104 - 496	732 - 733
T4	106 - 498	1049 - 1050

we need four pixels per laser, the data selector skips every four pixels to the particular buffer. For example, the buffer that is triggered by the signal T1 receives information from pixel 1 and at a later time receives information from pixel 5, pixel 9 and so on. The same happens to the remaining three other buffers.

Looking back to Figure 2-5, we observe four three-state buffers. These buffers have almost as many inputs as outputs. They receive the red component data and store it until the control signal Tn is triggered and enables the buffer to pass the information to the laser actuator. Meanwhile, when a buffer is not triggered, it remains at high impedance. A high impedance system could be compared to a system that is disconnected in such a way that no current may flow through the wires. High impedance is used in these buffers to avoid undesired digital information that may interfere with one buffer or another. When all four buffers are completely loaded,

a horizontal scan line has been accomplished. When the next scan line begins the buffers will acquire their information right after they send the previous data to the laser actuator. The process keeps going until it reaches the last horizontal scan line. Remember from Table 2.2 that the first set of data is triggered when T1 is at logic one, two clock cycles before we begin to acquire the next scan line information. This means that the projected display is only *one scan line* delayed from the VGA source image.

The laser actuator is a module that receives buffer information at the time Tn. This actuator is a set of state machines that are controlled by an internal counter, similar to the data selector counter. A state machine is a decision-making module that depends directly on a specific input, and the number of decisions is directly proportional to  $2^n$ , where n is the number of bits. For example, if we have a two bit statement or input, the only options we will have as an input are: "00", "01", "10", and "11" (i.e. only  $2^2 = 4$  options). In our particular case, these are the values of the red component that are stored on the three-state buffers. Each state machine will receive only one of the past options and will determine what to do with the received information. For example, if the input is a "11", which is brightness HIGH (from Table 2.1), then the state machine will "decide" for the designated laser to stay on for 314 clock cycles or 6.28  $\mu$ s. This is the most time that the laser stays on. When the input is a "10", brightness MEDIUM, the state machine will light up the laser for 157 clock cycles or 3.14  $\mu$ s. Finally, if the actuator receives a "01", brightness LOW", the laser will be on for only 79 clock cycles or 1.58  $\mu$ s. However, if the data received is a "00", laser OFF, the laser will remain off at least the 314 clock cycles.

To better understand Table 2.2, we will summarize the process. After all the buffers are completely loaded, the system is waiting for the trigger signals, which are: T1, T2, T3, and T4. The buffer triggered by T1 starts storing data when the clock cycle equals one hundred. This buffer will get data from the data selector every eight clock cycles, since we need to skip four pixels and each pixel takes two clock cycles to be acquired by the buffer. The following equation explains the behavior of the

internal counter hcnt:

$$hcnt = 100 + 8 * 49 = 492 \tag{2.2}$$

This equation helps us to determine the lower and upper bound for the first 50 skipped pixels which are from pixel 1, pixel 5, pixel 9, all the way to pixel 197. The rest of the table follows the same equation but starting from 102 for buffer T2, 104 for buffer T3, and 106 for buffer T4, which ends on pixel 200. The trigger range for buffer T1 starts on clock cycle 98 and lasts only two clock cycles for accessing the laser actuator. After 314 clock cycles, the buffer T2 is triggered at clock cycle 415 instead of clock cycle 413, giving only two more cycles to avoid any glitches. The same process is repeated for buffers T3 and T4.

#### **Piezo Electric Actuator**

As shown in Figure 2-5, there is a module which has as inputs T1, T2, T3, and T4 signals. These signals go to a submodule, not depicted in the figure that is a memory that retains the Tn information. The module's outputs go to an external digital to analog converter (**DAC**), which is a small ladder resistor array. This DAC produces a ladder like signal. This signal is very important in order to drive the piezo-electric device to make the right to left jitter.

#### **Brushless DC Motor Actuator**

Finally, the last module depicted in Figure 2-5 is the DC Brushless Motor Actuator. This actuator has an internal counter "*vcnt*" that is governed by the FPGA clock signal (CLK) and the V-Sync signal. As depicted on Figure 2-3 the V-Sync signal is not uniform, in other words, is not 50% duty cycle. Instead it is 96% duty cycle, which means that 96% of the signal periodic time stays on logic one, and the rest is on logic zero. The brushless motor actuator is nothing but a compensator. For example, the signal at 96% duty cycle is transformed to a signal at 50% duty cycle. This signal is critical to synchronize the prism's faces attached to the brushless motor, in order to project the display without wiggling.

## 2.2 VCSEL Technology and Fabrication

The acronym VCSEL stands for Vertical Cavity Surface Emitting LASER and contains itself another acronym, which is LASER. This word stands for Light Amplification through Stimulated Emission of Radiation. To better understand this concept we must start by explaining some basics. First, let us start with the concept of light generation. Consider any particular metal, and no matter which one we pick it has atoms. Next, we concentrate on one atom and analyze its structure. As we may know, atoms are composed of neutrons, protons, and electrons. The electrons normally travel within an orbit around the nucleus, which is composed of neutrons and protons. When the electrons travel in their own orbit we might say that the atom is in ground state [13]. Supposing that energy is applied to this atom (e.g., thermal or electrical excitation), and this energy is strong enough to move one or more electrons to a "higher" orbit or energy level, then we may say that the atom is in the **exited** state. After this phenomenon and naturally, this electron will return to its equilibrium or its original orbit; in other words, the atom goes from the exited to ground state. This action will constitute a photon liberation. Light is composed of photons, so the process described above will be is light generation.

Light amplification is the next thing to consider for lasers. Consider a third-level system that consists of a ground state, an exited state, and a super-excited state. To begin with the example, let us say that a set of atoms in the system are at ground state, and after some type of excitation most of the atoms go from ground to the excited state. The super-excited state is utilized as a short-lived "pivot point." When a photon enters into the system it "knocks" an electron from the excited state down to ground state, thus creating a new photon. Two photons will leave the system instead of one. After these photons leave the system, they are reflected by a mirror sending these two photons back to the system. At this moment, they enter the system and another two atoms get knocked down from the excited to ground state. This time there will be four photons leaving the system in the opposite direction and that are then reflected by a second mirror. This amplification process is called **stimulated** 

#### emission.

A laser is composed of three main things: a gain medium (e.g., third-level system), a fully reflecting mirror, and a partially reflecting mirror. With all these three elements we can consider this as a positive feedback system. Laser light is then created by the following procedure; an external force like electricity, excites the atoms within the gain medium, and a photon enters the gain medium. As an output we will see more photons and they will be bounced back by the partially reflecting mirror. Then, these photons re-enter the system and light gets amplified again. As an output we will see even more photons exiting in the opposite direction and getting bounced back by the fully reflecting mirror, forcing to re-enter the system so light is amplified again. When the light is strong enough, it penetrates the partially reflecting mirror and we get what is called a coherent light.

The VCSEL emission light occurs perpendicular to the wafer surface as depicted in Figure 2-6. As seen in the figure, the topology of the VCSEL has been designed in such a way that allows two distributed Bragg reflector (DBR) mirrors to form the laser cavity [11] [9]. VCSELs have three major characteristics: a low divergence circular beam, the ability to wafer level test *before* packaging, and the capability of fabricating a dense two-dimensional laser arrays [14].

These three characteristics helped us to reduce costs on the MLPP implementation. For example, the low divergence circular laser beam reduced the complexity of designing or implementing a complex lens for collimating the laser beams; instead a simple plano-convex lens was used for this purpose. Testing before packaging represents a big advantage because we can have a relatively high production yield which in the longer term will enable high volume and low-cost VCSEL manufacturing. The third characteristic is the best for the MLPP because it is possible to fabricate these lasers by arrays.

As mentioned before, laser light depends on mirrors. The DBR mirrors on the VCSELs are the key elements for these devices. These DBR mirrors are fabricated by techniques such as Molecular Beam Epitaxy (MBE) and Metal Organic Vapor Phase Epitaxy (MOVPE) [14] [2], including the development of *in situ* growth diagnostics.

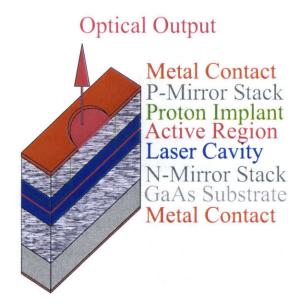


Figure 2-6: VCSEL Cross Section

These DBR mirrors are separated by a thickness of a multiple of  $\lambda/2$  where  $\lambda$  is the wavelength of the emitted light [8]. The distance between the top and the bottom mirror stacks is called the optical cavity which is typically one wavelength long where we find the active laser region. These mirrors are designed in such a way so as to be almost 99% reflective. Monolithic semiconductor DBR mirrors have a lower index of contrast and thus require a higher number of mirror periods for high reflectivity [12]. As we can see from Figure 2-6, DBR mirrors come in stacks of which each layer is called a period. The higher the period, the more reflective it becomes. Notice that on the top of the VCSEL we have fewer mirror stacks than in the bottom [20]. The bottom layers have to be more reflective than the top so the light can "escape" on the top. The composition of the layers is chosen to maximize the index of contrast and to be transparent to the laser light [18]. For example, if we want a 850 nm or 650 nm wavelength laser we can use pairs of  $Al_{0.2}Ga_{0.8}As/AlAs$  or  $Al_{0.5}Ga_{0.5}As/AlAs$  respectively.

The VCSEL technology is mainly used for telecommunication purposes as well as for proximity sensors, encoders, laser range finders, laser printing, bar coders, and optical storage among others. Because of the nature of the typical VCSEL applications, they are commonly fabricated for wavelengths above 690 nm. This wavelength is barely visible to human eyes, since the range of eye sensitivity goes from 440 nm to 690 nm approximately [17]. Some VCSEL prototypes have been developed for wavelengths shorter than 690 nm, like the MLPP, almost on the infrared region. Laser image displaying has not been considered before, and the development of ideal VCSELs in wavelengths such as 440 nm (Blue range), 555 nm (Green range) and 625 nm (Red range) are now in research. In the near future VCSELs developed on the RGB range will help us to display a full color laser projection display.

## 2.3 Optical Devices

Thinking of the micro laser projector and its small lasers, the first big problem to solve was to focus all fifty lasers. A lot of ideas were considered. One of them was to use a holographic lens array to collimate them all. The idea was focused on collimating each and every single laser independently, which was a very difficult task. Thinking a little harder and realizing that the whole array is so small that it could be taken as a small point source, we could easily collimate every laser with a single plano-convex lens. Therefore, the optic devices related to the MLPP prototype were very simple to get.

Taking the **Thin-lens Equation**, [13] often referred to as the Lensmaker's Formula

$$\frac{1}{s_o} + \frac{1}{s_i} = (n_l - 1)(\frac{1}{R_1} - \frac{1}{R_2})$$
(2.3)

where  $s_o$  is the distance of the object or point source (in this case is the distance between the lasers and the lens),  $s_i$  is the distance of the image,  $R_1$  and  $R_2$  are the radii of the lens, and  $n_l$  is the refraction index of the lens, helped us in finding the right type of lens needed to collimate all lasers. For a plano-convex lens radii, values are  $R_1 = \infty$  and  $R_2 < 0$ , reducing equation 2.3 to the following equation

$$\frac{1}{f} = \frac{n_l - 1}{|R_2|} \tag{2.4}$$

where f is the focal point of the lens. Starting from the fact that our lasers were taken as a spherical wave point source which can be represented as

$$a(x, y, z, t) = A \frac{e^{j(kR - \omega t)}}{jR}$$

$$(2.5)$$

and we suppose the paraxial approximation (refer to Figure 2-7), a very short distance approximation. We end up with the following math expression for our point source which is

$$a(x, y, z) = Ae^{j2\pi\frac{z}{\lambda} + j\pi\frac{x^2 + y^2}{\lambda z}}$$

$$(2.6)$$

where  $k = \frac{2\pi}{\lambda}$  is the wave number where this time  $\lambda$  is the period of each outgoing wavefront.

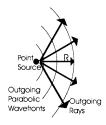


Figure 2-7: Spherical Wave in the paraxial approximation

The role of any thin lens is to *transform* a spherical wavefront into a plane wave and vice-versa. The problem now, is that when our point source (the lasers) is at the exact focal point f, we will produce an image of the size of the diameter of the lens. To avoid this inconvenience, it was necessary to move the focal point a little bit farther away such as  $f' = f + \Delta f$ , in other words, to de-focus the system a little bit. This action yielded an output divergence where the final image increased its size until we had the desired image size of one meter diagonal long. Some of the trade offs that we deal with when de-focusing the system was that our image blurred. This could be inconvenient, but fortunately was not, because blurring the image (not too much) helped to cover the inherent 100  $\mu$ m pitch gaps that the VCSELs have. Although, these gaps were not covered at 100%. Another special technique is going to be applied to enhance the image resolution. This technique is described in section 2.4, where piezo-electric devices will be used.

A previous MLPP prototype used a galvanometer for beam steering. It consisted in a single mirror attached to the galvanometer. This mirror was deflected by some degrees and then move back to the original position. The problem with this prototype was that the energy used to move the galvanometer was too much. For example, to move the mirror one degree, we needed 50 mA. Moving the mirror 24 degrees, would take about 1.2 A, which is considered too much current to drive. Another issue was that the galvanometer is very heavy and very big. A brushless or a stepper motor was considered as an alternate solution to the beam steering problem, including power consumption.

The other optic device used for the MLPP was a rotating mirror. This mirror must have certain characteristics so it can display video at a rate of 60 Hz. The first step was to choose a shape. The shape that was convenient for the application is a polygon. The first thing was to calculate the number of facets. The formula for the number of facets is given by [5]:

$$n = \frac{720(C)}{\theta} \tag{2.7}$$

where  $\theta$  is the active optical scan angle and C is the duty cycle. The number must be an integer, if not, then there is no exact solution for the system. In our case, we previously decided a duty cycle of 50% (section 2.1.2), and a 60° displayed angle, giving a total of  $\frac{720(0.5)}{60^\circ} = 6$  facets. The second step was to decide the size of each facet. We already know that the width has to be at least 15 mm long, because both laser arrays measure 12 mm long. The length of the facets must be calculated from [5]:

$$L(mm) = \frac{(D'+1)}{1-C}$$
(2.8)

where D' is the laser spot width, which experimentally measure approximately 0.2 mm, where the laser path is the two VCSEL arrays that measure all together physically from laser #1 to laser #50 10 mm. Our hexagon length must be equal to or greater than  $L(mm) = \frac{D'+1}{1-C} = \frac{0.2+1}{1-0.5} = 2.4mm$ . Some experiments were done while de-

signing gold plated hexagons. The results of these experiments showed that the bigger the facet length the bigger the displayed image. For experimental purposes, the size of the polygon facets are 12 mm each. In theory a 6 mm facet can be implemented.

## 2.4 Piezo-Electric Devices

Most piezos are made of polycrystalline oxide ceramic PZT. The letters PZT stand for lead (Pb), zirconium (Zi), and titanium (Ti), which are the basic components of most piezo devices. Basically, any piezo-electric device of any material reacts when applying a voltage to it. This kind of reaction is expressed into internal displacement or deformation. This reaction can be reversed and we might obtain voltage by deforming or pressing the piezo-device. Some commercial piezo-electric devices come in stacks. These stacks are often used to increase the displacement of the device, but the increase of displacement, even with the increase of stacks, will reduce the resonant frequency of the piezo devices.

The MLPP has a resolution of  $50 \ge 480$  pixels in its recent prototype. To increase its image resolution it is convenient to think in new ways of increasing the number of pixels. Fortunately the software provides the laser driver to make each laser display four different pixels. Nevertheless, a physical actuator has to be implemented to "move" the lasers from left to right four positions. An alternative to do this task is the use of piezo-electric devices.

Recalling from section 2.1.1 Figure 2-3 the H-sync signal has a frequency of approximately 32 KHz, and so the right shift piezo-electric signal *(RSPES)*. The idea is to apply the RSPES to a specific piezo-electric device to make a right shifting of a plano-convex lens. This action will give us the effect of "moving" the lasers from left to right four positions. The desired displacement for each shift is about 10  $\mu$ m since the diameter of each laser is about that size. The total desired displacement is 30  $\mu$ m at 32 KHz. Unfortunately there is no actual piezo-actuator that can give us this characteristics of moving at least 30  $\mu$ m with a resonant frequency of 30 KHz or higher [3] [1].

# Chapter 3

# Implementation

This chapter will describe how the MLPP has been implemented, including circuit and board designs as well as micro technology techniques. Some circuit designs and considerations on building micro-boards for the VCSELs will be described on this chapter.

## 3.1 Mini Board Design and Implementation

Before bonding lasers at size of  $10\mu$ m of diameter with a pitch of  $100 \mu$ m from one laser path to the other, a mini board has to be designed. Implementation of the ceramic board requires training at the Microsystems Technology Laboratories (MTL). The design is depicted in Figure 3-1. The board dimensions were designed in such a way

-	h. d	b	
Ξ			E
			E:
30	m	Г	TE
	111		

Figure 3-1: Laser Board Photo Mask (Actual Size)

that it must fit into another 1 in x 1 in board, which is a very common technique when a design this small has to be implemented into a "normal" size board. The mini

board is a ceramic made board with gold coating. The technique applied for this mini board is called the *"liftoff"* technique. It consists of the following procedures:

- 1 First we take a blank ceramic board. For best results, this board has to be immersed in acid called *piranha*, which is a 4:1 composition of sulfuric acid  $(H_2SO_4)$  and hydrogen peroxide  $(H_2O_2)$ . This process cleans all impurities that it might have so the following steps can be successful. Failure to a previous cleaning with "piranha" may cause future damage to the board design.
- 2 After carefully cleaning the board, we rinse the board, and blow dry with dry air. Then the board goes to the "Hard Bake" oven. This oven is around 120 to 130°
  C. We let it bake for 20 to 30 minutes.
- 3 Once the sample is ready, we take it out from the hard bake oven and begin with the small vacuum chamber, where we deposit the board and a small quantity of *HDMS* on a small cup on its side. This substance is nothing but a "primer" for the *photoresist* that will be poured afterwards. This primer is used so the surface of the board smoothes up, letting any type of liquid to attach uniformly on the board surface. We wait 10 minutes until the HDMS evaporates completely. At this point we take the sample into the *spinner*. We must have ready the photoresist AZ 5214-E. The photoresist is a layer which in the future we will use to make the liftoff. On the spinner, we control the *RPM (Revolutions Per Minute)* for two main purposes: the first is to pour the photoresist on the sample at a slow rate so it uniformly covers the board in order to avoid "gaps" or unfilled areas among the ceramic board; the second is to increase the speed (at 2500 RPM) to let it dry.
- 4 Now that the sample is uniformly covered by the AZ 5214-E photoresist, we take it to the "soft bake" oven (at 90° C) for about twenty minutes. This procedure is just to let it dry, but we must be careful not to exceed this time, otherwise the Az 5214-E will harden and will make it almost impossible to accomplish further steps.

- 5 The sample is now ready to make the photolithography process. Remember the board design from Figure 3-1, it was used to make a transparency photo mask. The mask is placed on top of the sample. Then an UV (ultra violet) light is exposed, as if we where taking a picture using a flash. The exposition time at this point is very critical. It must not be less than 2.5 seconds, and must not exceed 4 seconds. The consequences may vary. On one hand if the exposure is less than 2.5 seconds, then when we try to develop the sample, all the photoresist will be washed down. On the other hand, if the exposition time is exceeded, then it will not be developed at all.
- 6 After a successful UV exposure, again we take the sample into the soft bake oven for another thirty minutes. If we exceed this time, developing will not be successful.
- 7 The sample must be ready for a final exposure. This time we increase the exposure time to 30 seconds.
- 8 After the final exposure, we are ready to develop. The developer used is the AZ-422. We pour this developer on a recipient and immerse the sample in this substance. After some time, we will notice that the traces and paths start to appear. The developer dissolves the parts that are not exposed by the UV light. We must be very careful and be sure that the photoresist is completely dissolved on the desired areas. Failure to dissolve it will cause undesirable results. Once we see that the traces are well defined and the photoresist is dissolved completely, we immerse our sample in water to wash it from the developer to avoid damage to the board. Finally, we gently blow dry.
- 9 At this point we are ready to deposit gold on the board. To begin, we must put the sample on the *Barrel Asher*. This device is a plasma chamber that cleans all kind of impurities. After the Barrel Asher, we put the sample into the *E-Beam* (Electro Beam) chamber. This chamber is a high vacuum chamber where a metal is exposed to an electron gun beam that evaporates it at a very high temperatures. For our case, we will put titanium and gold. The Ti is evaporated

and its vapors will be poured on our sample. The desired deposition thickness is around 50 Å(Angstroms), so it will attach well to the ceramic board. For gold deposition it is desirable to have around 500 Å to 700 Å deposition thickness, so the inherent resistivity is not too big and avoid in-board heating when the lasers are all turned on.

10 Finally, after a successful Ti and Au deposition, we put our sample into acetone to remove the photoresist from the ceramic board. The final result is a board with gold traces, as depicted in Figure 3-2., ready to be bonded by the gold wire bonder.

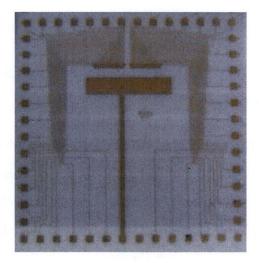


Figure 3-2: Ceramic Laser Board

#### 3.1.1 Wire Bonding

Another difficulty is to bond the fifty lasers into the ceramic board. As it seems, wire bonding is not an easy task. Previous knowledge of wire bonding is required. Before starting to bond, the two dice are glued with silver epoxy and we let dry for a few hours. A recommendation is to use isopropyl to clean the lasers before bonding. Patience is the key to bond the lasers to the ceramic board. Figure 3-3 shows one laser die ready to be bonded. The capillary is positioned on top of the VCSEL array

over the gold pads. Moving the capillary is like moving a computer mouse, which has one button that is used to trigger the capillary from top to bottom.

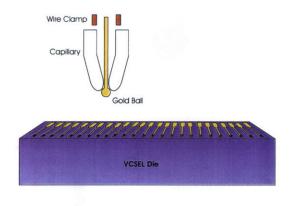
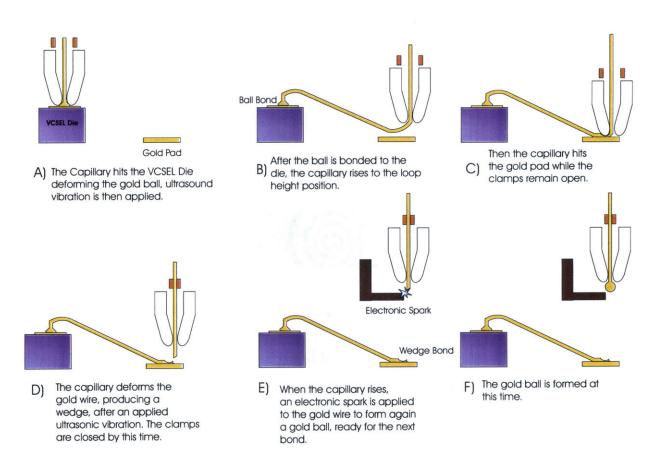


Figure 3-3: Gold Ball Bonder and a VCSEL Die

Figure 3-4 depicts the sequence of how the gold wire bonder works. Figure 3-4 A shows how the capillary hits the VCSEL's gold pad using ultrasonic vibration to make a good gold-to-gold bonding. The clamps are open at this moment so the gold wire moves freely down. After a successful gold-to-gold bonding, the capillary moves upright to the loop position height predetermined by a knob located on the wire bonder device. Then the capillary is moved towards the mini board gold pad as shown on Figure 3-4 B. When the capillary is in the right position the mouse button is triggered again and a wedge bonding is done using an ultrasonic vibration to bond the wire to the mini board as depicted in Figure 3-4 C and D. Finally, the clamps are closed when the capillary moves upright again, so the the gold wire stays down in the right position. A spark hits the tip of the wire to form a gold ball again ready to make the next bond. Figure 3-5 shows a real gold and wedge bonding.

### 3.2 Hardware Implementation

After wire bonding the lasers, an encapsulated device was fabricated to avoid dust and other impurities that may land on the device. A picture of the encapsulated device is depicted in Figure 3-6. On top of the protection device lies a plano-convex lens that focuses the laser beams. Above the lens, a prism mirror is attached to a





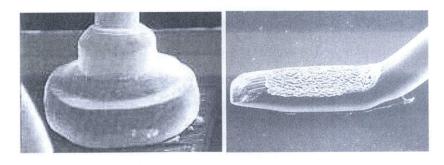


Figure 3-5: Gold Ball and Wedge bonding respectively [16]

brushless motor. This motor is the beam actuator that deflects the lasers in a vertical way. This steering subsystem is connected directly to the FPGA board. The internal FPGA design can be seen in Appendix A.

A large variety of mirrors were designed, from cover glass or acrylic, in different sizes. Both materials were gold coated on the E-Beam chamber with a variety of

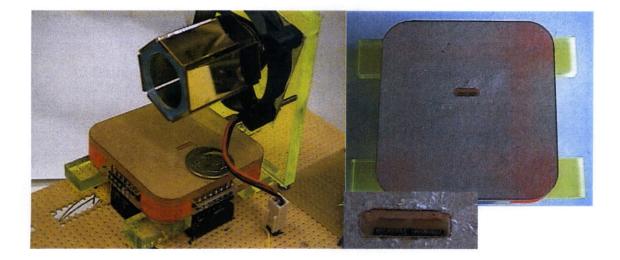


Figure 3-6: Laser Protection mounted over the board

gold thickness deposition in order to find the best reflective index, that allows the laser to be 99.9% reflected. The recommended thickness to get this reflective index is approximately 920 Å. A higher deposition may cause damage on the acrylic, burning it down and losing its mirror properties. As for the cover glass, a thicker deposition may cause blurriness on the sample, completely ruining its mirror properties. The deposition rate is also important. Making a deposition at a rate equal to or lesser than 1 Å/s is strongly recommended. A higher deposition rate will require more energy and therefore a higher temperature inside the vacuum chamber.

Many lens sizes were implemented to verify the most convenient type of planoconvex lens. The type of lenses selected are shown in Table 3.1. After trying many focal lengths and lens sizes, the most convenient lens to use was the 15 mm diameter lens because it focuses all lasers with the desired divergence, and also because it is at the limit of collimating these lasers. The 18 mm and 20 mm lenses could be used, but they occupy most of the space on the prototype. After the 15 mm lens selection, the next step was to cut this lens into three equal portions to reduce space, so the middle part was the only part used to collimate all the lasers (see Figure 3-7).



Figure 3-7: Implemented Lens

Table 3.1: Lens Dimension Options

Lens Diameter (mm)	Lens Focal Length (mm)
12	12
15	15
18	18
20	20

### 3.3 Results

The MLPP is displaying a video image with a resolution of 50x480 pixels. The implementation of a piezo-electric device is still under research. At the moment there is no piezo-electric device that can actually give at least 40  $\mu$ m displacement at a resonant frequency of 32 KHz. New ways of solving this issue have been taken into consideration. One possible solution is to reduce the number of horizontal scan lines so we can reduce the horizontal frequency, in order to reach a resonant frequency from approximately 12 to 15 KHz. At this frequency there are many piezos that produce a displacement of 40  $\mu$ m with a force of at least 5 N (Newtons), enough to move a glass lens from left to right. Another way to solve the problem is to implement a 4 sided mirror polygon spinning at a rate of approximately 32 KHz, where each of its sides has a degree of deflection different from the other. With these deflections it might be possible to achieve four different displacements from left to right, to increase the image resolution. One last solution is to build a denser VCSEL array so the gaps

between each laser are barely noticed. A problem with this will occur since VCSEL robustness is evidenced by maximum continuous wave lasing temperature  $T = 200^{\circ}$  C [19]. This means that VCSELs produce heat, and the more current they drive, the more heat they dissipate and vice-versa. Since its robustness is a function of heat, the closer the lasers on the VCSELs are placed against each other, the more heat is going to be transferred on each laser cavity. This information proves that the closer they get, the less current they have to drive and therefore the less energy we need to drive them, and the dimmer the image we get. Figure 3-8 depicts the laser images displayed by the MLPP. At the middle of the picture appears a black line. This black line is a gap that each VCSEL die has inherently. When gluing both VCSEL dies together, there is a gap of approximately 440  $\mu$ m, equivalent to 4 laser spots. We can overcome this problem by either getting a straight 50 laser array instead of two arrays of 25 lasers each, or to die saw both 25 array VCSELs. The second option is too risky and so it was not considered.



A) Original Image

B) Projected Image

Figure 3-8: Displayed Laser Image

The laser images are not bright enough at day light illumination due to the wave-

length of 670 nm. The brightness will increase in a non-linear way as the wavelength decreases up to 555 nm. Meanwhile, in a not too dark spot, a remarkable image is seen. Unfortunately present technology can not support brighter VCSELs because their applications are not as common as those in communication business. At the moment, blue VCSELs are being developed for high quality image processing as in printers and DVD devices.

#### 3.3.1 Evaluation

The MLPP prototype gives the final resolution of  $50 \ge 480$  pixels. However, the expected resolution of  $200 \ge 480$  pixels has not been implemented yet. The signal driver that controls the piezo-electric device is giving the correct signal output. This signal can be amplified and implemented directly to the piezo-electric device that makes the shifting to the right.

At this moment, the MLPP still a prototype that can be enhanced in many ways. So far, with the software control and the proof that it is possible to display video out of 50 micro lasers, the FPGA system can be shrunk into an ASIC (Application Specific Integrated Circuits) where this ASIC could be as small as 1 cm x 1cm. A relatively small brushless DC motor can be used, where a small mirror prism could be attached to it. The smaller the motor, the less current it will use. It can be estimated that this motor could use less than 20 mA, which is the actual current that the DC motor uses on the prototype. The plano-convex lens has dimensions of 15 mm x 7 mm x 5 mm.

### Chapter 4

# **Future Work and Applications**

The MLPP development grows as a function of technological innovations such as the development of green and blue VCSELs. However, this inconvenience will not stop the relevance of displaying laser images, since the technical process is almost straight forward when trying to upgrade the MLPP from monochromatic to full color video. Implementing the RGB VCSELs could be a problem when trying to align them. We might find that each VCSEL die will be independent and most likely we will not find a red and blue, a green and blue, or any possible combination of VCSEL colors on the same wafer. Fortunately, the freedom of having each RGB VCSEL arrays as an independent die will let us overcome with the problem, which is the chromatic aberration due to the different refraction index that the red, green and blue wavelengths have when passing from air through a glass or acrylic lens. Figure 4-1 A depicts an axial chromatic aberration (A·CA) for a plano-convex lens. As seen on the figure, when RGB light enters the lens, the RGB lasers demonstrate their different refraction index and therefore they have different focal points denoted as *FB*, *FG*, *FR* [13].

It is not easy to figure out the RGB VCSEL alignment since the focal points well as the deflection angle are different. A possible solution is to use a prism which will combine the three colors and transform them into a single RGB beam as depicted in Figure 4-1 B. After the light is combined, a thin achromatic doublet may be used to collimate all lasers. An achromatic doublet design should be done following some

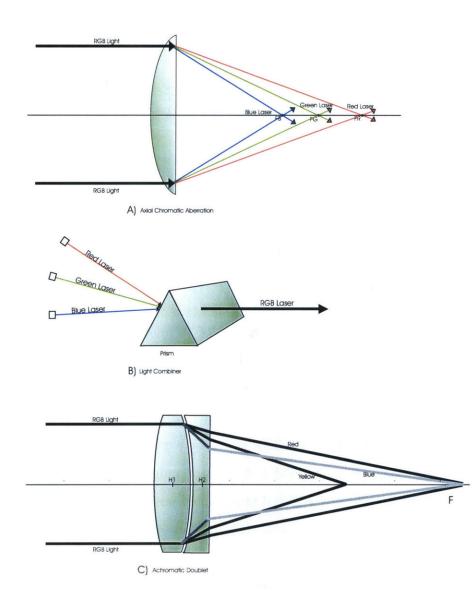


Figure 4-1: Chromatic Aberration

optic equations where the doublet can be considered as a positive and a negative lens separated by a distance d having a compound focal point f

$$\frac{1}{f} = \frac{1}{f_1} + \frac{1}{f_2} - \frac{d}{f_1 f_2} \tag{4.1}$$

Referring to equation 2.3 and considering that

$$\frac{1}{s_o} + \frac{1}{s_1} = \frac{1}{f} \tag{4.2}$$

we can re-arrange equation 2.3 and 4.2 to get  $\frac{1}{f_1} = (n_1 - 1)\rho_1$  and  $\frac{1}{f_2} = (n_2 - 1)\rho_2$ following the idea, we can write the following equation

$$\frac{1}{f} = (n_1 - 1)\rho_1 + (n_2 - 1)\rho_2 - d(n_1 - 1)\rho_1(n_2 - 1)\rho_2$$
(4.3)

This final equation describes the focal point of the doublet, and since FR and FB are the extreme focal points of the system, we want them to be equal such that  $\frac{1}{f_R} = \frac{1}{f_B}$ and using equation 4.3

$$(n_{1R}-1)\rho_1 + (n_{2R}-1)\rho_2 - d(n_{1R}-1)\rho_1(n_{2R}-1)\rho_2 = (n_{1B}-1)\rho_1 + (n_{2B}-1)\rho_2 - d(n_{1B}-1)\rho_1(n_{2B}-1)\rho_2$$

$$(4.4)$$

Suppose that the distance d = 0, where the two lenses are in contact. The expression reduces to

$$\frac{\rho_1}{\rho_2} = -\frac{n_{2B} - n_{2R}}{n_{1B} - n_{1R}} \tag{4.5}$$

The focal length of the compound lens [13]  $f_Y$  can be specified as that associated with yellow light. For the component lenses in yellow light,  $\frac{1}{f_{1Y}} = (n_{1Y} - 1)\rho_1$  and  $\frac{1}{f_{2Y}} = (n_{2Y} - 1)\rho_2$ . Hence

$$\frac{\rho_1}{\rho_2} = \frac{(n_{2Y} - 1)f_{2Y}}{(n_{1Y} - 1)f_{1Y}} \tag{4.6}$$

Equating 4.5 and 4.6 we get

$$\frac{f_{2Y}}{f_{1Y}} = -\frac{(n_{2B} - n_{2R})/(n_{2Y} - 1)}{(n_{1B} - n_{1R})/(n_{1Y} - 1)}$$
(4.7)

where the quantities  $(n_{2B} - n_{2R})/(n_{2Y} - 1)$  and  $(n_{1B} - n_{1R})/(n_{1Y} - 1)$  are known as the **dispersive powers** of the two materials forming the lenses and their reciprocals are known as the **Abbe numbers**. With these numbers we are able to look up a table that specifies the correct material for the achromatic doublets. For more information about doublet design please refer to Section 6.3.2 of [13].

The idea of having a full color display is more likely to gain more acceptance from the general consumer rather than a monochromatic display. Therefore the first thing to do is to develop an 8-bit resolution laser projector, increasing the image fidelity and quality. With this upgrade, people from around the world will be able to use the laser projector as part of their normal lives, because not only images but also e-mail messages and any kind of text formats will be displayed. For example, the resolution needed to display Japanese or Chinese characters, to avoid unnecessary misunderstandings, is higher than displaying a text written in English.

The MLPP is a versatile device that has a whole variety of applications and future developments. Venture applications could be applied in medicine where, in a configuration of two MLPPs with perpendicular beams impinging into an active plastic material, the Magnetic Resonance Imaging (MRI) could be displayed in a three-dimensional way. If a 3-D laser projector is created, then many research areas could be explored, from medicine, architecture and entertainment to aeronautical modelling.

The MLPP can be considered as an input device. Using the time of flight technology, the MLPP can be used as a 3-D scanner. With the same technology, the MLPP can perform keystone auto correction. Taking advantage of the laser sensor, the surface where the video is projected (i.e., the wall, the floor, or a table) can be used as a touch screen display.

It is important to remember that the goal for the MLPP is low power consumption, low cost, and a size small enough to make it portable and run on batteries. With these characteristics, it someday may substitute for the Liquid Crystal Displays of the lap-tops and PDA's. The MLPP is not far for a future implementation for a hand-held device, where it can gain a final size of 2 cm x 2cm x 2cm. If the MLPP gets market acceptance, VCSELs will become very cheap. The series production of MLPP will make all of its components even cheaper making an total estimate of 20 to 50 USD. Eye safety is one of the most important things to consider in all applications described before. Safety circuitry has to be designed to avoid direct eye contact. This last feature will make the MLPP a reliable device which no one wants to live without it, with a broad acceptance on the market.

# Appendix A

# Hardware Specification Data Sheets

### A.0.2 Analog to Digital Converter Data Sheet

ADC DATA SHEET can be found on the National web site at http://www.national.com/pf/AD/ADC08100.html

### A.0.3 FPGA Board Description

The FPGA board design can be found at the following Digilent web site http://www.digilentinc.com/Catalog/digilab\_2e.html

# Appendix B

# Software Source

The following sections will describe the code source for the MLPP. Note that the subindexes are shifted. At a similar fashion the names were changed. Here in this appendix the name of the modules are as they appear originally on the FPGA.

#### B.0.4 Main Driver 0

The original name of this submodule is "mainT0 sch". The code is shown below: library IEEE; use IEEE.STD LOGIC 1164.ALL; use IEEE.STD LOGIC ARITH.ALL; use IEEE.STD LOGIC UNSIGNED.ALL; -This is the main driver that controls all 50 lasers. At time T0, pixels p1, p5, p9, p13, p17 etc... are lit at this time. Buffer T0 will gather all the information for these pixels entity mainT0 sch is Port ( contador : in std ulogic; -internal counter rst : in std ulogic; -async reset input hsyncin : in std ulogic; -horizontal sync input vsyncin : in std ulogic; -vertical sync input rin0 : in std ulogic; -MSB for Red input

rin1 : in std ulogic; -LSB for Red input (out of 2 bits)

red0, red1, red2, red3, red4, red5, red6, red7,

red8, red9, red10, red11, red12, red13, red14,

red15, red16, red17, red18, red19, red20, red21,

red22, red23, red24, red25, red26, red27, red28,

red29, red30, red31, red32, red33, red34, red35,

red36, red37, red38, red39, red40, red41, red42,

red43, red44, red45, red46, red47, red48, red49,

red50, red51, red52, red53, red54, red55, red56,

red57, red58, red59, red60, red61, red62, red63,

red64, red65, red66, red67, red68, red69, red70,

red71, red72, red73, red74, red75, red76, red77,

red78, red79, red80, red81, red82, red83, red84,

red85, red86, red87, red88, red89, red90, red91,

red92, red93, red94, red95, red96, red97, red98,

red99 : out std ulogic; -all red outputs (50)

BE : out std ulogic; -enables all buffers

SME : out std ulogic; -enables State Machines

rout0, rout1 : out std ulogic; -vga red components

hsyncout : out std ulogic; -horizontal sync output vsyncout : out std ulogic

#### );

end mainT0 sch;

-Passtrhough

```
hsyncout \leq hsyncin;
vsyncout <= vsyncin;
rout0 \le rin0;
rout1 \leq rin1;
counter: Process(contador) – This is the counter process
begin
if (rst = '1') then
hcnt <= "0000000000";
- horiz. pixel counter increments on rising edge of dot clock
elsif (contador'event and contador='1') then
- horiz. pixel counter rolls-over after 1400 clock cycles
if hcnt < 1400 then
hcnt \leq hcnt + 1;
else
hcnt <= "0000000000";
end if;
end if;
end process;
Laser0: Process(hcnt)
begin
if (hent = 100) then -Laser 0 procedure
red0 \ll rin0;
red1 \leq rin1;
end if;
```

```
end process;
*
*
*
Laser49: Process(hcnt)
begin
if (hent = 492) then -Laser 49 procedure
red98 \le rin0;
red99 \le rin1;
end if;
end process;
Enable: Process(hcnt)
begin
if (hent = 492 \text{ or } hent = 493) then -Buffer Enable
BE <=' 1';
else
BE <=' 0';
end if;
end process;
SMEnable: Process(hcnt)
begin
if (hent = 98 \text{ or } hent = 99) then -Tri State Enable
SME \le '1'; - and Clock Enable for SME
else
SME <=' 0';
end if;
```

end process; end Behavioral;

#### B.0.5 Main Driver 1

The original name of this submodule is "mainT1 sch". The code is shown below: library IEEE;

use IEEE.STD LOGIC 1164.ALL;

use IEEE.STD LOGIC ARITH.ALL;

use IEEE.STD LOGIC UNSIGNED.ALL;

-This is the main driver that controls all 50 lasers.

At time T1, pixels p2, p6, p10, p14, p18 etc... are lit at this

time. Buffer T1 will gather all the information for these pixels entity mainT1 sch is

Port (

contador : in std ulogic; -internal counter

rst : in std ulogic; -async reset input

hsyncin : in std ulogic; -horizontal sync input

vsyncin : in std ulogic; -vertical sync input

rin0 : in std ulogic; -MSB for Red input

rin1 : in std ulogic; -LSB for Red input (out of 2 bits)

red0, red1, red2, red3, red4, red5, red6, red7,

red8, red9, red10, red11, red12, red13, red14,

red15, red16, red17, red18, red19, red20, red21,

red22, red23, red24, red25, red26, red27, red28,

red29, red30, red31, red32, red33, red34, red35,

red36, red37, red38, red39, red40, red41, red42,

red43, red44, red45, red46, red47, red48, red49,

red50, red51, red52, red53, red54, red55, red56,

red57, red58, red59, red60, red61, red62, red63,

```
red64, red65, red66, red67, red68, red69, red70,
red71, red72, red73, red74, red75, red76, red77,
red78, red79, red80, red81, red82, red83, red84,
red85, red86, red87, red88, red89, red90, red91,
red92, red93, red94, red95, red96, red97, red98,
red99 : out std ulogic; -all red outputs (50)
BE : out std ulogic; -enables all buffers
SME : out std ulogic -enables State Machines
);
end mainT1 sch;
architecture Behavioral of mainT1 sch is -Internal Signal
signal hcnt: std logic vector(10 \text{ downto } 0);
begin
counter: Process(contador) – This is the counter process
begin
if (rst = '1') then
hcnt <= "00000000000";
- horiz. pixel counter increments on rising edge of dot clock
elsif (contador'event and contador='1') then
- horiz. pixel counter rolls-over after 1400 clock cycles
if hcnt < 1400 then
hcnt \leq hcnt + 1;
else
hcnt <= "00000000000";
end if;
end if;
end process;
```

```
Laser0: Process(hcnt)
begin
if (hent = 102) then -Laser 0 procedure
red0 \ll rin0;
red1 \ll rin1;
end if;
end process;
*
*
*
Laser49: Process(hcnt)
begin
if (hent = 494) then -Laser 49 procedure
red98 \le rin0;
red99 \le rin1;
end if;
end process;
Enable: Process(hcnt)
begin
if (hent = 494 \text{ or } hent = 495) then -Buffer Enable
BE <=' 1';
else
BE <=' 0';
end if;
end process;
SMEnable: Process(hcnt)
```

```
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```

begin
if (hent = 415 or hent = 416) then -Tri State Enable
SME <= '1'; - and Clock Enable for SME
else
SME <=' 0';
end if;
end process;
end Behavioral;</pre>

#### B.0.6 Main Driver 2

The original name of this submodule is "**mainT2 sch**". The code is shown below: library IEEE;

use IEEE.STD LOGIC 1164.ALL;

use IEEE.STD LOGIC ARITH.ALL;

use IEEE.STD LOGIC UNSIGNED.ALL;

-This is the main driver that controls all 50 lasers.

At time T1, pixels p3, p7, p11, p15, p19 etc... are lit at this

time. Buffer T2 will gather all the information for these pixels

entity mainT2 sch is

Port (

contador : in std ulogic; -internal counter

rst : in std ulogic; -async reset input

hsyncin : in std ulogic; -horizontal sync input

vsyncin : in std ulogic; -vertical sync input

rin0 : in std ulogic; -MSB for Red input

rin1 : in std ulogic; -LSB for Red input (out of 2 bits)

red0, red1, red2, red3, red4, red5, red6, red7,

red8, red9, red10, red11, red12, red13, red14,

red15, red16, red17, red18, red19, red20, red21,

```
red22, red23, red24, red25, red26, red27, red28,
red29, red30, red31, red32, red33, red34, red35,
red36, red37, red38, red39, red40, red41, red42,
red43, red44, red45, red46, red47, red48, red49,
red50, red51, red52, red53, red54, red55, red56,
red57, red58, red59, red60, red61, red62, red63,
red64, red65, red66, red67, red68, red69, red70,
red71, red72, red73, red74, red75, red76, red77,
red78, red79, red80, red81, red82, red83, red84,
red85, red86, red87, red88, red89, red90, red91,
red92, red93, red94, red95, red96, red97, red98,
red99 : out std ulogic; -all red outputs (50)
BE : out std ulogic; -enables all buffers
SME : out std ulogic -enables State Machines
);
end mainT2 sch;
architecture Behavioral of mainT2 sch is -Internal Signal
signal hcnt: std logic vector(10 downto 0);
begin
counter: Process(contador) – This is the counter process
begin
if (rst = '1') then
hcnt <= "00000000000";
- horiz. pixel counter increments on rising edge of dot clock
elsif (contador'event and contador='1') then
- horiz. pixel counter rolls-over after 1400 clock cycles
if hcnt < 1400 then
hcnt \leq hcnt + 1;
```

```
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```

```
else
hcnt <= "0000000000";
end if;
end if;
end process;
Laser0: Process(hcnt)
begin
if (hent = 104) then -Laser 0 procedure
red0 <= rin0;
red1 \leq rin1;
end if;
end process;
*
*
Laser49: Process(hcnt)
begin
if (hent = 496) then -Laser 49 procedure
red98 \leq rin0;
red99 \le rin1;
end if;
end process;
Enable: Process(hcnt)
begin
if (hent = 496 \text{ or } hent = 497) then -Buffer Enable
BE <=' 1';
```

#### B.0.7 Main Driver 3

The original name of this submodule is "mainT3 sch". The code is shown below: library IEEE;

```
use IEEE.STD LOGIC 1164.ALL;
```

use IEEE.STD LOGIC ARITH.ALL;

use IEEE.STD LOGIC UNSIGNED.ALL;

-This is the main driver that controls all 50 lasers.

At time T1, pixels p4, p8, p12, p16, p20 etc... are lit at this

time. Buffer T3 will gather all the information for these pixels

entity mainT3 sch is

Port (

contador : in std ulogic; -internal counter

rst : in std ulogic; -async reset input

hsyncin : in std ulogic; -horizontal sync input

```
vsyncin : in std ulogic; -vertical sync input
rin0 : in std ulogic; –MSB for Red input
rin1 : in std ulogic; -LSB for Red input (out of 2 bits)
red0, red1, red2, red3, red4, red5, red6, red7,
red8, red9, red10, red11, red12, red13, red14,
red15, red16, red17, red18, red19, red20, red21,
red22, red23, red24, red25, red26, red27, red28,
red29, red30, red31, red32, red33, red34, red35,
red36, red37, red38, red39, red40, red41, red42,
red43, red44, red45, red46, red47, red48, red49,
red50, red51, red52, red53, red54, red55, red56,
red57, red58, red59, red60, red61, red62, red63,
red64, red65, red66, red67, red68, red69, red70,
red71, red72, red73, red74, red75, red76, red77,
red78, red79, red80, red81, red82, red83, red84,
red85, red86, red87, red88, red89, red90, red91,
red92, red93, red94, red95, red96, red97, red98,
red99 : out std ulogic; -all red outputs (50)
BE : out std ulogic; -enables all buffers
SME : out std ulogic –enables State Machines
);
end mainT3 sch;
architecture Behavioral of mainT3 sch is -Internal Signal
signal hcnt: std logic vector(10 \text{ downto } 0);
begin
counter: Process(contador) – This is the counter process
begin
if (rst = '1') then
```

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hcnt <= "0000000000";

```
- horiz. pixel counter increments on rising edge of dot clock
elsif (contador'event and contador='1') then
- horiz. pixel counter rolls-over after 1400 clock cycles
if hcnt < 1400 then
hcnt \leq hcnt + 1;
else
hcnt <= "00000000000";
end if;
end if;
end process;
Laser0: Process(hcnt)
begin
if (hent = 106) then -Laser 0 procedure
red0 <= rin0;
red1 <= rin1;
end if;
end process;
*
Laser49: Process(hcnt)
begin
if (hent = 498) then -Laser 49 procedure
red98 <= rin0;
red99 \le rin1;
end if;
```

```
end process;
Enable: Process(hcnt)
begin
if (hent = 498 \text{ or } hent = 499) then -Buffer Enable
BE <=' 1';
else
BE <=' 0';
end if;
end process;
SMEnable: Process(hcnt)
begin
if (hent = 1049 \text{ or } hent = 1050) then -Tri State Enable
SME \le '1'; - and Clock Enable for SME
else
SME <=' 0';
end if;
end process;
end Behavioral;
```

#### **B.0.8** Three-State Buffer

In this section the three-state buffer was divided into two submodules which are the "buffer" and the "three-state" submodules. The reason why it was chosen that way was because it was better to handle the variables by separate. The submodule "buffer" works as a local memory and stores the data information until the "BE" (Buffer Enable) signal triggers the buffer and transfers the data to the three-state submodule. This last submodule is like a stand by memory which is always at high impedance until SME (State Machine Enable) signal triggers this submodule and

transfers the data to the laser actuator (laser act).

#### Buffer (Buf50L)

library IEEE; use IEEE.STD LOGIC 1164.ALL; use IEEE.STD LOGIC ARITH.ALL; use IEEE.STD LOGIC UNSIGNED.ALL;

entity Buf50L is

#### Port (

BE : in std ulogic; -Enables Buffer rin0, rin1, rin2, rin3, rin4, rin5, rin6, rin7, rin8, rin9, rin10, rin11, rin12, rin13, rin14, rin15, rin16, rin17, rin18, rin19, rin20, rin21, rin22, rin23, rin24, rin25, rin26, rin27, rin28, rin29, rin30, rin31, rin32, rin33, rin34, rin35, rin36, rin37, rin38, rin39, rin40, rin41, rin42, rin43, rin44, rin45, rin46, rin47, rin48, rin49 : in std ulogic; -Buffer inputs rout0, rout1, rout2, rout3, rout4, rout5, rout6, rout7, rout8, rout9, rout10, rout11, rout12, rout13, rout14, rout15, rout16, rout17, rout18, rout19, rout20, rout21, rout22, rout23, rout24, rout25, rout26, rout27, rout28, rout29, rout30, rout31, rout32, rout33, rout34, rout35, rout36, rout37, rout38, rout39, rout40, rout41, rout42, rout43, rout44, rout45, rout46, rout47, rout48, rout49 : out std ulogic -Buffer outputs

);

end Buf50L;

```
architecture Behavioral of Buf50L is
```

begin

```
Process(BE)

begin

if BE = '1' then

rout0 <= rin0;

rout1 <= rin1;

*

*

rout48 <= rin48;

rout49 <= rin49;

end if;

end process;

end Behavioral;
```

#### Three State Buffer (Tristatebuff)

library IEEE; use IEEE.STD LOGIC 1164.ALL; use IEEE.STD LOGIC ARITH.ALL; use IEEE.STD LOGIC UNSIGNED.ALL; entity tristatebuff is Port ( SME : in std ulogic; -**Enables Buffer** rin0, rin1, rin2, rin3, rin4, rin5, rin6, rin7,

rin8, rin9, rin10, rin11, rin12, rin13, rin14, rin15, rin16, rin17, rin18, rin19, rin20, rin21, rin22, rin23, rin24, rin25, rin26, rin27, rin28, rin29, rin30, rin31, rin32, rin33, rin34, rin35, rin36, rin37, rin38, rin39, rin40, rin41, rin42, rin43, rin44, rin45, rin46, rin47, rin48, rin49 : in std ulogic; -Buffer inputs rout0, rout1, rout2, rout3, rout4, rout5, rout6, rout7, rout8, rout9, rout10, rout11, rout12, rout13, rout14, rout15, rout16, rout17, rout18, rout19, rout20, rout21, rout22, rout23, rout24, rout25, rout26, rout27, rout28, rout29, rout30, rout31, rout32, rout33, rout34, rout35, rout36, rout37, rout38, rout39, rout40, rout41, rout42, rout43, rout44, rout45, rout46, rout47, rout48, rout49 : out std ulogic -Buffer outputs );

end tristatebuff;

architecture Behavioral of tristatebuff is

```
begin Process(SME)
begin
if SME = '1' then
rout0 <= rin0;
rout1 <= rin1;
*
*
rout48 <= rin48;
rout49 <= rin49;</pre>
```

elserout0 <=' Z';rout1 <=' Z';\* \* \* rout48 <=' Z';rout49 <=' Z';end if; end process; end Behavioral;

#### **B.0.9** Laser Actuator

Better known as "laser act", this is how the state machines were declared. library IEEE; use IEEE.STD LOGIC 1164.ALL; use IEEE.STD LOGIC ARITH.ALL; use IEEE.STD LOGIC UNSIGNED.ALL; entity laser act is Port ( clk : in std ulogic; -Clock rst : in std ulogic; -Async reset rin0, rin1, rin2, rin3, rin4, rin5, rin6, rin7, rin8, rin9, rin10, rin11, rin12, rin13, rin14, rin15, rin16, rin17, rin18, rin19, rin20, rin21, rin22, rin23, rin24, rin25, rin26, rin27, rin28, rin29, rin30, rin31, rin32, rin33, rin34, rin35, rin36, rin37, rin38, rin39, rin40, rin41, rin42, rin43, rin44, rin45, rin46, rin47, rin48, 68

rin49 : in std ulogic; -Laser intensity MSB

SME : in std ulogic; -State Machine Enable led0, led1, led2, led3, led4, led5, led6, led7, led8, led9, led10, led11, led12, led13, led14, led15, led16, led17, led18, led19, led20, led21, led22, led23, led24 : out std ulogic -Laser outputs ); end laser act; architecture Behavioral of laser act is signal hcnt: std logic vector(8 downto 0);

begin

counter: Process(clk) – This is the counter process begin

if (rst = '1') then

hcnt  $\leq$  "000000000";

- horiz. pixel counter increments on rising edge of dot clock elsif (clk'event and clk='1') then

- horiz. pixel counter rolls-over after 316 clock cycles

if hcnt<316 then

#### -State Machines for all lasers;

#### \*\*\*\*\*\*\*

#### -Laser 0

laser0: process(clk, hcnt, rin0, rin1, SME)

type state type is (High, Medium, Low, Off); -Declare Laser States

variable laser state: state type;

begin

if (rin0 = '0' and rin1 = '0' and SME = '1') then -When the red inputs are in low

 $led0 \le '0'$  after 1 ns; -The laser is then Off

laser state := Off;

-default conditions

elsif (clk'event and clk = '1') then

 $led0 \le '0'$  after 1 ns;

-Define state transitions, note that end if is at the very end!

case laser state is

when Off =>

#### -Transition from Off to High

if (rin0 = '1' and rin1 = '1' and SME = '1') then

 $led0 \le 1$  after 1 ns;

laser state := High;

#### -Transition from Off to Medium

elsif (rin0 = 1 and rin1 = 0 and SME = 1) then

 $led0 \le '1'$  after 1ns;

laser state := Medium;

#### -Transition from Off to Low

elsif (rin0 = 0' and rin1 = 1' and SME = 1') then

 $led0 \le '1'$  after 1ns;

laser state := Low; -Holding term in off

else

laser state := Off; end if; -This state makes the Laser be on for 314 clock cycles because is at High state when High => -Transition from High to Off if (hcnt> 314) then

 $led0 \le '0'$  after 1ns;

laser state := Off;

-Holding term in High

else

 $led0 \le '1'$  after 1ns;

laser state := High;

end if;

-This state makes the Laser be on for 157 clock cycles because is at

Medium state

when Medium =>

-Transition from Medium to Off

if (hcnt>157) then

 $led0 \le '0'$  after 1ns;

laser state := Off;

-Holding term in Medium

else

 $led0 \le '1'$  after 1ns;

laser state := Medium;

end if;

-This state makes the Laser be on for 79 clock cycles because is at Low

state

when Low =>

-Transition from Low to Off

```
if (hcnt>79) then
led0 \le '0' after 1ns;
laser state := Off;
-Holding term in Low
else
led0 \le '1' after 1ns;
laser state := Low;
end if;
-Default term for dead states
when others =>
laser state := Off;
end case;
end if;
end process laser0;
*
*
```

#### -Laser 24

```
laser24: process(clk, hcnt, rin48, rin49, SME)
type state type is (High, Medium, Low, Off); -Declare Laser States
variable laser state: state type;
begin
if (rin48 = '0' and rin49 = '0' and SME = '1') then -When the red inputs are in
low
led24 <= '0' after 1 ns; -The laser is then Off
laser state := Off;
-default conditions
elsif (clk'event and clk = '1') then</pre>
```

 $led24 \ll 0$  after 1 ns;

#### -Define state transitions, note that end if is at the very end!

case laser state is

when Off =>

#### -Transition from Off to High

if (rin48 = '1' and rin49 = '1' and SME = '1') then

 $led24 \le '1'$  after 1 ns;

laser state := High;

#### -Transition from Off to Medium

elsif (rin48 = '1' and rin49 = '0' and SME = '1') then

 $led24 \ll '1'$  after 1ns;

laser state := Medium;

#### -Transition from Off to Low

elsif (rin48 = 0' and rin49 = 1' and SME = 1') then

 $led24 \ll '1'$  after 1ns;

laser state := Low; -Holding term in off

else

```
laser state := Off;
```

end if;

### -This state makes the Laser be on for 314 clock cycles because is at High

state

when High =>

#### -Transition from High to Off

if (hcnt > 314) then

 $led24 \le '0'$  after 1ns;

laser state := Off;

#### -Holding term in High

else

 $led24 \ll 1'$  after 1ns;

laser state := High;

end if;

-This state makes the Laser be on for 157 clock cycles because is at Medium state when Medium =>-Transition from Medium to Off if (hcnt>157) then  $led24 \le '0'$  after 1ns; laser state := Off; -Holding term in Medium else  $led24 \ll 1'$  after 1ns; laser state := Medium; end if; -This state makes the Laser be on for 79 clock cycles because is at Low state when Low =>-Transition from Low to Off if (hcnt>79) then  $led24 \leq 0$  after 1ns; laser state := Off; -Holding term in Low else  $led24 \leq 1'$  after 1ns; laser state := Low; end if; -Default term for dead states when others =>laser state := Off; end case; end if; end process laser24;

end Behavioral;

#### B.0.10 FPGA Pin out

NET "clkout" LOC = "P64"; NET "vsyncout" LOC = "P62"; NET "vsyncin" LOC = "P189"; NET "rout1" LOC = "P74"; NET "rout0" LOC = "P73"; NET "rin1" LOC = "P192"; NET "rin0" LOC = "P193"; NET "led49" LOC = "P154"; NET "led48" LOC = "P160"; NET "led47" LOC = "P161"; NET "led46" LOC = "P162"; NET "led45" LOC = "P163"; NET "led44" LOC = "P164"; NET "led43" LOC = "P165"; NET "led42" LOC = "P166"; NET "led41" LOC = "P167"; NET "led40" LOC = "P168"; NET "led39" LOC = "P169"; NET "led38" LOC = "P173"; NET "led37" LOC = "P174"; NET "led36" LOC = "P175"; NET "led35" LOC = "P176"; NET "led34" LOC = "P178"; NET "led33" LOC = "P179"; NET "led32" LOC = "P108"; NET "led31" LOC = "P102";

NET "led30" LOC = "P110"; NET "led29" LOC = "P109"; NET "led28" LOC = "P112"; NET "led27" LOC = "P111"; NET "led26" LOC = "P114" : NET "led25" LOC = "P113"; NET "led24" LOC = "P116"; NET "led23" LOC = "P115": NET "led22" LOC = "P121"; NET "led21" LOC = "P120"; NET "led20" LOC = "P123": NET "led19" LOC = "P122"; NET "led18" LOC = "P125"; NET "led17" LOC = "P81"; NET "led16" LOC = "P75"; NET "led15" LOC = "P83": NET "led14" LOC = "P82" : NET "led13" LOC = "P86"; NET "led12" LOC = "P84"; NET "led11" LOC = "P88"; NET "led10" LOC = "P87"; NET "led9" LOC = "P93"; NET "led8" LOC = "P89"; NET "led7" LOC = "P95"; NET "led6" LOC = "P94"; NET "led5" LOC = "P97": NET "led4" LOC = "P96"; NET "led3" LOC = "P99"; NET "led2" LOC = "P98"; NET "led1" LOC = "P101";

- NET "integra" LOC = "P152";
- NET "SME3" LOC = "P58";
- NET "SME2" LOC = "P59";
- NET "SME1" LOC = "P60";
- NET "SME0" LOC = "P61";
- NET "clock" LOC = "P80";
- NET "hsyncin" LOC = "P63";
- NET "hsyncout" LOC = "P70";
- NET "led0" LOC = "P100";

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