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Broadband surface-wave transformation cloak

Su Xu,a,b,1 Hongyi Xu,c,d,1 Hanhong Gao,e Yuyu Jiang,a,b, Faxin Yu,f, John D. Joannopoulosg,h,2 Marin Soljačićg, Hongsheng Chen,a,b,2, Handong Sun,c,d,2 and Baile Zhangc,d,2

*State Key Laboratory of Modern Optical Instrumentation, Zhejiang University, Hangzhou 310027, China; †The Electromagnetics Academy at Zhejiang University, Department of Information Science and Electronic Engineering, Zhejiang University, Hangzhou 310027, China; ‡Division of Physics and Applied Physics, School of Physical and Mathematical Sciences, Nanyang Technological University, Singapore 637371; ‡Centre for Disruptive Photonic Technologies, Nanyang Technological University, Singapore 637371; †Department of Electrical Engineering and Computer Science, Massachusetts Institute of Technology, Cambridge, MA 02139; School of Aeronautics and Astronautics, Zhejiang University, Hangzhou 310027, China; and ‡Research Laboratory of Electronics, Massachusetts Institute of Technology, Cambridge, MA 02139

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Guiding surface electromagnetic waves around disorder without disturbing the wave amplitude or phase is in great demand for modern photonic and plasmonic devices, but is fundamentally difficult to realize because light momentum must be conserved in a scattering event. A particular realization has been achieved by exploiting topological electromagnetic surface states, but this approach is limited to narrow-band light transmission and subject to phase disturbances in the presence of disorder. Recent advances in transformation optics apply principles of general relativity to curve the space for light, allowing one to match the momentum and phase of light around any disorder as if that disorder were not there. This feature has been exploited in the development of invisibility cloaks. An ideal invisibility cloak, however, would require the phase velocity of light being guided around the cloaked object to exceed the vacuum speed of light—a feat potentially achievable only over an extremely narrow band. In this work, we theoretically and experimentally show that the bottlenecks encountered in previous studies can be overcome. We introduce a class of cloaks capable of remarkable broadband surface electromagnetic waves guidance around ultrasharp corners and bumps with no perceptible changes in amplitude and phase. These cloaks consist of specifically designed nonmagnetic metamaterials and achieve nearly ideal transmission efficiency over a broadband frequency range from 0 to 6 GHz. This work provides strong support for the application of transformation optics to plasmonic circuits and could pave the way toward high-performance, large-scale integrated photonic circuits.

transformation optics | surface wave | invisibility cloaks | broadband

Significance

Guiding surface electromagnetic waves around disorder without disturbing the wave amplitude or phase is in great demand for modern photonic and plasmonic devices. In this work, we introduce a class of cloaks capable of remarkable broadband surface electromagnetic waves guidance around ultrasharp corners and bumps with no perceptible changes in amplitude and phase. This work provides strong support for the application of transformation optics to plasmonic circuits and could pave the way for high-performance, large-scale integrated photonic circuits.
ideal transmission, which allows “invisibility cloaking” of disorders, such as ultrasharp corners and bumps for surface electromagnetic waves. The remarkable broadband guidance, lacking in previous scattering-free topological electromagnetic surface states and free-space invisibility cloaks, is because of two reasons. First, the slow-wave property of surface electromagnetic waves can overcome the bottleneck of free-space invisibility cloaks, because its phase velocity does not need to exceed the vacuum speed of light. Second, we adopt a fully nonmagnetic design with naturally accessible dielectric parameters, whereas magnetic responses were necessary to open the topological band gap for topological electromagnetic surface states.

We start with the demonstration of bending a surface electromagnetic wave across sharp right-angle corners at microwave frequencies—similar to the previous demonstration of bending a guided topological surface electromagnetic wave in a photonic crystal (3). We call the bending adapter a “corner cloak,” because it effectively hides a corner to the wave as if the corner did not exist. Because metals at microwave frequencies are perfect electric conductors that generally do not support surface electromagnetic waves, here we adopt the approach of geometrically induced or spoof surface plasmons (31) [i.e., we use a grooved metallic surface (referred to as a patterned metal in Figs. 1–3) to support surface electromagnetic waves in the microwave regime]. Fig. 1A shows the experimental setup: a U-shaped surface-wave waveguide (a metal base with periodic grooves on its surfaces; i.e., the patterned metal) with two right-angle zero-radius corners. Given the sizes of the cloaks, the more confined the surface waves on the dielectric-metal interface, the better the cloaking performance. To enhance the confinement, we load the grooves with ceramic material with permittivity \( \varepsilon_{\text{ceramic}} = 21 \). We use glass with permittivity \( \varepsilon_{g} = 4.6 \) as the surrounding background that is to be impedance-matched with the cloaks. More details can be found in Fig. S1 and SI Methods.

The two identical corner cloaks locate at the two corners. This structure that consists of the U-shaped waveguide and two corner cloaks can be thought of as if it were transformed from a straight waveguide without any corner. A corner cloak, when transformed back, corresponds to a triangular space on top of the dielectric-metal interface, which has an area that is purposely chosen to be the same as the corner cloak. This area preserves nonmagnetism in the cloak design for surface electromagnetic waves (more details are in Fig. S2 and SI Methods). The two identical corner cloaks require anisotropic constitutive parameters. For each cloak, the required principal permittivities in two orthogonal directions, \( \varepsilon_{x} \) and \( \varepsilon_{y} \), after the procedure of diagonalization, where only components in the \( y \) plane are relevant, are \( \varepsilon_{x} = 10.7 \) and \( \varepsilon_{y} = 2.0 \). These cloak materials were implemented with a metamaterial consisting of a stack of the following two materials with subwavelength thicknesses: a microwave dielectric ceramic with permittivity \( \varepsilon_{\text{ceramic}} = 21 \) (WuxiXiaoying K-21; loss tangent: \( 1 \times 10^{-4} \); 1-mm thickness) and a polymer foam with permittivity \( \varepsilon_{\text{foam}} = 1.1 \) (Rohacell 71HF; loss tangent: \( 16 \times 10^{-6} \); 1.06-mm thickness). According to the standard formulas of effective medium theory, one can get

\[
\begin{align*}
\varepsilon_{x} & = \varepsilon_{\text{ceramic}} + (1 - r) \varepsilon_{\text{foam}} \\
\varepsilon_{y} & = (1 - r) \varepsilon_{\text{ceramic}} + r \varepsilon_{\text{foam}}
\end{align*}
\]

where the filling factor is given by \( r = 0.485 \). Fig. 1B shows the simulation of the transmission of surface electromagnetic waves when the corners are not cloaked by the corner cloaks; a dramatic scattering loss is evident. However, the transmission of surface electromagnetic waves across a sharp corner is perfect when both corners are cloaked by corner cloaks (Fig. 1C).

A fabricated model with two corner cloaks is shown in Fig. 1D. For comparison, we also fabricated a straight waveguide with similar grooves and the same total propagation distance. The transmission data measured on the U-shaped surface-wave waveguide from 0° (100 MHz) to 6 GHz without/with corner cloaks are normalized to the transmission data measured on the straight waveguide (Fig. 1E). Without corner cloaks, the transmission measured at the output of the U-shaped waveguide is close to zero, but when both sharp corners are hidden by the corner cloaks, the transmission is almost unity. The experimental result shows near-perfect cloaking of two right-angle zero-radius corners for surface electromagnetic waves in a broad bandwidth from 0° to 6 GHz (i.e., with a fractional bandwidth of 200%).

Next, we show a surface-wave carpet cloak used to cover an ultrasharp bump on a flat metal–dielectric interface. Fig. 2A shows our experimental setup. Like in the realization of the corner cloaks, we use a metal base with grooves to support surface electromagnetic waves. A sharp bump on the flat surface acts as an obstacle able to block the propagation of surface electromagnetic waves. The carpet cloak that can hide this sharp bump was designed with a similar transformation optics approach (more details can be found in Fig. S3 and SI Methods). The numerical simulation for the real structure in the presence of a sharp bump is shown in Fig. 2B and C: without a carpet cloak, most of the wave energy is scattered into the background medium near the apex of the bump; when the carpet cloak is put on top of the bump, however, the electromagnetic surface waves can be smoothly guided around the bump and returned to their original path as if the bump was not there. A fabricated model with a carpet cloak implemented with the same metamaterial used in the corner cloaks is shown in Fig. 2D. The measured transmissions, normalized to the

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**Fig. 1.** Surface-wave bending around sharp corners. (A) A U-shaped surface-wave waveguide with grooves on its surface covered by glass is illuminated by a dipole antenna. The two sharp corners are covered by two corner cloaks (i.e., layered structures with subwavelength foam and ceramic materials). A second dipole antenna located at the output of the waveguide measures the transmission. (B) Simulation of a surface wave when it encounters a sharp corner that is not covered by a cloak. (C) Simulation of a surface wave when the sharp corner is cloaked by a corner cloak. (D) Photo of a fabricated model. The transmitter is shielded by the microwave absorber material. (E) Measured normalized transmission of surface waves through the waveguide. Exp., experimental data; Sim., simulation data.
transmission through a straight waveguide without the bump, are shown in Fig. 2E. For the setup without a cloak, the normalized transmission is close to zero, indicating that the propagation of surface electromagnetic waves has been blocked by the bump. For the case with a carpet cloak, the normalized transmission approaches unity, showing near-perfect cloaking of a sharp bump for surface electromagnetic waves in a broad bandwidth from 0 GHz to 6 GHz.

A striking feature, absent in topological electromagnetic surface states, is that, when the surface waves are perfectly guided by the cloaks, the phase is preserved. We used a pulsed signal to show this behavior. Fig. 3 shows the dynamic propagation of a pulse through the cloaks, which was obtained with the commercial software COMSOL Multiphysics. A point source at port 1 excites a Gaussian-shaped pulse (bandwidth: 0 GHz to 6 GHz; center frequency: 3 GHz) at 0 ns. The magnetic

**Fig. 2.** Surface-wave carpet cloaking. (A) A straight surface-wave waveguide with a sharp bump is illuminated by a dipole antenna. The surface of the metal base is grooved similarly as that in Fig. 1A. The sharp bump is covered by a carpet cloak (i.e., a layered structure with subwavelength foam and ceramic materials). A second dipole antenna located at the output of the waveguide measures the transmission. (B) Simulation of a surface wave when it encounters the sharp bump without a carpet cloak. (C) Simulation of a surface wave when the sharp bump is cloaked by the carpet cloak. (D) Photo of a fabricated model. The transmitter is shielded by the microwave absorber material. (E) Measured normalized transmission of surface waves through the waveguide. Exp., experimental data; Sim., simulation data.

**Fig. 3.** A Gaussian-shaped pulse propagates on the patterned metal. A point source (port 1) generates the pulse at 0 ns. The bandwidth of the pulse is 6 GHz, and the center frequency is 3 GHz. The magnetic field distributions for three cases [(A) the corner cloaks, (B) the carpet cloak, and (C) the straight waveguide reference] are plotted to show the propagation of the pulse at five equivalent temporal sampling points.
field distributions are plotted to show the propagation of the pulse on the patterned metal for the setup with the corner cloaks (Fig. 3A), the carpet cloak (Fig. 3B), and a straight waveguide as a reference (Fig. 3C). For the realization with corner cloaks, the signal reaches the first and second sharp corners at 1.88 and 3.56 ns, respectively. At both sharp corners, the pulse signal is perfectly guided by the corner cloak, and at last, it leaves the patterned metal from port 2. In the case of the carpet cloak, the pulse reaches the bump at 2.68 ns, and it is guided smoothly across the bump by the carpet cloak without any loss. The pulse reaches the same positions as in the straight waveguide, with no relative delay, indicating that the phase is well-preserved in a broad bandwidth by the cloaks. Movies S1 and S2 show more details of the propagating pulse. Fig. 4 shows the measured phase for the corner cloak (Fig. 4A) and carpet cloak (Fig. 4B). The curves almost coincide with their references over the frequency band from 0 to 6 GHz, confirming that the phase of the surface wave is well-preserved by the cloaks.

The above results show scattering-free guidance of surface electromagnetic waves around large disorders, with both wave energy and phase undisturbed in a 200% broad-frequency band.

Fig. 4. Phase measurements. (A) The corner cloaks. (B) The carpet cloak. The phases with the straight waveguide are plotted for reference. In both cases, the phase curves almost coincide with their references over the frequency band from 0 to 6 GHz.

Switching from free-space electromagnetic waves to surface electromagnetic waves, transformation cloaks can find immediate applications without any fundamental limitations. The fully nonmagnetic design makes it feasible to further extend to higher frequencies and/or conventional surface waves. Our work, thereby, paves the way for the next generation of photonic and plasmonic devices, allowing for flexible design without concern in disorders.

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