Evaluation of the Performance of Coherent Optical Communications Commercial DSP ASICs in Low Earth Orbit Radiation Environments

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ABSTRACT

Coherent optical communications systems on satellites have the potential to contribute meeting to world-wide data capacity demand. Digital signal processing **(DSP)** application specific integrated circuits (ASICs) for coherent optical communications systems, first developed in **2008** with current capabilities of over **100** Gbps for commercial terrestrial applications, are a key technology needed for spacebased applications. However, in order to develop coherent optical communications systems for space applications, the performance of these commercial ASICs must be evaluated with consideration of the radiation effects from the space environment. This work investigates the performance of the Inphi **CL2001OA1** optical coherent **DSP** ASIC in a low earth orbit **(LEO)** radiation environment and assesses whether this ASIC could be a viable option for a coherent optical communications system on a **LEO** spacecraft.

The approach consists of simulation and experiment. First the radiation environment is modeled for three sample **LEO** orbits: International Space Station (ISS) orbit, **800** km polar orbit, and 1000 km 0[°] inclination, for 1-year, 5-year, and 10-year mission durations. Total ionizing dose (TID) requirements were determined for each mission and used to experimentally evaluate the TID tolerance of the **CL20010A1.** The **CL2001OA1** on an evaluation board system (EVK) is modeled and simulations with Stopping Range In Matter (SRIM) program are used to simulate 64.0 MeV protons penetrating through the system. The SRIM simulations are used to calculate the proton energy levels entering the silicon active region of the **CL2001OA1** and to determine the proton energy level sufficient for depositing ionizing dose in the active region and penetrating through the active region. The simulations determine the lower threshold of proton energy level needed for experimental testing.

Two proton test campaigns of the CL2001OA1-EVK were completed at **UC** Davis Crocker Nuclear Laboratory **(CNL)** and Tri-University Meson Facility (TRIUMF) using energy levels of 64.0 MeV and 480 MeV, respectively. The **CL2001OA1** ASIC survived and experienced no performance degradation from TID exposure up to **170** krad(Si). The measured **CL20010A1** single event effect **(SEE)** cross section was 2.46×10^{19} cm² at the 64 MeV proton energy level and 3.82×10^{10} cm² at the 480 MeV proton energy level.

The **SEE** cross section data from the proton test campaigns was used to calculate the **CL20010A1 SEE** rate for the sample **LEO** missions. The **SEE** cross section results were compared to the results from previous studies on proton-induced SEEs of other Complementary Metal Oxide Semiconductor **(CMOS)** devices with silicon active regions. Mitigation strategies against **LEO** radiation effects for the **CL2001A1** are considered, such as spot shielding, strategic placement in spacecraft, incorporation of protective electronic devices in the circuit system design, and programming periodic **CL20010A1** resets or full system power cycles. Expansion of this work, such as additional proton radiation test campaigns at different energy levels below 64.0 MeV and between 64.0 MeV and 480 MeV as well as heavy ion test campaigns to assess SEEs induced **by** heavy ions from galactic cosmic rays (GCRs) and solar energetic particles (SEPs), would provide additional insights on potential effects of the **LEO** radiation environment on the **CL20010A1.** Heavy ion test campaigns would provide **SEE** cross section data, which could be used to calculate the expected heavy-ion induced **SEE** rate for a given **LEO** mission. This work serves as an initial step toward the development of a **DSP** coherent optical communications transceiver for **LEO** satellite applications.

Thesis Supervisor: Kerri Cahoy Title: Associate Professor of Aeronautics and Astronautics

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List of Abbreviations

- **ADC -** Analog-to-Digital Converter
- **AFE -** Analog Front End
- ARTEMIS **-** Advanced Relay and TEchnology MIssion Satellite
- ASIC **-** Application Specific Integrated Circuit
- BCB **-** Benzocyclobutene
- BCD **-** Bulk Chromatic Dispersion
- BEOL **-** Back End of Line
- BER **-** Bit Error Rate
- **BJT -** Bipolar Junction Transistor
- BPSK **-** Binary Phase Shift Keying
- BT **-** Bismaleimide Trizxine
- **CCD -** Charged Coupled Device
- **CD -** Chromatic Dispersion
- **CFP2-ACO -** Compact Form Pluggable 2 Analog Coherent Optics
- CL20010Al-EVK **- CL2001OA1 ASIC** Evaluation Board System
- **CME -** Coronal Mass Ejections
- **CMOS -** Complementary Metal Oxide Semiconductor
- **CNL -** Crocker Nuclear Laboratory
- **COTS -** Commercial **Off** The Shelf
- CubeSat **-** Cube Satellite
- **DAC -** Digital-to-Analog Converter
- **DC -** Direct Current
- **DD -** Displacement Damage
- DLR **-** German Aerospace Center
- DP **-** Dual Polarization
- DRAM **-** Dynamic Random Access Memory
- **DSP -** Digital Signal Processing
- **DUT -** Device Under Test
- **EDFA -** Erbium Doped Fiber Amplifier
- **EDC -** Electronic Dispersion Compensation
- ELDRS **-** Enhanced Low Dose Rate Sensitivity
- **ESA -** European Space Agency
- **ETS -** Engineering Test Satellite
- EVK **-** Evaluation Board System
- **FEC -** Forward Error Correction
- **FET -** Field Effect Transistor
- FLARE **-** Freespace Lasercom and Radiation Experiment
- **FPGA -** Field Programmable Gate Array
- *FSOC* **-** Free Space Optical Communications
- Gbps **-** Gigabits Per Second
- **GEO -** Geostationary Earth Orbit
- GCR **-** Galactic Cosmic Ray
- *IC* **-** Integrated Circuit
- ICR **-** Integrated Coherent Receiver
- ILD **-** Interlayer Dielectric
- IP **-** Internet Protocol
- **JAXA -** Japanese Aerospace Exploration Agency
- **JPL -** Jet Propulsion Laboratory
- keV **-** kilo-electron-Volt
- **LADEE -** Lunar Atmosphere and Dust Environment Explorer
- **LCE -** Laser Communications Experiment
- **LCL -** Latch-Up Current Limiter
- LCP **-** Liquid Crystal Polymeric
- **LCT -** Laser Communications Terminals
- **LEO -** Low Earth Orbit
- LET **-** Linear Energy Transfer
- **LLCD -** Lunar Laser Communications Demonstration
- **LLST -** Lunar Laser Space Terminal
- LO **-** Local Oscillator
- LOL **-** Loss of Lock
- **LUCE -** Laser Utilizing Communications Equipment
- MAP **-** Constellation Mapping
- Mbps **-** Megabit per second
- **MBU -** Multiple Bit Upset
- **MCU -** Multiple Cell Upset
- **MEMS -** Micro-Electric Mechanical Systems
- MEO **-** Mid-Earth Orbit
- MeV **-** Mega-electron-Volt
- MIT **-** Massachusetts Institute of Technology
- **NASA -** National Aeronautics and Space Administration
- NICT **-** National Institute of Information and Communications Technology
- **NIEL -** Non-Ionizing Energy Loss
- NFIRE **-** Near Field Infrared Experiment
- **NLC -** Nonlinearity Compensation
- **NODE -** Nanosatellite Optical Downlink Experiment
- **NPTC -** Northeast Proton Therapy Center
- NRZ **-** Non Return to Zero
- **NTIA -** National Telecommunications and Information Administration
- OBPF **-** Optical Bandpass Filter
- **OCSD -** Optical Communications and Sensor Demonstration
- **OGS -** Optical Ground Station
- **OICETS -** Optical Inter-orbit Communications Engineering Test Satellite
- OMERE **-** Outil de Mod6lisation de 'Environnement Radiatif Externe
- OOK **-** On-Off Keying
- **OPALS -** Optical Payload for Laser Communications Science
- OSNR **-** Optical Signal to Noise Ratio
- **OTA -** Optical Transceiver Assembly
- Pbps **-** Petabits Per Second
- PCB **-** Printed Circuit Board
- PDL **-** Polarization Dispersion Loss
- PI **-** Polyimide
- PM **-** Polarization Multiplexed
- PMD **-** Polarization Mode Dispersion
- **PMQ -** Polarization-Multiplexed Quadrature
- POL **-** Point of Load
- PPO **-** Polyphenylene Oxide
- PTFE **-** PolyTetraFluoroEthylene
- **QAM -** Quadrature Amplitude Modulated
- **QPSK -** Quadrature Phase Shift Key
- Rad-Hard **-** Radiation-Hardened
- RF **-** Radio Frequency
- Rx **-** Receive
- **SAA -** South Atlantic Anomaly
- **SD -** Soft Decision
- **SEB -** Single Event Burnout
- SEDR **-** Single Event Dielectric Rupture
- **SEE -** Single Event Effect
- SEFI **-** Single Event Functional Interrupt
- SEGR **-** Single Event Gate Rupture
- **SEHE -** Single Event Hard Error
- **SEL -** Single Event Latch-Up
- **SEP -** Solar Energetic Particle
- **SESB -** Single Event Stuck Bit
- **SET -** Single Event Transient
- **SEU -** Single Event Upset
- SILEX **-** Semiconductor Inter-satellite Link Experiment
- *SOI* **-** Silicon On Insulator
- **SPE -** Solar Particle Event
- SPOT4 **-** Satellite Pour l'Observation de la Terre 4
- SRAM **-** Static Random Access Memory
- SRIM **-** Stopping Range in Matter
- TerraSAR-X **-** Synthetic Aperture Radar for Earth Observation
- TID **-** Total Ionizing Dose
- TNID **-** Total Non-Ionizing Dose
- TRIM **-** Transport of Ions in Matter
- TRIUMF **-** Tri-University Meson Facility
- Tx **-** Transmit
- VOA **-** Variable Optical Attenuator
- WDM **-** Wavelength-Division Multiplexed
- ZB **-** ZettaByte

Chapter 1

Introduction

Global Internet Protocol (IP) traffic is estimated to reach **2.3** zettabytes (ZB) (1021) per year **by** 2020 [Cisco, **2016].** Satellite communications systems have the potential to play a key role in meeting worldwide data capacity demand as they evolve towards providing broadband that can augment or compete with terrestrial broadband. Satellite systems are well-suited for providing connectivity to large areas with low population densities and enable subscribers to connect to broadband and Internet backbone networks rapidly with little setup overhead required for the ground terminal, only a modem and an antenna. The cost to field satellite terminals for providing connectivity may be competitive with costs for point-by-point roll out of terrestrial fiber networks [ITU and **UNESCO, 2015].**

The development and launch of satellites as well as the construction of satellite broadband infrastructure is a still significant investment. Second to satellite launch costs, which can reach tens of millions of dollars, capital and operating expenditures for radio frequency (RF) gateways dominates the cost of connectivity via broadband systems. Due to limited bandwidth $(e.g. K band)$, increasing capacity requires additional, spatially diverse ground gateway sites.

1.1 Free-Space Optical Communications Systems

Free space optical communications **(FSOC)** could serve as a viable and more costeffective solution for future satellite systems. The larger carrier frequency and narrow transmit beamwidth of optical or laser communications are the key advantages over RF [Hemmati, **2008].** Higher frequency provides for greater bandwidth, reduces system mass, volume, and power consumption for transmit and receive apertures based on mission application, and allows for these systems to not be subjected to frequency regulation and spectrum restrictions of RF communications systems [Toyoshima, **2005;** Cornwell, **2016].** The National Telecommunications and Information Administration (NTIA) of the United States Department of Commerce does not require authorization for the use of frequencies above **3000** GHz **[NTIA, 2015].** However, the higher frequency of optical communications yields vulnerability to weather outages, such as cloud coverage, if crosslinks and multiple optical ground stations are not incorporated in the overall system design. **A** narrow beamwidth provides for greater power efficiency and potentially increased security, but at the cost of precision pointing control [Yoon, **2017]. FSOC** systems are currently not as prevalent as RF communications systems on satellites due to challenges with fine beam pointing and cloud coverage [Hemmati, **2008].** As fine beam pointing technology further develops, mitigation strategies for cloud coverage progresses, and development of optical communications systems advances, there is significant potential for an increase in space-based, **FSOC** systems in the near future.

1.1.1 Space-Based **FSOC Systems in Development**

Currently, there are several efforts **by** industry, government entities, and academic institutions to develop **FSOC** systems for satellites. **A** few of the efforts are mentioned in this section, but many others exist.

Through funding with the National Aeronautics and Space Administration **(NASA)** Small Satellite Technology Program, the Aerospace Corporation is developing the Optical Communications and Sensor Demonstration **(OCSD)** for cube satellite (CubeSat) applications. **OSCD-B** on AeroCube-7B and **OSCD-C** on AeroCube-7C will be launched in late **2017** and are designed to downlink up to 200 Megabits per second (Mbps) at a **900** km distance from low Earth orbit **(LEO)** to a ground station. Each CubeSat with **OCSD** as payload is **2.5 kg** [Janson and Welle, **2016].**

The Aerospace Corporation and Draper Laboratory are in collaboration with BridgeSat to develop an optical communications network through the design of laser terminals on **LEO** satellites and optical ground stations [BridgeSat, **2017].** Vialight, founded in **2009,** is a spin-off company from the German Aerospace Center (DLR), working on the development of satellite-to-ground optical communications system [Vialight, **2017].** Laser Light Communications intends to deploy a global optical network, HALO, consisting of up to 12 optical satellites in medium Earth orbit (MEO) with data rates up to 200 Gigabits per seconds (Gbps) for inter-satellite links and up to **100** Gbps for uplinks and downlinks between MEO and ground stations [Laser Light Communications, 2014].

Sinclair Interplanetary is working on optical communications for small spacecraft. Datasheets have been released for a laser crosslink system with data rates up to **100** Mbps at **250** km range and up to **6.25** Mbps at **1000** km range. The transmitter is designed with **1** W optical output power and the system weighs -0.4 **kg. A** downlink laser transmitter system has been designed with data rates up to **1** Gbps for distances up to **1000** km and data rates up to **250** Mbps for distances up to 2000 km. The downlink transmitter system is designed with **1** W output power and has a mass of -0.34 **kg** [Sinclair Interplanetary, **2017].**

The Massachusetts Institute of Technology (MIT) is developing two **FSOC** demonstrations, the Nanosatellite Optical Downlink Experiment **(NODE)** and the Freespace Lasercom and Radiation Experiment (FLARE), for CubeSats. Both **FSOC** systems are designed with commercial off the shelf **(COTS)** components. **NODE** will be hosted on a CubeSat, which is scheduled for launch in **2017,** and is designed for an optical downlink with 0.2 W transmit power and data rates up to **70** Mbps. FLARE is designed to demonstrate optical crosslinks between two **LEO** CubeSats with data rates greater than **10** Mbps [Clements, **2016].**

1.1.2 Demonstrated Space-Based **FSOC** Systems

Within the past two decades, several **FSOC** systems have been successfully demonstrated with optical links between spacecraft and Earth ground stations as well as optical inter-satellite links.

The National Institute of Information and Communications Technology (NICT) performed the first bi-directional optical communications demonstration in 1994 between the Laser Communication Experiment **(LCE)** onboard the Engineering Test Satellite **(ETS-VI)** and both NICT optical ground station in Tokyo, Japan and Jet Propulsion Laboratory **(JPL)** optical ground station in Table Mountain, California. During the demonstration, **ETS-VI** was in a **highly** elliptical orbit of **38,700** km apogee, **8.560** km perigee, and 13-degree inclination. **LCE** weighed 22.4 **kg** and transmitted data with Manchester coded pulse modulation at **-1** Mbps. **13.8** mW average optical output power was transmitted over a **37,800** km link distance. [Araki et al., **1997].**

In 2001, the European Space Agency **(ESA)** demonstrated Semiconductor Intersatellite Link Experiment (SILEX), the first optical inter-satellite link from a **LEO** French spacecraft, Satellite Pour l'Observation de la Terre 4 (SPOT4), to a geostationary Earth orbit **(GEO)** spacecraft, Advanced Relay and TEchnology MIssion Satellite (ARTEMIS). The optical link was a distance of nearly 40,000 km with data rate of **50** Mbps using On-Off Keying (OOK) modulation with Non-Return to Zero (NRZ) coding [Tolker-Nielsen and Oppenhaeuser, 2002]. The SILEX terminals were about **100 kg** [Fletcher et al., **1991].** The Japanese Aerospace Exploration Agency **(JAXA)** designed an optical communications terminal, Laser Utilizing Communications Equipment **(LUCE),** onboard the **JAXA** spacecraft Optical Interorbit Communications Engineering Test Satellite **(OICETS).** The **LUCE** system weighed 140 **kg** and transmitted an average of **100** mW. In **2005, LUCE** and **SILEX,** onboard ARTEMIS, demonstrated the first bi-directional optical inter-orbit communications experiment between **LEO** and **GEO.** The optical link was nearly 40,000 km with the **LEO** to **GEO** uplink transmitting at data rate of **50** Mbps and the **GEO** to **LEO** downlink transmitting at data rate of 2 Mbps [Fujiwara et al., **2007;** Toyoshima et al., 2004].

The DLR developed laser communications terminals (LCTs) based on homodyne binary phase shift keying (BPSK). One **LCT** was flown on the **US** satellite, Near Field Infrared Experiment (NFIRE) and a second **LCT** was flown on German satellite, Synthetic Aperture Radar for Earth Observation (TerraSAR-X). In **2008,** the two LCTs demonstrated a **LEO** to **LEO** optical inter-satellite link over five thousand km distance at **5.625** Gbps [Gregory et al., 2011]. The **LCT** on TerraSAR-X also demonstrated an optical link at **5.625** Gbps from **LEO** to European Space Agency **(ESA)** Optical Ground Station **(OGS)** in Tenerife, Spain [Ciminelli et al., **2016].** The LCTs have a mass of **32 kg** and a peak transmit power of **0.7** W [Fields et al., **2009].**

MIT Lincoln Laboratory designed the Lunar Laser Communications Demonstration **(LLCD) FSOC** system, Lunar Laser Space Terminal **(LLST)** with mass of **30 kg,** onboard the **NASA** mission, Lunar Atmosphere and Dust Environment Explorer **(LADEE).** In **2013,** the **LLST** transmitted an optical downlink from lunar orbit to their Earth ground terminal with data rate of **622** Mbps and **0.5** W optical output power over 400,000 km maximum link distance [Robinson etal., 2011; Boroson etal., 2014].

In 2014, **NASA JPL** demonstrated the Optical Payload for Laser Communications Science **(OPALS)** system with an optical downlink from the International Space Station **(ISS)** to a ground station at Table Mountain, California. The **OPALS** system transmitted with a data rate of **50** Megabits per second (Mbps) using OOK modulation and Reed Solomon forward error correction **(FEC).** The system weighed **50 kg** and transmitted with an average optical output power greater than **0.83** W [Oaida etal., 2014].

Table **1.1** summarizes **FSOC** systems that have been successfully demonstrated and describes some of the key parameters of each system.

FSOC System	Mass	Maximum Link Distance	Optical Link Source	Optical Link End	Data Rate	Detection	Tx Power
OPAL_s (JPL 2014)	50 _{kg}	700 km	ISS	Optical Ground Terminal (TPL)	50 Mbps	Direct	> 0.83 W
LLCD (MIT LL 2013)	30 kg	400000 km	Lunar Orbit (LADEE s/c)	Optical Ground Terminal	622 Mbps	Direct	0.5W
LCT (DLR 2008)	32 kg	5100 km	LEO (NFIRE s/c)	LEO $(TerraSAR-X s/c)$	5.625 Gbps	Coherent	0.7W
LCT (DLR 2008)	32 _{kg}	5100 km	LEO (NFIRE s/c)	Optical Ground Terminal (ESA)	5.625 Gbps	Coherent	0.7W
LUCE (JAXA 2005)	140 kg	38000 km	LEO (OICETS s/c)	GEO (ARTEMIS s/c)	50 Mbps	Direct	100 mW
SILEX (ESA 2005)	100 kg	38000 km	GEO (ARTEMIS s/c)	LEO (OICETS s/c)	2.048 Mbps	Direct	37 mW
SILEX (ESA 2001)	100 kg	38000 km	LEO (SPOT4 s/c)	GEO (ARTEMIS s/c)	50 Mbps	Direct	60 mW
LCE (NICT 1994)	22.4 kg	38700 km	Elliptical Orbit (ETS-VI)	Optical Ground Terminal (NICT, JPL)	1.024 Mbps	Direct	13.8 mW

Table **1.1** Demonstrated **FSOC** systems and key parameters

A majority of the described **FSOC** systems were designed with direct detection optical communications systems and reached data rates on the order of 100s of Mbps. Figure **1.1** plots the demonstrated **FSOC** missions based on the maximum optical link distance and maximum achieved data rate. **LCT** designed **by** DLR achieved the highest data rate of **5.625** Gbps with a coherent optical communications system.

Figure **1.1** Demonstrated **FSOC** missions based on maximum optical link distance, data rate, and transmit power

In comparison to a direct detection system, a coherent system uses both the phase and intensity of a light wave to transmit information. In a direct detection system, the received optical signal is focused directly onto a photodetector, which generates a signal current proportional to the number of photons received. In a coherent system receiver, the received optical signal is superimposed with a light wave generated **by** a local oscillator (LO) laser and both optical signals are focused onto a photodetector. The signal current from the photodetector is non-linear and is dependent on the amplitude, phase, and polarization of the received optical signal and LO light wave [Pribil and Hemmati, **20081.** Coherent optical communications have the potential to meet high bandwidth and data capacity demands through multi-level modulation formats $[Xi$ ang et al., 2014.

1.2 Coherent Optical Communications Systems

Coherent optical communications systems will serve a key role in terrestrial optical network capacity expansion [Xiang et al., 20141. Coherent systems have potential to achieve highest receiver sensitivity, high spectral efficiency, tolerance against

dispersion effects, and superior performance over long transmission distances [Pfau et al., **2008].** The performance of optical coherent detection systems could enable future systems with transmission at 200 Gbps, 400 Gbps, and up to **1000** Gbps per wavelength [Roberts et al., **2009].**

Coherent optical communications were studied in the 1980s, however, interest shifted toward wavelength-division multiplexed (WDM) direct detection using erbium-doped fiber amplifiers (EDFAs). In the past few years, there has been revived interest in coherent systems since recent technological advancements in optical coherent receivers and high speed digital circuits now allow for realization of coherent optical communications [Kikuchi, 2010]. Coherent optical receivers require high-speed analog-to-digital converters (ADCs) with sampling rates which can reach the symbol rate of high data rate signals. Fast circuit ADCs have been developed with sampling rates greater than **10** Giga-samples per second to meet the needs of high data rate coherent optical communications systems [Pfau et al., **2008].**

Digital signal processing **(DSP)** through application specific integrated circuits (ASICs) is the key technological advancements to the realization of coherent optical communications systems. Prior to the development of optical coherent **DSP** ASICs in **2007,** coherent systems would store transmitted data in a computer and analyze bit errors offline. The combination of coherent detection and **DSP** through an **ASIC** provides for real-time operation of digital coherent receivers [Savory, **2008;** Kikuchi, 2010]. Advances to Complementary Metal Oxide Semiconductor **(CMOS) ASIC** technology affords the ADCs and **DSP** necessary to access and manipulate optical electric field signal for optical coherent systems [Roberts et al., **2009].**

Optical coherent **DSP** ASICs have streamlined complex optical and analog functions to simplify transceiver implementation. **DSP** ASICs have integrated capabilities of analog-to-digital conversion, **FEC** encoding, constellation mapping **(MAP),** fiber nonlinearity compensation **(NLC),** electronic dispersion pre-compensation **(EDC),** and link equalization [Rasmussen et al., **2013].** Fast demultiplexers built into ASICs provide high speed interfaces with the input signals from ADCs [Pfau et al., **2008]. DSP** ASICs optimize optical coherent system performance **by** providing robustness to additive noise and tolerance to linear transmission impairments, such as chromatic dispersion **(CD)** and polarization mode dispersion (PMD) [Rasmussen et al., **2013].**

Optical coherent **DSP** ASICs can also reduce the size, cost, and power consumption of optical communications systems. Optical coherent ASICs with advanced **DSP** techniques can improve transmission capacity and performance for a given set of optoelectronic components, and **DSP** can be used to compensate for high dispersion, error, or noise [Xiang et al. 20141. Implementation of a **DSP ASIC** with integrated low-cost optoelectronic components can yield an overall lower-cost optical communications system. However, customized optical coherent **DSP** ASICs can have high development costs in the range of tens of millions of dollars with **CMOS** technologies [Rasmussen et al., **2013].** Significant investments of time and money would be required to fabricate space-qualified versions of commercial ASICs.

ASICs created with state-of-the-art commercial processes have been at least two generations ahead of space-qualified ASICs created with radiation-hardened (radhard) processes. Alternatives to creating space-grade ASICs with cost and time intensive rad-hard processes include **(1)** applying **CMOS** hardness-by-design approaches (e.g. guard banding around **MOS** transistors or using redundant latches) to custom ASICs created in commercial foundries and (2) testing or qualifying commercially-made ASICs against space mission radiation requirements. The latter option of using commercial ASICs has been recently approached based on increasing demand for higher system performance with the continuously evolving commercial processes [Lacoe, 2000]. With the small satellite revolution, and even for larger satellites, there has been a shift over the past decade toward evaluating, qualifying, and using commercial components for space applications [Sinclair and Dyer, **2013].**

In order for a commercial ASIC to be used for a space mission, space systems engineers typically require the device to undergo extensive qualification tests for survivability and reliability in the space environment, including exposure to ionizing radiation. Spacecraft mission parameters, such as orbit and duration, as well as spacecraft materials and geometry play a role in the radiation environment and total ionizing dose (TID) that spacecraft components will encounter. Non-rad-hard, commercial components must be tested for survival at the TID requirements of the mission as well as tested for exposure to high energy protons and heavy ion particles. Radiation damage from high-energy protons and energetic proton events in the space environment can have detrimental effects on microelectronics, such as increased noise in photonics and single event effects (SEEs) [Petersen, 2011; Hiemtra & Blackmore, **2003].** The purpose of this thesis is not focused on describing in detail the effects of radiation on materials and devices in spacecraft applications. There **is** literature that can provide further detail on this topic [Hastings and Garrett, 2004; Petersen, 2011; Stark, 2011].

The lowest lead time and cost option for applying current optical coherent **DSP** ASICs to a space-based system would be to test them to specific space-mission radiation requirements. Optical coherent **DSP** ASICs are used in commercial optical transceiver packages. In **2007,** an Agile Engine **DSP** ASIC was implemented in an optical coherent dual polarization (DP) quadrature phase shift key **(QPDK)** receiver. The coherent system demonstrated real-time, continuous measurements of 40 Gbps optical signal transmission [Sun et al., **2008;** Roberts et al., **2009].** This laboratory demonstration was a key milestone in the development of coherent optical communications systems using **DSP** ASICs, and the Agile Engine **DSP ASIC** was one of the first known optical coherent **DSP** ASICs commercially designed and successfully demonstrated [Kikuchi, 2010]. Acacia Communications, Inc. (Acacia) developed the first industry product line of **100** Gbps coherent optical transceivers with **DSP** ASICs for analog-to-digital conversion and soft-decision **(SD) FEC.** With long-haul system experiments over **1000** km, Acacia demonstrated the robustness of

their 120 Gbps coherent, polarization-multiplexed (PM) **QPSK** optical transceiver package [Nelson, 2012].

The Inphi **CFP2-ACO** coherent optical transceiver module was designed with a polarization-multiplexed quadrature **(PMQ)** modulator transmitter, an integrated coherent receiver (ICR), and a **DSP ASIC (CL2001OA1).** The Inphi **CFP2-ACO** coherent optical module with ASIC **CL2001OA1** was tested over long-haul distances of **7000** km and **1600** km to demonstrate transmission with **100** Gbps and 200 Gbps rates [Reis, **2016].** This work will focus on the Inphi **CL2001OA1** since this commercial optical coherent **DSP** ASIC has the highest achievable data rate out of all current commercially available coherent optical communications **DSP** ASICs.

1.3 Thesis Structure

Studies on space radiation testing and qualification of commercial optical coherent **DSP** ASICs, such as those mentioned in Section 1.2, have not yet been published and current work on radiation testing of these types of ASICs is unknown. This work investigates if a commercially-available, optical coherent **DSP** ASIC, specifically the Inphi **CL2001OA1,** can meet **LEO** space mission radiation requirements.

In Chapter 2, the space radiation environment for sample **LEO** missions are modeled and analyzed using radiation environment modeling software, Outil de Modélisation de l'Environnement Radiatif Externe (OMERE) (English Translation: Modeling Tool for Space Radiation Environment) version **5.0,** created **by** TRAD Tests and Radiation in Toulouse, France. Models of the sample **LEO** missions are used to generate dose depth curves and determine TID requirements for testing the **CL2001OA1** ASIC. The three sample **LEO** missions included **ISS** orbit, **800** km polar orbit, and **1000** km **0*** inclination orbit, for 1-year, 5-year, and 10-year mission durations. Chapter 2 also discusses background of the **LEO** radiation environment and describes the radiation environment modeling for each of the sample **LEO** missions.

The Inphi **CL2001OA1 ASIC** and the **CL2001OA1 ASIC** evaluation board system (CL20010A1-EVK) for experimental testing are described in Chapter **3.** Potential radiation damage to the **CL2001OA1 ASIC** in a **LEO** space environment is also discussed.

Chapter 4 discusses simulations with the nuclear modeling program, Stopping Range In Matter (SRIM). SRIM analyses are used to simulate protons penetrating the CL2001OA1-EVK and to evaluate the minimum proton energy level to reach the **CL2001OA1** silicon active region.

Two proton radiation test campaigns are completed to assess the TID effects and SEEs from ionizing proton radiation on the **CL20010A1. A** description of the test campaigns is in Chapter **5.** Proton **SEE** cross section results from the test campaigns are used to calculate the **SEE** rate of the **CL2001OA1** for the three sample **LEO** missions. The **SEE** rate calculation is detailed in Chapter **6.**

Radiation mitigation strategies for the **CL2001OA1** are determined for potential application to a **LEO** space mission in Chapter **7.** Finally, the **CL2001OA1** is assessed for a **LEO** space mission, based on the initial radiation requirements from the sample **LEO** missions modeled and the results of the investigation. Figure 1.2 outlines the investigative approach, which is used to evaluate the performance of the **CL2001OA1** ASIC for a **LEO** space radiation environment.

Figure 1.2 Approach used in this work to evaluate the **CL2001OA1** ASIC for **LEO** space radiation environment

Overall, the results of this work can be used to evaluate if the commercial Inphi **CL2001OA1** optical coherent **DSP** ASIC can be feasibly implemented for a **LEO** space application. This work is a key step toward the widespread implementation of high data rate, coherent optical communications systems on satellites.

Chapter 2

Low Earth Orbit Space Radiation Environment

This chapter introduces the **LEO** radiation environment, discusses the radiation sources and effects that are relevant to the focus of this thesis to **DSP** ASICs, and models the radiation environment of **LEO** missions to create a set of TID requirements. Figure 2.1 shows the high level approach used in this work, based on the detailed approach in Figure 1.2, and highlights the relevant contributions from this Chapter toward assessing the **CL2001OA1** ASIC for application in the **LEO** radiation environment.

Figure 2.1 High-level approach used in this work to evaluate the **CL2001OAl** ASIC for **LEO** space radiation environment with Chapter 2 accomplishments highlighted in yellow.

2.1 Low Earth Orbit Radiation Environment

The radiation environment for **LEO** space missions consists primarily of trapped particles in the inner Van Allen Radiation Belt, galactic cosmic rays (GCRs), and solar energetic particles (SEPs) that penetrate into the Earth's magnetosphere [Badhwar, **1997].**

2.1.1 Trapped Particles in the Inner Radiation Belt

The inner belt has a typical altitude range of **1000** km to **6000** km [Ganushkina et al., 2010]. Within the inner belt, charged particles from GCRs and solar winds are trapped in the Earth's magnetic field and follow the magnetic field lines. Energetic trapped protons dominate the inner belt and reach energy levels from **10** kiloelectron-volts (keV) to **300** Mega-electron-volts (MeV) [Suparta, 2014]. Trapped electrons reach energy levels from **10** keV to **5** MeV [Stassinopoulous, **1988]** in the inner belt. For **LEO,** trapped particles dominate at low inclination orbits with high altitude. Trapped proton fluxes in the inner belt are anti-correlated to solar activity [Suparta, 2014, Varotsou, **2017].**

Models of **1** MeV trapped protons and **1** MeV trapped electrons within the Van Allen radiation belts are shown in Figure 2.1 and Figure 2.2, respectively. The greatest concentration of high energy trapped particles are in the region between the inner and outer radiation belts in Figure 2.2 and Figure **2.3.** The OMERE (version **5.0)** flux mapping module is used to model the trapped particle fluxes for a 1-year mission with ISS orbit starting on January **1,** 2020. **A** trapped proton model, AP8-MIN, and trapped electron model, **AE8-MAX** are used with the IGRF magnetic field model. These trapped particle models are described in further detail in Section **2.3.**

Figure 2.2 Integral flux mapping for **1** MeV trapped protons. The OMERE flux mapping module is used to model trapped protons with AP8-MIN and IGRF for Earth's Magnetic Field. Red represents trapped protons with highest fluxes between \sim 1 x 10⁶ and \sim 1 x 10⁷ particles/cm²/s, yellow represents fluxes between \sim 5 x $10⁴$ and \sim 1 x 10⁶ particles/cm²/s, and green represents fluxes between \sim 1 x 10² and \sim 5 x 10⁴ particles/cm²/s.

Figure **2.3** Integral flux mapping for **1** MeV trapped electrons. The OMERE flux mapping module is used to model trapped protons with **AE8-MAX** and IGRF for

Earth's magnetic field. Red represents trapped protons with the highest fluxes between \sim 1 x 10⁶ and \sim 1 x 10⁷ particles/cm²/s, yellow-orange represents fluxes between \sim 5 x 10⁴ and \sim 1 x 10⁶ particles/cm²/s, and green represents fluxes between \sim 1 x 10² and \sim 5 x 10⁴ particles/cm²/s.

The South Atlantic Anomaly **(SAA)** is a region with increased energetic particle fluxes [Stark, 2011]. **Highly** energetic protons in the **SAA** can affect spacecraft at orbital altitudes of up to **-1000** km and orbital inclinations of **35-60** degrees [Lohmeyer, **2015].** The **SAA** is caused **by** an offset between the Earth's magnetic field axis and Earth's rotational axis by angle of \sim 11 degrees and by an additional Northward offset of **- 500** km [Varotsou, **2017; ESA, 2008;** Stark, 2011]. The **SAA** is generally located in the region to the southeast of Brazil, but the geomagnetic field is dynamic and influences the geographic location of the **SAA** [Heynderickx, 2002].

The OMERE flux mapping module is used in Figure 2.4 to show the **SAA** for the trapped proton environment of the 1-year mission with ISS orbit starting on January 1, 2020. Figure 2.4 shows the **SAA** at 400 km altitude with a concentration of high energy trapped protons near the region southeast of Brazil.

Figure 2.4 OMERE model of trapped protons. Concentration of high energy protons near the region southeast of Brazil highlights the South Atlantic Anomaly. Red represents trapped protons with the highest fluxes between \sim 1 x 10⁷ and \sim 1 x 10⁸ particles/cm²/s, yellow-orange represents fluxes between $\sim 5 \times 10^4$ and $\sim 1 \times 10^7$ particles/cm²/s, and green represents fluxes between \sim 1 x 10² and \sim 5 x 10⁴ particles/cm 2/s.

2.1.2 Galactic Cosmic Rays

GCRs originate from outside the solar system and are formed from diffusive shock acceleration of supernova remnants. GCRs detected within our solar system reach energy levels of ~ 1 GeV [Suparta, 2014]. A spacecraft in ISS orbit can encounter proton energy range greater than 1 GeV and up to $\sim 10 \text{ GeV}$ with fluences of $\sim 10^{-2}$ particles/m² sec [Hastings and Garrett, 1996]. Ions from GCRs can reach energy levels up to **-300** MeV per nucleon and tend to have lower fluxes but higher energy levels [Varotsou, **2017].** Primary GCRs consist of protons, alpha particles, and heavy nuclei. Secondary GCRs consist of neutrons, pions, positrons, and muons, and are a result of primary GCR decay **by** interaction with Earth's atmosphere. GCRs are anticorrelated with the solar cycle, reaching a maximum during solar minimum, and can be deterred **by** solar wind, which is strongest during solar maximum. GCR particles move perpendicular to the Earth's magnetic field lines and can be deflected at the equator, and funneled toward the poles. Thus, for **LEO** spacecraft, GCRs are most relevant at high altitude and high inclination, polar orbits [Suparta, 2014].

2.1.3 Solar Energetic Particles

SEPs consist of electrons, protons, and heavy ions from solar flares and coronal mass ejections (CMEs). The Earth's magnetosphere helps to shield spacecraft from solar particles, especially ions and electrons, but high-energy protons may not be deterred **by** the magnetosphere. This yields a greater amount of solar protons in comparison to electrons and heavy ions, and these solar protons dominate among the SEPs due to the high energy level. Solar particle events (SPEs) are solar flares and CMEs,

which yield protons with energies greater than **30** MeV with flux up to **107** cm-2 and protons with energies greater than **10** MeV with higher fluxes of up to **1010 cm-2** [Suparta, 2014]. Solar energetic protons have energy levels that can range from a few keV up to **500** MeV, while solar energetic ions have energy levels that can range from **¹**to **100** MeV per nucleon [Varotsou, **2017].** SPEs are correlated with the solar cycle and tend to occur during solar maximum. Solar particles are a significant source of radiation for high altitude and **high** inclination Earth orbits, since high energy protons tend to penetrate the magnetosphere in these regions [Suparta, **2009].**

2.1.4 Summary of Low Earth Orbit Radiation Environment

LEO spacecraft are affected **by** trapped protons and trapped electrons in the inner belt, protons and ions from SEPs, and protons and ions from GCRs. Table 2.1 below summarizes the types of particles from each **LEO** radiation environment source, the particle energy levels in **LEO,** and the relevant areas in **LEO** that are affected **by** each type of particle.

Table 2.1 Summary of **LEO** radiation sources, particles from radiation sources, particle energy level in **LEO,** and affects areas in **LEO**

Radiation		Particle Energy	Affected Areas in
Source	Particle Type	Level in LEO	LEO
Inner Belt	Protons	$10 \text{ keV} - 300$	SAA and high
		MeV	inclinations
	Electrons	$10 \text{ keV} - 5 \text{ MeV}$	SAA and high
			inclinations
GCR	Protons	~ 1 GeV	High inclinations
	Ions	$<$ \sim 300 MeV/n	High inclinations
SEP	Protons	$keV - 500 MeV$	High altitudes and
			low inclinations
	Ions	$1-100$ MeV/n	High altitude and
			low inclinations

Particles from the space radiation environment generate TID effects, total nonionizing dose (TNID) or displacement damage **(DD)** effects, and SEEs. Figure **2.5** outlines the main **LEO** radiation environment sources, the types of particles from each source, and the radiation effects caused **by** the particles. The radiation effects caused **by** the particles of each **LEO** radiation source are further described in Section 2.2.

Figure **2.5 LEO** radiation environment sources, types of particles from each source, and radiation effects caused **by** each type of particle

2.2 Low Earth Orbit Radiation Environment Effects on Spacecraft Components

The three main radiation effects in the **LEO** space environment are TID effects, TNID or **DD** effects, and SEEs. TID and TNID are cumulative radiation effects based on long-term exposure to a large number of charged particles penetrating into an electrical device. Thus, TID and TNID are related to mission duration and result in progressive degradation. SEEs are non-cumulative and are the result of a single charge particle penetrating into device material and causing functional disruption or failure in the electrical device [McMorrow et al., 2004].

2.2.1 Total Ionizing Dose Effects

Trapped protons and electrons from the radiation belts, protons from SPEs, and protons from GCRs are sources of TID effects on spacecraft electronics [Varotsou, **2017].** TID effects occur when charged particles deposit ionizing energy into target material and the ionization alters the material **by** generating electron-hole pairs and inelastic Coulombic scattering [Alig and Bloom, **1975;** Miroshnichenko, **2003].**

TID level depends on the type and energy of the incident particle(s) and the density and composition of the target material. TID is generally calculated as the amount of total energy deposited **by** all incident particles in the target material divided **by** the total mass of the incident particles. TID is expressed in units of Joules per kilogram, Gray, or rad **(1** Gray **= 100** rad). Equation 2.1 outlines the general calculation for TID.

$$
TID = \frac{Energy\ Deposited}{Mass} \left[\frac{J}{kg} = Gray = 100\ rad\right]
$$
 (Equation 2.1)

TID can be described with the ionizing stopping power or linear energy transfer (LET) of a single particle. As a proton or electron penetrates an electronic device, energy **is** transferred from the particle to the device material and the particle slows down. Electrons within the device material act as a viscous medium to slow down particles and cause ionizing energy loss. LET is a function of the ionizing energy loss **by** path length of a particle and normalized **by** the target density [McMorrow et al., 2004]. LET is expressed in units of MeV cm² per μ m. Equation 2.2 displays the expression for LET.

$$
LET = \left(\frac{1}{\rho}\right) \cdot \left(\frac{dE}{dx}\right)_{ionizing} \left[\frac{MeV \cdot cm^2}{\mu m}\right] \text{(Equation 2.2)}
$$

The LET and the fluence of incident particles are multiplied to calculate TID. Equation **2.3** shows the relationship between TID and LET with fluence of incident particles.

$TID = \Phi \cdot \text{LET}$ (Equation 2.3)

TID effects in electronic devices result from charge trapping in insulators, such as gate oxides, and semiconductor/insulator interfaces. TID effects are observed as electrical device parameters drift [Tuite, **2013].** Trapped particles in the inner belt can cause interference with sensors and degradation of spacecraft electronic parts from accumulated TID [Stark, 2011; Suparta, **2009].**

Technologies sensitive to TID effects include **MOS** (Metal Oxide Semiconductor) transistors, field effect transistors (FETs), linear integrated circuits (ICs), and bipolar junction transistors (BJTs). TID effects can cause a shift in the threshold voltage of **NMOS** devices and a decrease in the drive current of PMOS devices. For **CMOS** devices, TID effects can result in decreased switching speed and increased leakage current. The effects of ionizing radiation on **CMOS** devices are further detailed in Chapter **3.2.** In junction gate FETs or JFETs, TID effects include enhanced sourcedrain leakage currents. TID effects on BJTs include degradation to the gain current for low current conditions [Poizat, **2009].**

Common spacecraft electronic devices with **MOS, FET,** IC, and **BJT** technologies include charged-coupled devices (CCDs), optical components, micro-electromechanical systems **(MEMS),** memory chips, bipolar electronic devices, and general microelectronics. The most sensitive devices to TID failures are bipolar electronic devices. CCDs and optical components can experience increased dark current. Charge build-up in the dielectric layers of **MEMS** from TID can lead to a shift in response. Memory devices, such as static random access memory (SRAM) chips are also sensitive to TID failures due to **CMOS, BJT,** and **FET** technologies within the chips [Sukhaseum, **2017].** Potential TID damage on digital microelectronics includes enhanced transistor leakage and logic failures due to reduced gain in BJTs or shifted threshold voltages and reduced switching speeds in **CMOS.** TID effects can cause gain degradation and change the offset voltage, offset current, and bias-current of analog microelectronics [Poizat, **2009].**

TID sensitivity of an electronic device is dependent on the technology passivation, device structure, and when and where the device was manufactured. Different foundries manufacture electrical devices with different doping levels that produce cause dose sensitivity levels. Although manufacturing quality of semi-conductor regions is controlled based on doping levels, there are differences in the manufacturing quality of the oxide layer among different foundries or even among different manufacturing lots of the same foundry [Dodd et al., 2010].

2.2.2 Total Non-Ionizing Dose or Displacement Damage Effects

Trapped protons and electrons from the radiation belts, protons from SPEs, and protons from GCRs cause **DD** or TNID effects on spacecraft electronics [Varotsou, **2017]. TNID** effects are caused **by** incident particles entering the target material and interacting with nuclei. When an incident particle hits a nucleus, the nucleus can become displaced in the crystal lattice, causing elastic scattering and nuclear reactions. Displacement damage results from the cumulative physical degradation of the lattice material [Miroshnichenko, **2003].**

TNID can be described **by** the nuclear stopping power or non-ionizing energy loss (NIEL) of a particle penetrating the target material and transferring energy to the nucleus. **NIEL** is a function of the nuclear energy loss **by** path length unit of a particle and normalized by the target density. NIEL is expressed in units of MeV cm² per μ m. Equation 2.4 shows the expression for NIEL.

$$
NIEL = \left(\frac{1}{\rho}\right) \cdot \left(\frac{dE}{dx}\right)_{nuclear} \left[\frac{MeV \cdot cm^2}{\mu m}\right] \text{(Equation 2.4)}
$$

The **NIEL** and the fluence of incident particles are multiplied to calculate **TNID.** Equation **2.5** shows the relationship between TNID and **NIEL** with fluence of incident particles. TNID is expressed in units of Joules per kilogram, Gray, or rad.

 $TNID = \Phi \cdot NIEL$ (Equation 2.5)

The TNID level or the amount of **DD** at a sensitive region of the device is calculated for a specific mission based on the radiation environment. TNID effects cause stable defects, which behave as recombination centers, trap centers, and diffusion centers. Recombination centers impact the carrier lifetime, trap centers impact resistivity of electronic circuits, and diffusion centers impact mobility of electrons and holes. The following types of devices are sensitive to displacement damage effects and are listed in order of most sensitive: opto-electronics, bipolar ICs, **MOS** transistors, power MOSFETs, and **CMOS.** The most sensitive devices are minority carrier devices [Sukhaseum, **2017].**

2.2.3 Single Event Effects

Protons and heavy ions from both GCRs and solar flares as well as trapped protons in the inner radiation belt cause SEEs in electronic devices on **LEO** spacecraft [Varotsou, **2017].** The greatest potential hazard from GCRs for **LEO** spacecraft are SEEs, caused primarily **by** abundant deposition of iron nuclei [Suparta, **2009].** Energetic heavy ions from GCRs can deposit significant amounts of energy in spacecraft electronics, and large amount of kinetic energy from GCRs can have permanent, damaging effects on materials through which the particles may pass **[ESA, 2008;** Stark, 2011]. Heavy ions from GCRs and energetic protons can trigger latch up **by** producing charge in the well-substrate junction of electronic devices. In some **LEO** environments, trapped protons can have greater energy levels (such as in the **SAA)** than GCR heavy ions and may induce more SEEs [Petersen, **1996].**

SEEs occur when a charged particle is deposited or passed through active components with electrical circuits, such as memory, power, and logic devices. Specifically, **SEE** radiation damage is caused when a charged, high-energy particle, commonly a proton, impacts the energy structure and lattice structure of a semiconductor material in an electronic device. The charged particle transmits energy to the semi-conductor material and can cause a displacement of a lattice atom in the material. Local ionization occurs within the material from the generated charge along the ion path and an impulsive release of charge from the ionization process causes a **SEE** [Stark, 2011]. Within microelectronic devices, SEEs result from energetic protons interacting with silicon nuclei and producing ionizing nuclear recoils [Hiemstra, **2003].** SEEs cause a disruption in electronic device operation and can have either destructive or non-destructive consequences to devices.

2.2.3.1 Types of Single Event Effects

SEEs can be destructive or non-destructive to spacecraft components. Figure **2.6** outlines the types of SEEs based on destructive and non-destructive effects.

Figure **2.6** Non-destructive and destructive **SEE**

SEEs are particular to certain electronics and components, based on the structure or technology. Table 2.2 lists the types of SEEs as well as some devices and technologies that are susceptible to each type of **SEE.** The types of SEEs relevant to the

CL2001OA1 ASIC, which is a **CMOS** device, are highlighted and indicated in Table 2.2 and include single event upsets (SEUs), single event functional interrupts (SEFIs), multiple bit upsets (MBUs), multiple cell upsets (MCUs), single event latchups (SELs), and single event hard errors (SEHEs) [Mutuel, **2016].** The non-relevant types of SEEs to the **CL2001OA1,** such as single event transients (SETs), single event gate ruptures (SEGRs), single event burn-outs (SEBs), and single event dielectric rupture (SEDR), are further discussed in Section **A. 1** of the appendix.

Non-destructive SEEs relevant to the **CL2001OA1** include SEUs, SEFIs, MBUs, **MCUs,** [Samaras, 2014; Mutuel, **2016].** SEUs are one of the most common SEEs for spacecraft electronics. SEUs occur in digital, analog, and optical components, and device areas sensitive to SEUs include memories, buffers, and latches. SEUs change the logic state of a circuit in an electronic device and induce soft errors [Baker, 2002]. Within **MOS** devices, **Off NMOS** and PMOS drain transistors are sensitive to SEUs [McMorrow, 2004]. **A** reset or rewrite of the device logic after an **SEU** occurrence can typically restore normal operation of the device.

SEFIs are characterized **by** a loss of normal device operation and are identified **by** device functional "hangs" in which the device falls into an unknown or unmanaged state. SEFIs can also result in permanent damage to a device. To counteract a SEFI, devices must be reset or power cycled to recover normal operation. SEFIs occur in complex devices with built-in state or control sections such as field programmable gate arrays (FPGAs), ASICs, dynamic random-access memory (DRAM), ADCs, and digital-to-analog converters (DACs) [Normand, 2004; Samaras, 2014].

Both MBUs and MCUs occur when a single particle triggers several bit flips, typically within memory devices. **A MBU** induces the corruption of several bits in the same memory word or address [Normand, 2004]. **A MCU** triggers several upsets or transients as it penetrates through a device or system.

Destructive SEEs relevant to the **CL20010A1** include SEHEs and SELs [Samaras, 2014; Mutuel, **2016].** SEHEs result from a **SEU** causing permanent change to the state of a memory element or buffer. SELs have an induced high current state, which can result in the loss of device functionality or potentially permanent damage. SELs are short-circuits between the ground and supply voltage and occur in **CMOS** and BiCMOS technologies. **A** power reset can be used to return to normal operation after occurrence of an **SEL,** but this type of **SEE** can also be destructive [Samaras, 2014].

2.2.3.2 Single Event Effect Rate

SEE device sensitivity is characterized **by** the probability of an **SEE** occurrence or the device cross section. The device cross section is determined based on specific conditions, such as the type of device, particle, and **SEE.** The **SEE** cross section of a device at a given particle energy level is the ratio of the total number of events from SEEs to the particle fluence, expressed in units **of cm ² .** Equation **2.6** shows the calculation for **SEE** cross section of a device.

$$
\sigma = \frac{\text{Total Number of Events}}{\text{Particle Fluence}} \, [\text{cm}^2] \, (\text{Equation 2.6})
$$

For ions, the device cross section is defined **by** the LET of heavy ions. There is a critical LET level or the LET threshold, over which a particle will induce a **SEE** [Hastings and Garrett, **1996].** For protons, the device cross section is defined **by** the energy level transported. Data points for the device cross section at specific LET or energy levels are collected from radiation testing with protons or heavy ions. These data points are used to generate a Weibull distribution and a cross section curve fitted to the test data. The Weibull distribution is the typical distribution used to parametrize **SEE** cross section data since it levels to a limiting value [Petersen et al., **1992].** The Weibull distribution provides better empirical description of cross section data in comparison to other distributions, such as Bendel 2-parameter fits [Tylka et al., **1996].** The cross section curve is then used to calculate the **SEE** rate for a device

[Samaras, 2014]. This process is used to calculated the proton **SEE** rate for the **CL2001OA1** ASIC in Chapter **6. SEE** data for the calculations are collected from proton radiation test campaigns described in Chapter **5.**

The software program, OMERE, calculates the ion **SEE** rate for one sensitive volume within a device at all LET values and integrates over the ion **SEE** cross section curve. OMERE uses the Cosmic Ray Effect on Electronics (CREME) model **by** Adams **[1986]** to generate a CREME rate. The CREME model assumes a rectangular parallelepiped sensitive volume, constant LET along the ion path, and a critical charge of the sensitive volume. The CREME rate at each LET is generated **by** modeling charged ions through a volume and defines that an event occurs if the charge of a penetrating ion is greater than the critical charge of the sensitive volume. The product of the CREME rate at each LET and the LET threshold distribution is integrated over the device cross section curve to generate the device **SEE** rate [Samaras, 2014]. Equation **2.7** shows the calculation for the device **SEE** rate with the parameters as the CREME rate, T, and the LET threshold distribution, ζ , as functions of LET, L.

$$
\tau = \int_{L0}^{\infty} T(L) \cdot \zeta(L) \, dL \, \left[\frac{Number \, SEES}{Unit \, Time} \right] \, (\text{Equation 2.7})
$$

The process for calculating the ion **SEE** rate will be completed in future work with heavy ion test campaigns, as described in Section **8.2.2.**

The proton cross section curve from the Weibull distribution fit is expressed as a function of the incident energy. The proton **SEE** rate is calculated **by** integrating the product of the cross section and proton flux at each energy level [Samaras, 2014]. Equation **2.8** displays the calculation for the proton **SEE** rate with parameters of proton cross section, Σ , and the proton flux, φ , as functions of proton energy, *E*.

$$
\tau = \int_{E0}^{\infty} \varphi(E) \cdot \Sigma(E) dE \, \left[\frac{\text{Number SEES}}{\text{Unit Time}} \right] \, (\text{Equation 2.8})
$$

The OMERE **SEE** rate calculation module uses Equation **2.7** and Equation **2.8** to determine the expected ion and proton **SEE** rates, respectively, of a device for a given mission. Prior to **SEE** rate calculations, the mission radiation environment must be modeled to determine input parameters to the equations, such as LET threshold distribution and proton flux. The proton **SEE** rate calculation for the **CL20010A1** is based on sample **LEO** missions modeled in Section **3.3.**

For a device and spacecraft mission, the **SEE** rate is calculated to gain an understanding of the device sensitivity to SEEs and to determine how often the device should be power cycled or reset in order to restore nominal operation from **SEE** occurrences.

2.2.4 Summary of **Low** Earth **Orbit Radiation Effects on Spacecraft Devices**

The mission orbit determines the types of radiation effects that the spacecraft and components could encounter, while the mission duration and timeline determine the potential amount of cumulative dose or rate of single events. Each type of radiation effect is characterized to generate mission radiation requirements. Table **2.3** summarizes the types of radiation effects in a **LEO** environment based on the types of particles inducing each effect, examples of susceptible devices and technologies, and characterization of each type of radiation effect.

Table **2.3** Summary of radiation effects, types of particles inducing each effect, susceptible devices and technologies to radiation effects, and characterization of

radiation effects

2.3 OMERE Radiation Environment Analyses for Sample Low Earth Orbit Missions

Radiation analyses are conducted to evaluate the radiation environment for sample **LEO** missions. The web interface and space radiation environment simulation, OMERE version **5.0** is used to calculate the expected TID levels for the sample missions. TID radiation requirements based on the OMERE analyses are used to evaluate the radiation tolerance level of the commercial **ASIC.**

The three **LEO** space missions with different orbital parameters are modeled, **(1) ISS** Orbit (400.2 km perigee, 409.5 km apogee, 51.64 inclination), (2) **800** km altitude polar orbit, and **(3) 1000** km altitude, **0*** inclination orbit, with OMERE for mission durations of **1** year, **5** years, and **10** years with an arbitrary mission start date of January **1,** 2020. Table 2.4 summarizes the **LEO** space missions modeled.

LEO Mission	Orbit Altitude	Orbit Inclination	Orbit Type
	(ISS) 400.2 km perigee, 409.5 km apogee	51.64°	Circular
	800 km	Polar	Circular
	1000 km	∩∘	Circular

Table 2.4 **LEO** space missions modeled with OMERE

OMERE simulates the radiation environment for each mission based on models for trapped proton and electron fluxes, solar particle fluxes, and GCRs. Trapped proton fluxes are modeled using the European standard trapped proton model AP8-MIN and trapped electron model **AE8-MAX,** which were both created **by NASA.** The AP8-MIN model includes proton energy levels from **0.1** MeV to 400 MeV and represents the worst cases trapped proton fluxes, which occur during solar minimum. The **AE8-MAX** model includes electron energy levels from 0.04 MeV to **7** MeV and represents the worst case trapped electron fluxes, which occur during solar maximum. AP8-MIN is used to model all missions except for **GEO** and MEO. OMERE uses the data from both AP8-MIN and **AE8-MAX** models to calculate the mean flux for trapped particles **by** calculating the flux spectrum at each orbit point and taking the average of the flux spectrum [Varotsou, **2017].**

The fluxes and fluences from solar ions and protons are calculated with OMERE solar particle modules: *Average Statistical Models and Solar Flare Models.* Cumulated solar particle effects are calculated with *Average Statistical Models.* The mean solar proton fluxes and fluences for each mission are calculated with the **ESP** model for a **99%** confidence level. The **ESP** model is an **ECSS** 10-04 standard for solar proton fluences and covers energy levels ranging from **1** MeV to **300** MeV. The mean solar ion fluxes and fluences for each mission are calculated with the PSYCHIC solar ion model for a **99%** confidence level. Peak ion and proton fluxes during solar flares for single event rates are calculated using *Solar Flare Models.* Proton solar events for each mission are modeled using the **ONERA** model for the worst case 5-minute proton fluxes at each energy level. Heavy ion solar events for each mission are modeled using the CREME **96** model for elements He to **U.** CREME **96** is the **ECSS** 10-04 standard for heavy ion solar event fluxes [Varotsou, **2017].**

GCRs are modeled with the OMERE *Cosmic Ray module.* The GCR **ISO 15390** model is used to calculate the proton and heavy ion fluxes from cosmic rays for each mission. The GCR ISO **15390** is the standard model for cosmic rays. The 1996.4 Solar Minimum temporal configuration is used with the model, since this represents a worst case scenario for GCRs.

TID is quantified using dose model **SHIELDOSE-2,** with shielding configuration set as center of aluminum spheres and target material set as silicon. For a given orbit, **SHIELDOSE-2** determines the absorbed dose behind a range of aluminum shielding thicknesses on different detector materials from user-input electron and proton fluences [Seltzer, **1980;** Seltzer, 1994]. **SHIELDOSE-2** quantifies total ionizing dose based on calculations of electrons, bremsstrahlung photons, and protons in the radiation environment of a given orbit. On-orbit dose rates generated **by** SHIELDOSE-2 are generally consistent with dose rates less than or equal to **10** mrad/s specified in the Enhanced Low Dose Rate Sensitivity (ELDRS) radiation test procedure in **MIL-STD-883H** [Bogorad et al., 2010; Lohmeyer, **2015].**

Figure **2.7,** Figure **2.8,** and Figure **2.9** display the dose depth curves for various aluminum shielding thicknesses estimated with the SHIELDOSE-2 model for the chosen orbits and mission durations of 1-year (a), 5-years **(b),** and **10** years (c). **A** comparison of the TID curves for the **3** different mission durations is shown in Figure **2.7(d),** Figure **2.8(d),** and Figure **2.9(d).** The dose depth curves display TID contributions from trapped electrons, trapped protons, solar protons, and gamma photons, as indicated with the different colored curves defined in the legend. For a space mission, the dose depth curve is used to define the top-level dose requirement assuming a conservative shielding thickness, typically **100** mils (2.54 mm) of aluminum shielding [Poivey and Day, 2002]. The TID values at **100** mils aluminum shielding thickness are marked on the TID curves in the figures.

Figure **2.7** (a,b,c,d). Dose depth curves for **LEO** Mission at ISS Orbit for (a 1-year, (b) 5-year, and (c) 10-year (bottom left) mission durations, showing TID and contributions from trapped particles, bremsstrahlung protons, and solar protons. Estimated TID comparison for all three mission durations (d). The TID values at

100 mils aluminum shielding thickness are marked and labeled.

Figure **2.8** (a,b,c,d). Dose depth curves for **LEO** Mission at **800** km polar orbit for **1** year (a), 5-year **(b),** and 10-year (c) mission durations, showing TID and contributions from trapped particles, bremsstrahlung protons, and solar protons. Estimated TID comparison for all three mission durations **(d).** The TID values at

100 mils aluminum shielding thickness are marked and labeled.

Figure **2.9** (a,b,c,d). Dose depth curves for **LEO** Mission at **1000** km, **0*** inclination orbit for 1-year (a), 5-year **(b),** and 10-year (c) mission durations, showing TID and contributions from trapped particles, bremsstrahlung protons, and solar protons. Estimated TID comparison for all three mission durations **(d).** The TID values at **100** mils aluminum shielding thickness are marked and labeled.

As displayed in the TID curves in Figure **2.7,** Figure **2.8,** and Figure **2.9,** the magnitude of TID decreases as aluminum shielding thickness increases. For a given aluminum shielding thickness and orbit inclination, accumulated TID steeply increases as **LEO** altitude increases, and for a given aluminum shielding thickness and altitude, TID increases as **LEO** inclination increases [Lohmeyer, **2015].** For a

given aluminum shielding thickness, **LEO** altitude, and inclination, the TID increases as mission duration increases.

Table **2.5** lists the TID values at **100** mils aluminum shielding thickness for each orbit and mission duration, based on the dose depth curves.

Total Ionizing Dose for **100** mils (2.54 mm) Aluminum Shielding Thickness 1 -Year 5 -Year Orbit Mission Mission 10-Year Mission **Mission ISS 0.50** krad 2.48 krad 4.96 krad **1000** km, **0*** Inc. **1.89** krad 9.46 krad **18.9** krad 800 km, Polar | 2.94 krad | 24.2 krad | 44.3 krad

Table **2.5** Estimated total ionizing dose levels for sample **LEO** missions, calculated from OMERE radiation environment simulation.

With the selected dose requirement, a general guideline is to use a factor of two margin on the expected TID of the mission for qualifying and testing spacecraft components [Lohmeyer, **2015].** Table **2.6** defines top-level dose requirements for the sample **LEO** missions with a factor of two margin from the estimated TID in Table **2.5.** These TID requirement values will be used to assess if the **CL2001OA1** ASIC is suitable for use in **LEO** missions.

Table **2.6** Total Ionizing Dose Requirements for sample **LEO** missions based on **100** mils aluminum shielding thickness and factor of 2 margin.

Total Ionizing Dose Requirement with 2x Margin							
for 100 mils (2.54 mm) Aluminum Shielding Thickness							
Orbit	1 -Year	5-Year	10-Year				
	Mission	Mission	Mission				
ISS	1.0 krad	4.96 krad	9.92 krad				
1000 km , 0 $^{\circ}$ Inc.	3.78 krad	18.92 krad	37.8 krad				
800 km, Polar	5.88 krad	48.4 krad	88.6 krad				

Chapter 3

CL2001OA1 Optical Coherent DSP ASIC

3.1 CL2001OA1 ASIC **Description**

The Inphi **CL20010A1** is the device-under-test **(DUT)** investigated for this work. The **CL20010A1** is an optical coherent **DSP** ASIC modem. The ASIC is a monolithic, **28** nm **CMOS** modem, which supports transmission and detection of **100** Gbps and 200 Gbps information rates with polarization-multiplexed differential and nondifferential **QPSK** and **16** quadrature amplitude modulated **(QAM)** signals. The **CL20010A1** handles host and line framing/de-framing, **SD-FEC** encoding/decoding and high-speed **(-60 GSPS)** analog input and output [Inphi, 2014]. In addition, the receive channel performs high-speed **DSP** functions comprising polarization rotation, carrier phase recovery, and optical fiber dispersion compensation.

3.1.1 CL2001OA1 ASIC Functionality

The **CL2001OA1** ASIC can benefit coherent optical transceiver systems through its integrated transmitter, Tx **DAC,** framer, and digital algorithms. An integrated transmitter removes the need for a separate Tx multiplexer. **A** Tx **DAC** supports spectral shaping for tight wavelength packing. The framer enables optical lane alignment, detection, and automatic correction for phase rotation, and polarization reversal without external support. The digital algorithms programmed in the **ASIC** provide consistent performance and eliminate the need for optical dispersion compensation with losses and associated amplification [Inphi, 2014].

The **CL2001OA1** ASIC interfaces between a host system and optical hardware or components (line side). Figure **3.1** below displays the high-level data path model through the ASIC. In the Tx host-to-optics path or "egress" path, Tx data from the host is input to the ASIC for **DSP,** framing, and **FEC** processing. The Tx processed data from the ASIC is then output to optical transmitter components, such as an optical modulator, laser, etc. For the receive (Rx) optics-to-host path or "ingress" path, Rx data from optical receiver components, such as a coherent receiver, are input to the ASIC. The Rx processed data from the ASIC is then sent to the host system [Inphi, 20141.

Figure **3.1** Egress and ingress data paths for **CL2001OA1** ASIC as interface between host system and optical components [Inphi, 2014].

Figure **3.2** Data Path through **CL2001OAl** ASIC [Inphi, 2014].

The detailed data path through the ASIC is modeled in Figure **3.2.** For the egress data flow, Tx data initially enters host analog front end **(AFE)** circuitry, where clock and data are recovered for individual lanes, and signal interfaces, which align the signal to support client rate modes. The lane signals are then multiplexed and sent to the egress host framer and host **FEC** decoder. Tx data is forwarded to the egress line framer and line **FEC** encoder. Tx data then undergoes constellation mapping and Tx **DSP** with spectral shaping, pre-compensation, and skew compensation. The processed, digital Tx data is converted into four analog signal lanes (HI, **HQ,** VI, **VQ) by** DACs within **AFE** circuitry.

The ingress data flow is complementary to the egress data flow. Line side data **is** received at the line **AFE** circuitry and sampled **by** an **ADC. DSP** of the Rx signal includes bulk chromatic dispersion (BCD) equalization and timing recovery, impairment equalization, such polarization mode dispersion (PMD) and polarization dispersion loss (PDL), and carrier recovery. The 4 lanes (HI, **HQ,** VI, **VQ),** are deskewed and the constellation de-mapper is used to de-map the symbols. The ingress data is then sent to the line ingress framer and line **FEC** decoder, followed **by** the host ingress framer and **FEC** encoder. The data is de-multiplexed and is processed through the host ingress interface to support host protocols.

3.1.2 CL2001OA1 ASIC Evaluation Board System

The **CL2001OA1** was integrated on a custom-developed, coherent optical transceiver evaluation board (EVK). Due to high power consumption (~50W) of the CL20010A1, a large copper heat sink with a dedicated fan was integrated onto the EVK above the ASIC. For a space application, an alternative thermal design would need to be created to manage the **CL2001OA1** heat dissipation. This work does not focus on evaluating the entire CL2001OA1-EVK system for a space application. Rather, the purpose of this work is to assess the performance of the **CL2001OA1** based on potential radiation effects from the **LEO** environment.

An optical transceiver, the Finisar FTLC3321x3NL Compact Form Pluggable 2 Analog Coherent Optics **(CFP2-ACO)** transceiver module was also integrated onto the EVK to serve as the transmit and receive optical line side. An external power supply provides **+3.3V** to the **CFP2-ACO.**

The EVK comprises of a printed circuit board (PCB), **FPGA,** and Ethernet interface. Ethernet connection between a computer and the EVK allows for a graphical user interface (GUI) to control the **CL2001OA1** and the EVK. An external power supply provides +12.0 V to the EVK. Figure **3.3** shows the block diagram of the integrated system with the **CL2001OA1,** external heat sink, and optical transceiver on the EVK, connected to a computer/GUI and power supplies. Figure 3.4 shows the test setup of the CL2001OA1-EVK system connected to a computer/GUI and power supplies

Figure **3.3 CL2001OAl** ASIC integrated on EVK with optical transceiver. Computer and power supply connected to EVK.

Figure 3.4 Test setup of CL2001OAl-EVK system. CL2001OA1-EVK connected to computer and power supplies.

Detailed modeling of the PCB and **CL2001OAl** layers are discussed in Chapter 4 for SRIM simulations. The EVK system and **CFP2-ACO** optical transceiver have components potentially susceptible to radiation damage, such as the **FPGA,** direct current to direct current **(DC-DC)** converters, ADCs, DACs, and point of loads (POLs). Consideration was taken to decouple the effects of radiation on the EVK and CFP2- **ACO** components versus on the **CL2001OA1** from the test campaigns (Chapter **5).** The diameter and geometry of the proton beams used in the radiation test campaigns were controlled so that only the **CL2001OA1** was directly irradiated. The EVK was also designed with strategic placement of all radiation-sensitive commercial components sufficiently far away from **CL2001OA1** in case there was slight scattering of protons from the center of the beam.

3.2 Potential Radiation Damage to the CL2001OA1 ASIC in a LEO Space Environment

As a **CMOS** device with silicon active region, the **CL2001OA1** ASIC is susceptible to both ionizing and non-ionizing radiation damage. Ionizing radiation damage to **MOS** devices can occur from TID effects and SEEs, and non-ionizing radiation damage can occur from TNID or **DD** effects.

MOS devices exposed to ionizing radiation have degradation in performance parameters and changes to electrical characteristics, such as shifts in threshold voltage, decrease in gain, slowing down of integrated circuits, increased leakage currents, and ceasing to function properly [Ma and Dressendorfer, **1989]. DD** causes a reduction in minority carrier lifetime in the Si substrate of a **MOS** device. Most **MOS** devices, however, are not significantly affected **by** minority carrier lifetime [Ma and Dressendorfer, **1989].** The main concerns for the **CL2001OA1** are ionizing radiation damage from TID effects and SEEs. As mentioned in Section **2.2.3** and highlighted in Table 2.2, the types of SEEs relevant to the **CL2001OA1** ASIC include single event upsets (SEUs), single event functional interrupts (SEFIs), multiple bit upsets (MBUs), multiple cell upsets (MCUs), single event latch-ups (SELs), and single event hard errors (SEHEs) [Mutuel, **2016].**

3.2.1 Ionizing Radiation Damage to **CMOS** Devices

Ionizing radiation leads to damage because it generates mobile electrons and holes in both the $SiO₂$ insulator layer and Si substrate in MOS devices. The insulating $SiO₂$ layers of circuit structures within the **ASIC** are most susceptible to ionizing radiation damage. Radiation damage to the $SiO₂$ layer consists of the build-up of trapped charge in the oxide layer, an increase in interface traps, and an increase in bulk oxide traps. Charge and trap sites created **by** ionizing radiation originate from three different processes: electron-hole pair creation in $SiO₂$ layer, and bond breaking and atomic relaxation **of** electron-hole pairs, and electron and hole transport [Srour and McGarrity, **1988;** Ma and Dressendorfer, **1989].**

Ionizing radiation and internal photoemission from the device contacts can create electrons and holes within the $SiO₂$ layer. These electrons and holes can recombine within the $SiO₂$ layer or can be transported through the $SiO₂$ layer. Electrons are

mobile within the $SiO₂$ layer and move to the contacts. Holes have low mobility and can become trapped in the $SiO₂$ layer, generating a positive charge. If holes can move to the SiO2*/Si* interface and capture electrons, an interface trap is created. The electrons and holes transported from the $SiO₂$ layer, along with ionizing radiation, can break chemical bonds in the $SiO₂$ structure. If the broken chemical bonds do not reform from the recombination of electrons and holes, electrically active defects are formed. These defects become interface traps or trap sites for carriers. Broken bonds, specifically from hydrogen or hydroxyl groups, can release impurities which mobilize and migrate to the SiO₂/Si interface, resulting in an interface trap [Ma and Dressendorfer, **1989].**

3.2.2 Total Ionizing Dose and Single Event Upset Trends in **CMOS** Devices

There has been a recent trend toward increased TID hardness levels in **CMOS** ICs as manufacturers strategically design these technologies to be less susceptible to ionizing radiation effects. **CMOS** technologies at the **0.13** to **0.25** gm feature size have exhibited TID tolerance for up to 300 krad $(SiO₂)$ [Dodd et al., 2010]. From this, it could be hypothesized that the **CL2001OA1** should be radiation tolerant to TID levels of the **LEO** space missions modeled in Section **2.3.** Table **2.6** indicates **88.6** krad as the maximum TID requirement for all the modeled **LEO** space missions based on **100** mils aluminum shielding and factor of 2 margin. However, variation exists between manufacturers of **CMOS** technologies and even between **CMOS** devices manufactured in different lots, which could result in different radiation hardness levels [Dodd et al., 2010]. Since there have been no published studies on TID testing of the **CL2001OA1,** it is necessary to test the **CL2001OA1** ASIC for radiation damage up to TID levels from requirements of the modeled **LEO** missions.

With improvements to power efficiency, the amount of charge that represents stored information has decreased and results in increased sensitivity of **CMOS** devices to single-particle charge collection transients. Additionally, there has been increased sensitivity of **CMOS** devices to SETs as operating speeds have increased [Dodd et al.,

2010]. SEUs are the single largest contributor to device soft failure rates in **CMOS** devices. SELs are a major concern for **CMOS** devices since this type of **SEE** can be destructive. Complex devices, such as the **CL2001OA1,** are also susceptible to SEFIs [Samaras, 2014]. It is critical that the **CL2001OA1** be tested for SEEs.

Proton radiation test campaigns are completed in Chapter **5** to assess effects of TID and SEEs on the **CL20010A1.** For **SEE** testing, the CL2001OA1-EVK system is powered during proton irradiation to monitor real-time **SEE** occurrences, the time duration of an **SEE** occurrence is recorded, and SEEs effects on the system performance are observed, such as losing connection between the **CL2001OA1** and EVK and receiver loss of lock (LOL). For TID testing, the CL2001OA1-EVK system is also powered to monitor ionizing radiation effects on the performance of the **CL2001OA1** through degradation to bit error rate (BER).

Chapter 4

SRIM *Analyses* **of CL2001OA1 ASIC**

Proton beam radiation is used to test the **CL2001OA1** for ionizing radiation damage, specifically TID effects and SEEs. Prior to radiation testing, analyses are conducted to determine the threshold proton energy level required to penetrate through the active silicon region of the **CL20010A1.** Figure 4.1 highlights the relevant contributions from Chapter 4 in the high-level approach of this work toward assessing the **CL2001OA1** ASIC for application in the **LEO** radiation environment.

Figure 4.1 High-level approach used in this work to evaluate the **CL2001OA1** ASIC for **LEO** space radiation environment with Chapter 4 accomplishments highlighted in yellow.

The external heat sink above the **CL2001OA1** on the EVK precludes proton beam irradiation from the front side. The most direct path to irradiate the **CL2001OA1** is through the back side of the 24-layer printed circuit board (PCB). Detailed calculations through all PCB layers, **ASIC** ball grid array, and other packaging layers were performed to ensure the analysis of energy deposition in the active die was accurate based on the model of the CL2001OA1-EVK.

As protons travel through a material, energy is lost due to protons ionizing atoms and depositing dose along the travelled path in the material. The range that the proton penetrates through a material is dependent on the initial energy level and the target material properties, such as density, electronic characteristics, and nuclear characteristics [Buchner et al., 2002]. The maximum range that the proton penetrates through a target material is characterized **by** the Bragg peak, which is the point of maximum ionizing energy loss for the proton. Figure 4.2 displays the relationship between ionizing energy loss and the range for 64.0 MeV protons in silicon.

Figure 4.2 Energy loss for 64.0 MeV protons in silicon as a function of distance from material surface. Bragg peak indicated.

The range at which the Bragg peak occurs is calculated for radiation test campaigns when it is desired to evaluate the ionizing radiation effects for protons stopping in

the target material. However, this work focuses on evaluating the ionizing radiation effects of protons penetrating completely through the **CL2001OA1** silicon active region, which is most representative of what the ASIC will experience in a **LEO** mission for protons with energy levels high enough to penetrate through the CL2001OA1-EVK system layers. There is only a small range of energy levels near the Bragg peak energy level for which protons will stop in the active region. Otherwise, protons will either be stopped in the CL2001OA1-EVK layers above and below the active region or will completely penetrate the active region and induce ionizing radiation effects.

The CL2001OA1-EVK system modeled for the SRIM analyses in this work is not high fidelity due to the lack of proprietary information on material properties of the PCB and packaging material. With an assumed model (further described in Section 4.1), the proton energy level needed for the Bragg peak to occur in the **CL2001OA1** silicon active region cannot be accurately calculated. The **CL2001OA1** silicon active region layer is less than **1** mm thick, thus any slight deviations in the calculation could result in protons without sufficient energy to reach the target region. Overall, this work is focused on ensuring that protons have sufficient energy levels to penetrate through the **CL20010A1** silicon active region in order to produce ionizing radiation effects.

4.1 CL2001OA1-EVK System Modeling

The computer program, SRIM, specifically the Transport of Ions in Matter (TRIM) module, is used to model the **CL2001OA1** behind the EVK PCB layers and packaging material. TRIM is used to analyze the energy level needed for protons to penetrate through the PCB and packaging layers prior to reaching the **CL20010A1** silicon active region. As described in Section **3.1.2,** the **CL20010A1** was mounted between the EVK and external heat sink. Figure 4.3 shows the model of the integrated EVK system with **CL2001OA1** and displays the different layer components modeled in TRIM.

Figure 4.3 Model of **CL2001OA1** between EVK and Heat Sink

The chemical stoichiometry and material properties of the PCB and flip chip substrate are proprietary information. Because complete information would not be provided **by** Inphi, the manufacturer of the EVK, we instead approximate chemical compositions and densities of these layers for SRIM modeling.

The EVK PCB consists of 24 layers of copper with inter-layer dielectric (ILD) material between each layer. The ILD is composed of fiber-glass cloth and resin compound, Megtron **6.** Megtron **6** Prepreg cloth style **1078** was used to model the EVK PCB ILD. The Megtron **6** Prepreg **1078** compound is a polyphenylene oxide (PPO) blend resin system with 72% resin content [Panasonic, 2016]. Assuming 72% PPO (C_8H_8O) and 28% fiber glass cloth (SiO_2) , an effective density of 1.52 g/cm³ and an effective chemical stoichiometry of $C_{60}H_{60}O_{19}Si_6$ is used as a combination of PPO and fiber glass to model the ILD layers.

The flip chip substrate layer from Phoenix Precision Technology Corporation patent consists of an organic polymer resin, such as bismaleimide trizxine (BT), polyimide (PI), benzocyclobutene (BCB), liquid crystal polymeric (LCP), and polytetrafluoroethylene (PTFE) [Hsu, **2008].** PTFE is used as the material for modeling the flip chip substrate since it had the highest density (2.2 **g/cm3). A** higher density material causes higher energy loss for protons travelling through the material and this results in a more conservative TRIM calculation than a lower density material.

The external solder balls are modeled with chemical stoichiometry $\text{Sn}_{96,5}\text{Ag}_{3}\text{Cu}_{0,5}$ [Topline, **2017].** The **CL20010A1** is modeled as a silicon dye. The back end of line (BEOL) components in the **CL20010A1,** such as transistors, capacitors, etc., are not modeled since these detailed components are small features and the active silicon region is the main region of interest for radiation effects. Each layer is modeled in TRIM based on atomic composition, density, and layer thickness, as detailed in Table 4.1.

Table 4.1 CL2001OA1-EVK system layers modeled in TRIM based on atomic composition, density, and layer thickness. Parenthesis in thickness column indicate number of layers. Assumptions are indicated with (*).

4.2 TRIM Analyses of CL2001OA1-EVK System

We simulate five thousand protons $(H⁺ ions)$ with 64 MeV energy level penetrating through the modeled layers with TRIM. **A** proton energy level of 64.0 MeV is set as the baseline for the SRIM analyses since this energy level is one of the typical proton energy levels for radiation test campaigns at **UC** Davis Crocker Nuclear Laboratory **(CNL),** with maximum proton beam energy level of **67.5** MeV [Hartman, **2013].** As discussed in Chapter 2 and summarized in Table **2.1,** the **LEO** radiation environment includes trapped protons with energy levels between **10** keV to **300** MeV, solar energetic protons with energy levels up to **500** MeV, and GCR protons with energy levels up to **1** GeV [Varotsou, **2017;** Hastings and Garrett, **1996].**

TRIM simulations were approached with the intention to conduct proton test campaigns with energy levels that would span the potential **LEO** environment proton energy range. The **high** financial costs of proton test campaigns limited this work to conduct only two tests. The SRIM simulations were used to evaluate the lower energy level that could be used for a proton test campaign of the **CL20010A1.** The lower proton energy level must have sufficient energy to reach the end of the **CL2001OA1** ASIC active region after travelling through the modeled CL2001OA1-EVK system. The higher proton energy level selection was based on the upper edge of the high energy range of solar energetic protons **(500** MeV) [Varotsou, **2017].**

Within TRIM, the calculation model "Detailed Calculation with Full Damage Cascades" is used. To simulate the proton beam penetrating through the bottom of the EVK, layers are modeled in TRIM in the following order: PCB layers, external solder balls, flip chip substrate, internal **CL2001OA1** solder balls, and **CL2001OA1** silicon die.

To address concerns that **100** data points from the ionization data file would not be sufficient for accurately summarizing the ionization energy data for a complex, multi-
layer system, we use a layer-by-layer comparison approach. Two approaches are used to model the CL2001OA1-EVK system with TRIM. In the first approach, the **CL20010A1 ASIC-EVK** system is modeled with all layers of the PCB, external solder balls, flip chip substrate, internal solder balls, and **ASIC** silicon die input to TRIM as an integrated system simulation. In the second approach, the system is modeled as individual layers with a simulation for each layer. The results of these approaches are compared to ensure that a proton energy level of 64 MeV would be sufficient to reach the **CL20010A1** active area. For both approaches, the calculated ionization energy levels at the end of the **CL2001OA1** active region are analyzed.

For both approaches, the TRIM ionization data from the file IONIZ.txt is used to calculate the proton energy level through the system or individual layers. The file IONIZ.txt contains **100** data points for each simulation. With only **100** data points representing **51** modeled layers for the full integrated CL20010A1-EVK system, the ionization loss for a particular data point could encompass multiple layers. For example, some PCB ILD layers are tens of micrometers while other layers are hundreds of micrometers. We conduct a layer-by-layer approach for modeling the system to verify the results of the integrated system simulation in case there were any inaccuracies with the first modeling method.

IONIZ.txt contains an array of **3** columns: target depth in units of angstroms, the ionization energy loss from ions per target depth thickness in units of eV per angstrom, and the ionization energy loss from recoils per target depth thickness in units of eV per angstrom. Figure 4.4 below shows a section of the IONIZ.txt TRIM file for the first EVK PCB copper layer in the layer-by-layer TRIM model approach.

	SRIM-2013.00	====== H (64000) into PCB Copper Layer 1 =======================
	Ion and Recoil IONIZATION	
		See SRIM Outputs\TDATA.txt for details
		See file : SRIM Outputs\TDATA.txt for calculation data
$=$ H Ion	Energy = 64000 keV	
		============== TARGET MATERIAL ===================================
Laver 1 : Cu L1		
	Layer Width = $17780.E+01 A$;	
		Layer # 1- Density = 8.491E22 atoms/cm3 = 8.96 g/cm3
		Laver # 1- Cu = 100 Atomic Percent = 100 Mass Percent
	Total Ions calculated =005073.00	
		Ionization Energy Units are >>>> eV /(Angstrom-Ion) <<<<
TARGET	IONIZ. IONIZ.	
DEPTH	by	by
	IONS	
	177801.E-02 6035.05E-04 1284.05E-09	
	355601.E-02 6034.75E-04 0000.00E+00	
	533401.E-02 6011.48E-04 8923.87E-11	
	711201.E-02 5996.55E-04 6032.41E-11	
	889001.E-02 6010.35E-04 3499.95E-10	
	106680.E-01 6003.63E-04 5425.71E-11	
	124460.E-01 6002.04E-04 1022.91E-10	
	142240.E-01 5997.93E-04 3234.58E-11	
	160020.E-01 5996.19E-04 4382.51E-10	
	177800.E-01 6004.41E-04 4914.51E-10	
	195580.E-01 6010.76E-04 1941.87E-10	
	213360.E-01 6017.47E-04 3450.17E-10	

Figure 4.4 TRIM ionization data file "IONIZ.txt" used for calculating energy loss through EVK PCB copper layer.

A Matlab script, ioniz.m, calculates the total ionization energy loss of the protons through the modeled layer or integrated system and is included in Appendix section **A.2** Matlab Script for TRIM Simulation Data Analyses. Data from IONIZ.txt was input to the Matlab script. Equation 4.1 calculates the total ionization energy loss **by** ions *(dE)* **by** multiplying the total layer thickness *(dx)* with the average ionization energy loss per target depth thickness $\left(\frac{dE}{dx}\right)_{avg}$. The average ionization energy loss is calculated **by** taking the average of the "Ionization **by** Ions" column of data in the IONIZ.txt file. Ionization energy loss is in units of eV.

$$
dE = \left(\frac{dE}{dx}\right)_{avg} \cdot dx \left[\frac{eV}{Ang} \cdot Ang = eV\right] \text{ (Equation 4.1)}
$$

Equation 4.2 calculates the final ionization energy level of the protons (E_F) in the integrated CL2001OA1-EVK system approach, **by** subtracting the total ionization energy loss (dE) from the initial energy level of protons entering the system (E_0) or specific layer (based on which approach was used). In the layer-by-layer approach, Equation 4.3 calculates the ionization energy level of the protons at the end of a layer (E_x) by subtracting the ionization energy loss through that layer (dE_x) from the energy level of the protons penetrating the prior region $(E_{x}I)$.

 $E_F = (E_0 - dE)$ [eV] (Equation 4.2)

$$
E_x = (E_{x-1} - dE_x) [eV]
$$
 (Equation 4.3)

4.2.1 Integrated CL2001OA1-EVK System TRIM Simulation

For the integrated CL2001OA1-EVK system model approach, all layers of the EVK PCB, external solder balls, flip chip substrate, internal solder balls, and **CL2001OA1** ASIC silicon die are input to TRIM as a single model simulation. Figure 4.5 below displays a block diagram of the integrated system TRIM simulation method.

Figure 4.5 Integrated CL2001OA1-EVK System Model TRIM Simulation Method

The average ionization energy loss per target depth thickness is ~ 0.24 eV/angstrom and the total ionization energy loss through the integrated system is **-15.60** MeV. The ionization energy level of protons penetrating into the **CL200101A1** ASIC silicon region is **-50.22** MeV, and the final ionization energy level of protons at the end of the ASIC active region is \sim 48.40 MeV. Table 4.2 summarizes the results of integrated

system TRIM simulation for 64.0 MeV protons penetrating through the modeled CL200101A1-EVK system.

Table 4.2 TRIM simulation results for CL2001OA1-EVK integrated system model.

Figure 4.6 shows the TRIM simulation of five thousand protons penetrating completely through the integrated system model. The integrated system simulation method verifies that protons with 64.0 MeV energy level would be sufficient to penetrate the **CL2001OA1** active region. With a final ionization energy level of 48.40 MeV at the end of the **CL2001OA1** active region, there is more than 40 MeV of margin for 64.0 MeV protons to penetrate through the end of the active region. This amount of margin is more than sufficient to account for deviations from the actual total ionization energy loss through the system.

Figure 4.6 Integrated system TRIM simulation of five thousand 64.0 MeV protons penetrating through CL2001OAl-EVK system.

4.2.1.1 Worst-Case Integrated System TRIM Simulation

We also conduct an integrated CL2001OA1-EVK system TRIM simulation for the worst-case scenario of protons travelling through a path of copper-filled vias within the EVK PCB. The EVK PCB is modeled as an individual layer of copper. The following layers of external solder balls, flip chip substrate, internal solder balls, and **CL2001OA1** silicon die are modeled as previously described.

The "worst-case" integrated system TRIM simulation method calculates an ionization energy level of **-23.30** MeV entering the **CL2001OA1** ASIC silicon active region and an ionization energy level of **-19.87** MeV for protons penetrating through the end of the ASIC silicon active region. Table 4.3 summarizes the results of the "worst-case" integrated system TRIM simulation.

Table 4.3 TRIM simulation results for CL2001OA1-EVK integrated system model worst-case scenario.

Average Ionization Energy Loss Per	0.68 eV/angstrom	
Target Depth Thickness		
Total Ionization Loss Through		
System	$44.13\ \mathrm{MeV}$	
Ionization Energy Level Penetrating	23.30 MeV	
into CL20010A1 Active Region		
Final Ionization Energy Level at end		
of CL20010A1 Active Region	19.87 MeV	
Ionization Energy Loss in	3.43 MeV	
CL20010A1 Active Region		

Figure 4.7 shows the TRIM simulation of five thousand 64.0 MeV protons penetrating completely through the PCB layers, modeled as copper, and through the **CL2001O1A1** ASIC silicon active region (last layer to the right side of Figure 4.6). Even in this

worst-case scenario, protons with 64.0 MeV energy can still penetrate through the copper-filled vias in the PCB to reach the **CL2001OA1** active region. With a final ionization energy level of **19.87** MeV at the end of the **CL2001OA1** active region, there is more than **10** MeV margin for 64.0 MeV protons to penetrate through the end of the active region if the TRIM simulations slightly deviate from the actual total ionization energy loss through the system.

Figure 4.7 Integrated system TRIM simulation of five thousand 64.0 MeV protons penetrating through CL2001OA1-EVK system. Worst case scenario is modeled for protons through copper-filled vias in PCB.

4.2.2 Layer-By-Layer CL2001OA1-EVK System TRIM Simulations

Each layer in the CL2001OA1-EVK system was individually modeled and simulated in TRIM. The TRIM ionization data (IONIZ.txt) of each simulated layer was input to the Matlab script, ioniz.m. For each simulated layer, ioniz.m calculates the average ionization energy loss per target depth thickness, the total ionization energy loss, and the ionization energy level of protons at the end of the layer. The ionization energy level of protons at the end of each layer is used as an input for the proton energy level for the following layer simulation. Figure 4.8 below shows a block diagram of the layer-by-layer TRIM simulation method.

Figure 4.8 Layer-by-layer TRIM Simulation Method. The ionization energy level is input to a feedback loop, unlike the integrated simulation method in Figure 4.5.

For example, the first TRIM simulation for this method is a model of the copper layer on the bottom of the PCB. Protons with 64.0 MeV energy level are used for the first simulation through the first copper layer. The Matlab script calculated the ionization energy level of the protons at the end of the first copper layer as **-63.89** MeV. This is the proton energy level input for the TRIM simulation of the next layer, which was an ILD layer in the PCB. Each layer of the CL2001OA1-EVK system was modeled with the same process as described for the first copper layer until the final ionization energy level is calculated for the protons through the **CL2001OA1** ASIC active region. Table 4.4 shows the calculations for the average ionization energy loss per target depth thickness, the total ionization energy loss, and the final ionization energy level for each layer.

Table 4.4. CL2001OA1-EVK system layer-by-layer TRIM model calculations for average ionization energy loss per target depth thickness, the total ionization energy loss, and the final ionization energy level at the end of the layer.

The layer-by-layer TRIM simulation method calculates an ionization energy level of **-49.97** MeV entering the **CL2001OA1** ASIC silicon active region and an ionization energy level of \sim 48.16 MeV for protons penetrating through the end of the ASIC silicon active region. Figure 4.9 displays the TRIM simulation of five thousand protons with **-49.97** MeV energy level penetrating completely through the **CL200101A1** ASIC silicon active region.

Figure 4.9. Layer-by-layer TRIM simulation of five thousand **49.97** MeV protons penetrating through **CL2001OA1** ASIC silicon active region

The layer-by-layer method verifies that protons with 64.0 MeV energy level would be sufficient to reach the **CL2001OA1** active region. With a final ionization energy level of 48.16 MeV at the end of the **CL2001OA1** active region, there is more than 40 MeV margin for 64.0 MeV protons to penetrate through the end of the active region if the TRIM simulations slightly deviate from the actual total ionization energy loss through the system.

4.2.2.1 Worst-Case Layer-by-Layer System TRIM Simulation

A layer-by-layer TRIM simulation is also conducted for the worst-case scenario of protons travelling through a path of copper-filled vias within the EVK PCB. Table 4.5 shows the calculations for the average ionization energy loss per target depth thickness, the total ionization energy loss, and the final ionization energy level for each layer in the worst-case model.

Table 4.5 CL2001OA1-EVK system layer-by-layer TRIM model calculations for "worst-case" scenario of protons through PCB copper-filled vias. Calculations for average ionization energy loss per target depth thickness, the total ionization energy loss, and the final ionization energy level at the end of the layer.

The "worst-case" layer-by-layer TRIM simulation method results in an ionization energy level of ~23.43 MeV entering the CL20010A1 ASIC silicon active region and an ionization energy level of **-19.99** MeV for protons penetrating through the end of the ASIC silicon active region. Figure 4.10 displays the TRIM simulation of five thousand protons with -23.43 MeV energy level penetrating completely through the **CL200101A1** ASIC silicon active region. Even in this worst-case scenario, protons with 64.0 MeV energy level can still penetrate through the copper-filled vias in the PCB to reach the **CL2001OA1** active region. Based on the calculated energy level of protons penetrating through the active region, there is almost 20 MeV of margin if the TRIM simulations slightly deviate from the actual total ionization energy loss through the system.

Figure 4.10 "Worst case" layer-by-layer TRIM simulation of five thousand 23.43 MeV protons penetrating through **CL2001OA1** ASIC silicon active region.

4.3 TRIM Analyses Summary

TRIM is used to analyze if 64.0 MeV protons could penetrate through the bottom side of the EVK and reach the **CL20010A1** ASIC silicon active region. TRIM simulations of both the integrated system model and layer-by-layer model of the CL2001OA1-EVK reveals that 64.0 MeV protons would reach the ASIC active region with margin. The integrated system simulation indicated an energy level of **50.22** MeV entering the **CL2001OA1** active region and the layer-by-layer model indicated an energy level of **49.97** MeV entering the **CL20010A1** active region. The results of the two models did not significantly differ.

Simulations with both approaches are also conducted for the worst case scenario of protons travelling through a path of only copper-filled vias in the PCB. The worstcase simulation for the integrated model indicated an energy level of **23.30** MeV entering the **CL2001OA1** active region and the worst-case simulation for the layerby-layer model indicated an energy level of 23.43 MeV entering the **CL2001OA1** active region. These results also did not significantly differ.

The results of the TRIM simulations are used to help plan proton radiation testing for SEEs and TID effects at the 64.0 MeV energy level.

Table 4.6 Summary of TRIM Simulation Results for ionization energy level of protons penetrating into and completely through the **CL2001OA1** ASIC active

region.

Chapter 5

Single Event Effect and Total Ionizing Dose Assessment of the CL2001OA1 ASIC with Proton Radiation Test Campaigns

Proton test campaigns are conducted to investigate the susceptibility of the **CL20010A1** to SEEs as well as the **CL2001OA1** performance for increasing levels of TID. One proton test campaign was performed at **UC** Davis **CNL** with 64 MeV proton energy level and a second proton test campaign was conducted at Tri-University Meson Facility (TRIUMF) National Laboratory with 480 MeV proton energy level in January **2017.** The 64 MeV proton energy level was selected based on the SRIM simulations of the CL2001OA1-EVK system model described in Chapter 4. **A** higher proton energy level of 480 MeV was selected since solar energetic protons reach levels up to **500** MeV and GCR protons can reach energy levels up to **1** GeV [Varotsou, **2017;** Hastings and Garrett, **1996].** Figure **5.1** highlights the relevant contributions from Chapter **5** in the high-level approach of this work and shows the path from SRIM simulations described in Chapter 4.

for LEO space radiation environment with Chapter 5 accomplishments highlighted Figure 5.1 High-level approach used in this work to evaluate the CL20010A1 ASIC in yellow.

5.1 Experimental Approach

5.1.1 Single Event Effect and Total Ionizing Dose Testing at Crocker Nuclear Laboratory Facility

SEE and TID measurements were performed at **CNL** using the custom EVK with the **CL20010.A** ASIC and **CFP2-ACO** optical transceiver module. **All** radiation- sensitive commercial components were placed sufficiently far away from **CL2001OAl** on the EVK. The ASIC evaluation was performed in noise-loaded optical loopback (Figure **5.2).** The transmit path was noise-loaded to set the optical signal-to-noise ratio (OSNR) level near the CL20010A1 receiver FEC correction threshold. This configuration represents the most stress for the optical transceiver system, because the optical communication link is signal-starved and the receiver is operating near the **FEC** threshold. The transmit and receive wavelengths of the CFP2 module were set to **1550.92** nm and transmit power to **0** dBm. The noise loading was accomplished **by** connecting the CFP2 module transmit output to a variable optical attenuator (VOA) followed **by** an erbium doped fiber amplifier **(EDFA).** The amplifier output was filtered with a **100** GHz optical band-pass filter (OBPF) centered at **1550.92** nm. The VOA attenuation was set to produce a pre-FEC bit BER of **0.01.** The OSNR at this BER is **1.6** dB higher than that required for BER of 0.02 which is the **FEC** "breaking" threshold.

Figure **5.2 CL2001OA1 SEE** and TID experimental block diagram for **CNL** Testing

The EVK setup at **CNL** is shown in Figure **5.3,** with the bottom of the board directly exposed to the incident proton beam. **A** laser was used to align the center of the proton beam to the center of the ASIC.

Figure **5.3** EVK proton irradiation test setup at **CNL;** (left) front side of board with **CL20010A1** in top right corner and **CFP2-ACO** in top left; (right) irradiated back side of board and laser used for proton beam alignment.

At the beginning of each **SEE** test campaign, the proton beam was powered on and the start time recorded. The pre-FEC BER, the number of uncorrected **FEC** errors, and the connection to the EVK were actively monitored. Two types of SEEs were observed, SEUs and SEFIs, based on the following occurrences: EVK lost connection to **CL2001OA1** device and the receiver loss of lock (LOL). It could not be determined which type of **SEE** between SEUs and SEFIs specifically corresponded to each observed occurrence. After detection of a **SEE,** the proton beam was powered off, and the average beam flux and end time stamp were recorded. The **CL20010A1** was subsequently power cycled, restoring all pre-SEE functionality. The **SEE** test was then repeated until a total of **10 SEE** data points were collected.

After completion of **SEE** testing, performance of **CL200010 ASIC** was measured for increasing levels of TID. The **CL2001OA1 ASIC** had been exposed to a cumulative **1211.35** rad(Si) from **SEE** testing. For the first five sets of TID testing, the proton beam current was set to 5 nA for five minute intervals, yielding TID of \sim 20 krad(Si) in each interval. The last round of TID testing irradiated the **ASIC** with the same beam current, but for a 10-minute interval, providing for additional **50** krad(Si). Hence, the ASIC was irradiated to a cumulative TID of **170** krad(Si). This level of TID sufficiently encompasses the **LEO** TID requirement levels determined in Chapter 2 from modeling the sample **LEO** missions and shown in Table **2.6.** After each round of TID testing, the **CL20010A1** was power cycled and the noise-loaded optical system performance was thoroughly characterized.

5.1.2 Single Event Effect Proton Test **Campaign at TRIUMF National Laboratory**

An identical EVK with **CL2001OA1** was used for proton testing at TRIUMF National Laboratory. In contrast to the test setup at **CNL,** the EVK at TRIUMF used a Finisar ML4030 **CFP2-ACO** transceiver module in electrical loopback mode as the line-side interface, and the ASIC evaluation was not performed with noise-loading. The TRIUMF test setup diagram is shown in Figure 5.4.

Figure 5.4 **CL2001OAl SEU** experimental block diagram for TRIUMF testing

The EVK setup at TRIUMF is shown in Figure **5.5,** with the bottom side of the board directly exposed to the incident proton beam, as in the **CNL** setup. The center of the proton beam was aligned to the center of the ASIC with a laser. A 1-inch by 1-inch square aperture was used to focus the proton beam to a size closely encompassing the ASIC on the EVK.

Figure **5.5** EVK proton irradiation test setup at TRIUMF. Front side of board with **CL2001OA1** in top right corner and **CFP2-ACO** in top left. Laser used for proton beam alignment.

5.2 Experimental Results

5.2.1 Single Event Effects

The specific types of SEEs observed include SEFIs and SEUs. No SELs were observed. Table **5.1** and Table **5.2** list the durations, average flux and accumulated fluence for each **SEE** test campaign at **CNL** with 64 MeV protons and at TRIUMF with 480 MeV protons, respectively.

SEE Number	Time Duration [seconds]	Average Flux [protons/second]	Fluence [$protons/cm2$]
	183	3.23×10^6	5.91×10 ⁸
$\overline{2}$	26	3.20×10^6	8.32×107
3	13	3.20×10^6	4.16×10^{7}
	422	1.24×10^6	5.23×10^8
5	195	1.85×10^6	361×10 ⁸
6	26	1.85×10^6	3.07×107
冖	65	1.86×10 ⁶	1.21×10^{8}

Table **5.1 CL2001OA1 SEE** data from **CNL** test campaign with 64 MeV protons

Table **5.2 CL2001OA1 SEE** data from TRIUMF test campaign with 480 MeV protons.

The test data from both proton test campaigns are used to calculate the proton **SEE** cross-sections in Chapter **6.** The results are further discussed in Chapter **6** and are used to calculate the **CL2001OA1 SEE** rate for the **LEO** missions modeled in Chapter 2.

5.2.1 Total Ionizing Dose

The summary of TID runs is listed in Table **5.3.** Following each run, the **CL20010A1** had to be power cycled to restore operation. The pre-FEC BER was measured and

used to assess any potential degradation in the transmit and/or receive paths of the ASIC. In particular, any degradation in the mixed-signal portion of the device would have degraded the signal quality and therefore the BER prior to **FEC** correction. No significant measurable change in ASIC performance was observed (Figure **5.6)** up to a TID level of **170** krad(Si). The test was stopped after the sixth round of TID irradiation due to facility closure.

Table **5.3 CL2001OA1** ASIC TID data summary from **CNL** test campaign with 64

Time Duration $[\mathrm{seconds}]$	Average Beam Flux [protons/second]	Cumulative Fluence [$protons/cm2$]	Cumulative TID [krad]	Average Pre-FEC BER
348	5.93x10 ⁸	1.87×10^{11}	25.01	$9.41x10^{-3}$
300	5.56x10 ⁸	3.57×10^{11}	47.71	$9.41x10^{-3}$
300	5.69x10 ⁸	5.28x10 ¹¹	70.51	8.53×10^{-3}
300	5.75x10 ⁸	7.01x10 ¹¹	93.61	8.53×10^{-3}
300	6.42x108	8.94x10 ¹¹	119.41	9.02×10^{-3}
600	6.22×10^8	1.27×10^{12}	169.31	$9.11x10^{-3}$

MeV protons

Figure **5.6 CL2001OAl** Pre-FEC BER vs TIDs

Chapter 6

Single Event Effect Calculations for CL2001OA1 ASIC

Proton **SEE** data from the **CNL** and TRIUMF test campaigns are used to calculate the proton **SEE** cross sections for energy levels 64 MeV and 480 MeV, respectively. Figure **6.1** highlights the relevant step accomplished from Chapter **6** in the high-level approach of this work and shows the transition in the path with **SEE** data collected from the proton test campaigns. The proton **SEE** rate calculations in this chapter will be used assess mitigation strategies against radiation effects on the **CL2001OA1** (Chapter **7),** such as programming periodic resets or power cycles.

Figure **6.1** High-level approach used in this work to evaluate the **CL2001OA1** ASIC for **LEO** space radiation environment. Chapter **6** accomplishments highlighted in yellow. The transition in the path with the **SEE** data collected from the proton test

campaigns is showed. **SEE** rate calculation will be used to assess mitigation strategies against radiation effects on the **CL20010A1.**

6.1 Proton Single Event Effect Cross Section Calculations and Analyses

Equation **2.6** was used to calculate the **SEE** cross sections. The calculated **SEE** cross section values, based on proton energy level, are listed in Table **6.1** and plotted in Figure **6.2.**

Table **6.1 SEE** cross section results from proton beam testing of **CL2001OA1** ASIC

Proton Energy Level [MeV]	Single Event Effect Cross Section [cm ²]
64	2.46×10^{-9}
480	3.82×10^{-10}

Figure **6.2 CL2001OA1** ASIC proton single event effect cross-section data

The **SEE** cross-section at 480 MeV proton energy level is nearly an order of magnitude lower in comparison to that at the lower 64 MeV proton energy level. These results differ from the expectation of a higher proton **SEE** cross section for a higher proton energy level. After the Bragg peak energy level, higher proton energy levels have a

lower LET or stopping power. Figure **6.3** shows the LET curve as a function of energy level for protons penetrating through silicon material. The Bragg peak energy level for silicon is 55 keV with LET 5.382×10^{-1} MeV/(mg/cm²), and proton energy levels greater so **55** keV have decreasing LET values.

Figure **6.3** LET versus energy curve for protons through silicon target material

Previous studies on proton-induced SEEs in **CMOS** technologies, specifically SRAMs, observe results consistent this work, where lower proton energy levels induce higher **SEE** cross section values in comparison to higher proton energy levels [Heidel et al., **2008;** Cannon et al., 2010. Guillermin et al., **20161.** Both attribute the results of higher proton **SEE** cross sections for lower proton energy levels to direct ionization effects. In commercial **CMOS** circuit devices for space applications, proton-induced SEEs are dominated **by** secondary ions generated from nuclear collision events rather than **by** direct ionization [Heidel et al., **2008].** However, the studies **by** Heidel et al. **[20081,** Cannon et al. [2010], and Guillermin et al. **[2016]** find that a significant number of SEEs are produced from direct ionization from protons relative to high energy collision events. The proposal is that for a certain range of low energy protons (prior to the Bragg peak energy level), the ionization rate as the particles cross a device is

higher in comparison to the rate for higher energy protons. These low energy protons have high LETs and are losing energy rapidly [Petersen, **1996].**

The Cannon et al. [2010] study assesses proton **SEE** sensitivity of a 90-nm SRAM device. One proton test campaign in the study assesses an energy range from **0.6** MeV to 2.0 MeV at the Boeing Radiation Effects Laboratory (BREL). For proton energy levels less than **1** MeV, the cross section curve increased to a peak value at ~ **0.7** MeV then decreased for following energy levels to ~ 1 MeV. There was a slight increase in **SEE** cross section values for energy levels between 1 MeV to 2 MeV. In a second proton test campaign of the same unit, the **SEE** cross sections were determined for energy range ~20 MeV to 200 MeV at Indiana University Cyclotron Facility (IUCF). The highest proton **SEE** cross section was observed slightly above 20 MeV and the **SEE** cross section values decreased for the data collected at increasing proton energy levels. In both test campaigns, the greatest **SEE** cross section values occurred at energy levels above the silicon Bragg peak value of **55** keV.

In the Heidel et al. **[2008]** study, proton **SEE** sensitivity is assessed for a silicon on insulator **(SOI)** SRAM device. **SEE** data over a **1** to **500** MeV energy range was collected from proton test campaigns at **5** different accelerators. Similar to the Cannon et al. [2010] study, the cross section data showed significant rise in SEEs approximately below the **1** MeV energy level. Data collected from the **CNL** test campaign with the 14.6 MeV proton beam showed an increase in **SEE** cross section values between **10** MeV and **30** MeV. The cross section was also observed to slightly increase at around the **30** MeV level for proton test campaigns at TRIUMF and Northeast Proton Therapy Center **(NPTC),** and the cross section values tended to decrease as proton energy increased between **30** MeV and **100** MeV.

For a proton beam energy level of 64.0 MeV, CL2001OA1-EVK SRIM model simulations from Chapter 4 predict that the proton energy level entering the **CL2001OA1** silicon active region would be **-50** MeV for the nominal case model and **-23** MeV for the worst case model. The worst case model represents the scenario of protons travelling through a path of copper-filled vias in the PCB. The results of the **CNL** proton test campaign with 64.0 MeV protons could be considered consistent with observations of the Heidel et al. **[2008]** and Cannon et al. [2010] studies, if the worst case SRIM model simulation of the CL2001OA1-EVK occurred with **-23** MeV protons entering the silicon active region. In the Cannon et al. study [2010], the proton **SEE** cross section values decreased as proton energy levels increased between 20 MeV and 200 MeV. Although this similar trend was observed **by** Heidel et al. **[2008],** the observation was over the proton energy range between **30** MeV and **100** MeV.

The Cannon et al. study [2010] also modeled the **CMOS** SRAM **DUT** in SRIM to evaluate the proton energy levels through the active region. However, the study described limitations in the SRIM program with modeling detailed layers of the device and approximating metallization layers as a uniform layer of dielectric. The Canon et al. SRIM model did not account for different proton trajectories crossing different materials within a layer and introducing spread in the final proton energy distribution. Device metallization and passivation layers significantly broaden the proton energy spectrum at the active silicon region despite the tight energy spread of the proton beam [Cannon et al., 2010].

A similar situation to the Cannon et al. [2010] study could have occurred with this work due to limitations with the SRIM model of the CL20010A1-EVK. As described in Chapter 4, the CL2001OA1-EVK SRIM model was based on assumptions of the chemical composition of the PCB and packaging material layers as well as approximated detailed, metallization and passivation layers. In the proton test campaigns at **CNL** and TRIUMF described in Chapter **5,** broadening of the proton energy spectrum through the **51** layers of PCB and packaging material is **highly** likely to have occurred. The protons in the silicon active region of the **CL2001OA1** could have been in the energy range (near **1** MeV for the two studies mentioned) for causing direct ionization. The low energy protons between **0.5** MeV to **1.0** MeV entering the

CL20010A1 could have stopped inside the active region and the Bragg peak could have been located within the active region [Guillermin et al., **2016].**

Overall, the effects of proton direct ionization in the Heidel et al. **[2008]** and Cannon et al. [2010] appear to be consistent with the results of this work. **A** greater number of proton **SEE** data points at different energy levels will need to be collected through future test campaigns and compared to the Heidel et al. **[2008]** and Cannon et al. [2010] studies. Additional proton test campaigns are proposed and described in Section **8.2.1.**

6.2 Proton Single Event Effect Rate Calculations for the CL2001OA1 ASIC

The **SEE** proton cross section results from experimental testing of the **CL2001OA1** are used to calculate the **SEE** rate of the **CL2001OA1** for the sample **LEO** missions described in Chapter 2 and summarized in Table 2.4. Section **2.2.3.2** describes the process and details of **SEE** rate calculations. The **SEE** rate calculation module within OMERE was used for the calculation.

The **CL2001OA1 ASIC** is modeled into the component database of **SEE** rate calculation module as a component of 1 sensitive cell with a 775 μ m cell depth. The **SEE** proton cross section data from Table **6.1** was used for a Weibull fit curve. Equation **6.1** is the functional form of the Weibull function used to calculate the proton SEE cross section $(F(x))$ in units of cm² [Petersen et al., 1992; Tylka, et al., **1996].** The proton energy in units of MeV is represented **by** x, the limiting cross section is represented **by** *A,* the Weibull width parameter is represented **by W,** and the Weibull dimensionless exponent parameter is represented **by** *s.*

$$
F(x) = A\left(1 - e^{-\left[\frac{x - x_0}{W}\right]^s}\right) \, [cm^2] \, (\text{Equation 6.1})
$$

Table **6.2** displays the Weibull width parameter, W, and the Weibull dimensionless exponent parameter, s, for the **CL2001OA1** proton **SEE** cross section Weibull distribution fit. Figure 6.4 displays the Weibull curve fit from the proton **SEE** cross section data.

Table **6.2** Weibull parameters for curve fit from proton **SEE** cross section data.

0.84112	-0.28699

Figure 6.4 **CL2001OA1** proton **SEE** cross section Weibull fit curve

Due to the proton **SEE** cross section results, the **CL2001OA1** Weibull fit curve does not follow the expected Weibull distribution or behavior. The nominal Weibull fit distribution behavior is shown in Figure **6.5,** with the proton **SEE** cross section having initial logarithmic-like growth and reaching a plateau as proton energy level increases [Petersen, **1996].** The **CL2001OA1** proton **SEE** cross section Weibull fit curve has a leveled, exponentially decreasing behavior with decreasing proton **SEE** cross section for increasing proton energy levels. Additional proton test campaigns, described in section **8.2.1,** will be conducted in future work to include more proton **SEE** cross section data points at energy levels below 64 MeV and between 64 MeV and 480 MeV. Additional **SEE** cross section data points will provide greater fidelity in the Weibull fit curve and in the final **SEE** rate calculation.

Figure **6.5** Nominal proton **SEE** cross section Weibull distribution behavior

Proton mission environment data for each of the **LEO** missions is taken into account with the proton Weibull fit curve to calculate the **SEE** rate specific to each mission. The proton mission environment was defined **by** trapped proton fluxes, solar flare proton fluxes, and cosmic ray proton fluxes transported through **100** mils of aluminum shielding. The proton fluxes transported were calculated through the OMERE Transport module, which generated a transport proton flux file as an input to the environment data field of the OMERE **SEE** rate calculation module. Table **6.3** shows the proton **SEE** rate calculations for each sample **LEO** orbit.

Table **6.3** Proton **SEE** rate calculations for each modeled **LEO** orbit.

	Proton SEE Rate	Proton SEE Rate
LEO Mission Orbit	[SEEs/day]	[SEEs/year]
ISS	3.82×10^{-3}	\sim 1.38
1000 km, 0° Inclination	2.61×10^{-3}	~ 0.95

The calculated **SEE** rates for the sample **LEO** missions estimate that the **CL2001OA1** could experience under -2 proton-induced SEEs per year in an **ISS** orbit, **-1** protoninduced SEE per year in a 1000 km 0° inclination orbit, and as many as ~ 20 protoninduced SEEs per year in an **800** km polar inclination orbit. These **SEE** rates could be used to program a reset of the **CL20010A1** or power cycle the full **OTA** system, as discussed in Chapter **7.**

 \mathcal{A}

 \sim

 \mathbb{R}^2

Chapter 7

Mitigation Strategies Against Radiation Damage to the CL2001OA1 ASIC

The main mitigation strategies to protect the **CL2001OA1** against radiation damage are physical techniques of spot shielding, strategic placement in the spacecraft, and incorporation of diagnostic or protective electronic devices in the circuit board design with the CL20010A1. Software-based approaches to help resolve SEEs could be implemented, such as programming device or system resets or power cycles based on the calculated **SEE** rate could be used to resolve SEEs. Software mitigation strategies (like voting logic) cannot be applied directly to the ASIC as a pre-programmed device. Other strategies could include design and development of a radiation-tolerant ASIC, but the intent of this work is to explore the potential of using the commercial **CL20010A1** for a space application in the **LEO** radiation environment.

7.1 Spot Shielding

Spot shielding on sensitive areas of a device or system is a common mitigation strategy used **by** spacecraft designers against TID effects. Specifically, spot shielding is effective in protecting against electrons and low-energy protons. The dose depth curve from space environment modeling of a mission can be used to determine the appropriate amount of shielding needed for a component or system to meet TID requirements. Aluminum is typically the baseline shielding material used to calculate

the TID curve for a given mission and can be used for additional low mass spot shielding. Tantalum, tungsten, and lead are also typically used for spot shielding since these materials are high electron or high-Z materials with higher the density than aluminum [Maurer, **2008].** These materials could be difficult to use for masssensitive spacecraft, but the high density allows for thinner shields than aluminum. Thin shields could be specifically applied to regions with tightly packed printed circuit boards, such as an optical transceiver assembly with the **CL20010A1.**

Besides an increase in mass and volume on a spacecraft, spot shielding may have other disadvantages, such as inducing additional radiation effects. Bremsstrahlung or secondary electrons and photons can be generated from ionizing radiation penetrating through **high-Z** material spot shields [Biddle **&** Monteiro, 2012]. **A** thin inner layer of aluminum can be applied under the high-Z spot shields and at the integrated circuit die to shield against potential Bremsstrahlung radiation effects. It is important to make sure that any shielding is properly grounded. However, thick spot shields with **high-Z** materials can increase the **SEE** rate on a device from the generated Bremsstrahlung particles [Maurer, **2008].**

7.2 Strategic Placement in Spacecraft

Strategic component or system placement within the spacecraft is another common mitigation strategy. Other spacecraft components can serve as additional sources of shielding to sensitive devices or systems. Modeling can be used to determine the impact of radiation effects on a component in different locations on a spacecraft. Tools or programs, such as FASTRAD and **GEANT4,** can be used in the spacecraft design phase to ensure that sensitive components or systems are strategically placed in areas with sufficient shielding from radiation effects.

7.3 Incorporation of Protective Electronic Devices in Circuit System Design

Protective components can be incorporated to the circuit board design of an optical transceiver assembly with the **CL2001OA1** ASIC to prevent destructive damage from SEEs. **A** Latch-Up current limiter **(LCL)** is a protection device incorporated in digital circuit systems, such as ASICs and FPGAs, sensitive to SELs. SELs can have destructive effects on a device and are observed through over-current levels. LCLs, such as the **3DPM0168-2** from **3D** Plus, monitor the power supply line of the radiation sensitive device and instantaneously switches it off when it detects over-current levels in the occurrence of radiation-induced SELs **[3D** Plus, **2013].** The **3D** Plus **3DPM0168-2 LCL** is "Rad Hard **by** Design", designed with radiation mitigation techniques and utilizes space design derating rules. This **LCL** could be used in the circuit system with the **CL2001OA1** and programmed with current thresholds specific to the **CL20010A1.**

A power distribution load switch is another protective device that could be incorporated in the board layout with the **CL20010A1.** This device would also protect the **CL2001OA1** from SELs. The **AP22802 by** Diodes Incorporated is a single channel current-limited integrated high-side power switch used as a protection solution against heavy capacitive loads, short circuit, and over-current [Diodes Incorporated, **2015].**

7.4 Programming Periodic CL2001OA1 Resets or Full System Power Cycles

Resetting the device or power cycling the full system are two ways to restore loss of normal operation in the **CL2001OA1** caused **by** SEEs. Based on the **SEE** rate calculation for a given **LEO** mission described in Chapter **6** and summarized in Table **6.3,** the **CL20010A1** could be programmed with periodic resets or the full optical transceiver system could be programmed with periodic power cycles. The device resets or the full system power cycles would need to be strategically programmed to not interfere with optical transceiver operation during the mission. Power cycling the

full system with the **CL2001OA1** is an aggressive solution to restoring normal operation so it is recommended that more periodic device resets occur than **full** system power cycles.

Chapter 8

Conclusions and Future Work

8.1 Conclusions

The Inphi **CL2001OA1,** a commercial, optical coherent **DSP** ASIC, was evaluated for application in a **LEO** radiation environment. This work introduced the **LEO** radiation environment, described the main sources of radiation expected for the orbits of interest, and focused on the effects that are most relevant for the device characteristics. The radiation environment for sample **LEO** missions were modeled using OMERE and the expected TID to the **CL2001OA1** was calculated for each mission based on dose depth curves (Table **2.5). A** set of TID requirements for each **LEO** mission was created in Table **2.6.** The **CL2001OA1** was analyzed, and the potential radiation damage to this **CMOS** device was evaluated. SRIM analyses of the CL2001OA10-EVK system were conducted to evaluate the proton energy sufficient to penetrate through the EVK system and reach the silicon active region of the **CL20010A1.** Simulations showed that 64 MeV protons have sufficient energy to penetrate through the active region, and there is margin if TRIM simulations slightly deviate from the actual total ionization energy loss through the system. Two proton radiation test campaigns at 64 MeV and 480 MeV were conducted to assess the performance of the **CL2001OA1** to TID and SEEs. Proton **SEE** cross section data from the proton test campaigns was used to calculate the **SEE** rate. Potential radiation damage mitigation strategies for the **CL2001OA1** were proposed.

Based on modeling the radiation environment for the three sample **LEO** missions **(ISS** orbit, **1000** km with **0'** inclination, and **800** km in polar orbit) over 1-year, **5** year, and 10-year durations, the maximum expected TID that the **CL20010A1** ASIC would encounter for **100** mils aluminum shielding is 44.3 krad for a 10-year, **800** km polar orbit. For the TID requirements with 2x margin, the highest expected TID is **88.6** krad. The **CL2001OA1 ASIC** survived and experienced no performance degradation from TID exposure up to **170** krad(Si). Thus, the **CL2001OA1** meets the TID requirements for the sample **LEO** missions and has a TID tolerance of at least **170** krad (Si), which surpasses the requirements.

When the **CL2001OA1** is implemented into an optical transceiver assembly **(OTA)** and into a spacecraft payload, additional shielding from the **OTA** housing, payload components, and spacecraft structure will further protect the **CL2001OA1** from TID effects. Careful consideration should be taken to evaluate if other components may be susceptible to TID at lower doses. Detailed analyses with modeling tools, such as FASTRAD and **GEANT4,** can be used to evaluate the **CL2001OA1** in an **OTA** and in a spacecraft.

The measured **CL2001OA1** proton **SEE** cross section was 2.46x10-9 cm2 at the 64 MeV energy level and 3.82x10-10 cm2 at the 480 MeV energy level. The results of the **SEE** cross section data are attributed to proton direct ionization effects. Based on the cross section data, the calculated **SEE** rates for the sample **LEO** missions estimated that the CL20010A1 could experience under \sim 2 proton-induced SEEs per year in an ISS orbit, **-1** proton-induced **SEE** per year in a **1000** km **0*** inclination orbit, and as many as ~ 20 proton-induced SEEs per year in an **800** km polar inclination orbit. These **SEE** rates could be used to program a reset of the **CL2001OA1** or power cycle the full **OTA** system.

Overall the findings of this work suggest that the Inphi **CL2001OA1** ASIC could be feasibly implemented on a spacecraft application in the **LEO** radiation environment.

8.2 Future Work

Further evaluation of the performance of the **CL2001OA1** at different proton energy levels and up to higher levels of TID could be completed through additional proton radiation test campaigns. Heavy ion test campaigns of the **CL2001OA1** would also provide useful insight on the effects of heavy ions. Data from additional test campaigns could be used to better understand **CL2001OA1** radiation tolerance and effects to **CL2001OA1** performance from radiation damage. This work could be used as an initial step toward developing a coherent optical communications optical transceiver with **COTS** components, such as the **CL20010A1.**

8.2.1 Additional Proton Radiation Test Campaigns

Additional proton radiation test campaigns could be completed on the previously tested **CL2001OA1** ASICs, as described in Chapter **6,** at different proton energy levels. Proton SEE cross section data for energy levels between ~45 MeV to 64 MeV and between 64 MeV and 480 MeV could be used to evaluate **if** proton direct ionization yields effects on the **CL20010A1.** 45 MeV is preliminarily indicated as the lowest energy level for an additional test campaign, based on SRIM simulations results in Chapter 4 and summarized in Table 4.6. The SRIM simulation results showed that 64.0 MeV protons in the modeled CL2001OA1-EVK system have ionization energy levels of \sim 20 MeV and \sim 48 MeV at the end of the active region in the worst case model and baseline model, respectively. An energy level of 45 MeV protons appears sufficient for protons to still reach the end of the active region. Additional SRIM simulations with 45 MeV protons will need to be completed to confirm this hypothesis, prior to future proton test campaigns.

SEE cross section data points from additional proton test campaigns could be further compared to the two studies mentioned in Chapter **7** with relevant results due to proton direct ionization effects. The results from additional proton radiation test campaigns at different energy levels could also provide other proton **SEE** cross section data points, which could be incorporated in the Weibull-fitted **SEE** proton cross
section curve in Chapter **7.** Additional data points would provide for a more statistically significant Weibull-fitted proton cross section curve of the **CL20010A1.** Thus, a higher fidelity **CL2001OA1 SEE** rate could be estimated for a specific mission.

An additional proton radiation test campaign at the 64 MeV energy level could be conducted on the same **CL20010AI ASIC** to evaluate the ultimate TID tolerance of the device. The test campaign assessed the **CL2001OA1** up to TID level of **170** krad (Si). Further testing to higher TID levels were not completed due to time limitations at the test facility. **A** TID level of **170** krad(Si) is sufficient for a **LEO** environment, but if there is interest in testing for **GEO** environment, higher TID levels could be tested.

The 64 MeV and 480 MeV proton test campaign in Chapter **6** could be repeated for **CL20010A1** ASICs manufactured in different lot. The results of all test campaigns could be compared to provide a better understanding of differences in radiation tolerance for **CL20010A1** ASICs based on manufacturing lot and to evaluate consistency in the Inphi manufacturing process for these devices.

Table **8.1** summarizes the additional proton test campaigns that could be completed for future work. Data from these test campaigns could provide useful information to further assess proton radiation effects on the **CL20010A1.**

Table **8.1** Proposed future proton test campaigns to further assess performance of **CL2001OA1** to **LEO** radiation effects. *Indicates additional SRIM simulations needed to determine the lowest energy level to reach the end of the **CL2001OA1**

active region.

8.2.2 Heavy **Ion Radiation Test Campaigns**

 $\bar{\gamma}$

Proton beam testing was used to generate TID effects and SEEs on the **CL2001OA1** ASIC. However, it is also important to complete heavy ion radiation testing of the **CL2001OA1** to evaluate SEEs caused **by** heavy ions from GCRs and solar flares. Protons undergo nuclear interactions, which then subsequently produce SEEs through direct ionization. In contrast, most heavy ion-induced SEEs are from direct ionization [Buchner et al., 2002].

TRIM simulations and analyses, as described in Chapter 4, could be repeated with ions instead of protons prior to heavy ion testing. These simulations would be used to evaluate the required energy levels of ions sufficient to penetrate through the bottom side of the **OTA** and into the **CL2001OA1** silicon active region. The experimental methodology and setup for heavy ion testing of the CL2001OA1-EVK would be similar to the **CNL** and TRIUMF proton beam tests described in Chapter **5.** The experimental data from the heavy ion tests would be used to calculate ion **SEE** cross section data points for specific LET levels. **A** Weibull distribution curve would be fitted to the ion **SEE** cross section data from heavy ion testing, and the Weibull curve would be used to calculate the estimated **SEE** rate for a specific mission.

8.2.3 Development of a Coherent Optical Transceiver **for LEO spacecraft application** The Inphi **CL2001OA1 DSP** ASIC is an integral component needed for the development of a coherent, **DSP** optical transceiver, which can transmit and receive optical signals with data rates up to 200 Gbps. Future work will focus on designing and developing an **OTA** with the **CL2001OA1** and commercial optical components for line side.

The commercial optical components for **OTA** line side would include an optical modulator, modulator driver, and optical coherent receiver. Research will be conducted on the susceptibility of these optical components to **LEO** radiation damage. The performance of these components would be assessed for radiation damage through radiation testing. Optical components are susceptible to TID and **TNID** effects from trapped particles, solar protons, and GCR protons in the **LEO** radiation environment. Radiation testing for TID effects and displacement damage effects is also necessary for these components. After testing the commercial optical components individually, a fully integrated **OTA** with the **CL2001OA1** and line-side optical components, such as an optical modulator, modulator driver, and optical coherent receiver, could be tested for ionizing and non-ionizing radiation damage.

This work is a first step toward assessing the feasibility of a commercial-based 200 Gbps coherent optical communications transceiver for use on a spacecraft in the **LEO** radiation environment.

 \cdot

Appendix

A. 1 Single Event Effects

An **SEU** can also trigger an **SET,** which is an electrical pulse of signal or voltage generated in the device and propagated from the device to the system [Normand, 2004]. SETs occur in analog and linear devices, such as voltage references, operational amplifiers, voltage regulators, and comparators [Samaras, 2014]. Peaks in voltage amplitude and high voltage durations of a device characterize a SETAlthough a **SET** can be non-destructive to a device, the effects of a **SET** can be destructive to a system if there are connected components which take voltage input from the SET-affected device.

Single event stuck bits (SESBs) result in the change to device functionality or operation from a stuck bit in a memory device. SEBEs are a type of **SEHE** with only semi-permanent damage since annealing can recover functionality of the memory device [Samaras, 2014].

SEGRs result from an ion-induced formation of a conducting path in the gate oxide of N-channel or P-channel power MOSFETs. **A** SEGR is observed through the dielectric breakdown of a power **MOSFET** gate [Samaras, 2014]. SEBs occur in power transistors, specifically N-channel power MOSFETs. These type of SEEs result in a high current state and are caused **by** the activation of parasitic **NPN** bipolar transistor in vertical power **MOSFET** [Normand, 2004, Miroshnichenko, **2003].** SEDRs are the result of the breakdown of thin oxide layer in programmable **IC,** linear devices, such as FPGAs.

A.2 Matlab Script for TRIM Simulation Data Analyses

Filename: ioniz.m

```
1 %% TRIM Data Simulation Analyses
 2 - \text{clc};
 3- clear all;
 4- close all;
 5
 6- datafile = 'fullsys.csv'; % CSV Version of TRIM IONIZ data file
 7- EO = 64.0E6; % [MeV] Initial proton energy level into layer/system
 8
 9- ioniz_data = csvread(datafile);
10- layer = ioniz data(1:length(ionizdata),1); % [Angstrom]
11- ioniz_by_ions = ioniz data(1:length(ioniz_data),2); % [eV/Angstrom]
12- ioniz_by_recoils = ioniz_data(1:length(ioniz_data),3); % [eV/Angstrom]
13
14- dE dx = mean(ioniz byions); % [eV/Ang] Ionization energy loss per thickness
15- dx = (layer(length(layer))); % [Ang] Total layer thickness
16- dE = dE dx*dx; % [eV] Total ionization energy loss
17
18- E = (EO - dE)/(1E6); % [MeV] Ionization energy for next layer
19
20<sup>-</sup> disp(['Avg. Ionization Energy Loss per Target Depth Thickness: ' ...
21 num2str(dE dx) ' [eV/Angstrom]']);
22- disp(['Total Ionization Energy Loss: ' num2str(dE/(1E6)) ' [MeV]']);
23- disp(['Final Ionization Energy: ' num2str(E) ' [MeV]']);;
```
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