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Inverse-Designed Lithium Niobate Nanophotonics

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1 Title

2 Inverse-designed lithium niobate nanophotonics

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19 Abstract

Lithium niobate-on-insulator (LNOI) is an emerging photonic platform that exhibits favorable material 20 properties (such as low optical loss, strong nonlinearities, and stability) and enables large-scale 21 integration with stronger optical confinement, showing promise for future optical networks, quantum 22 processors, and nonlinear optical systems. However, while photonics engineering has entered the era 23 of automated "inverse design" via optimization in recent years, the design of LNOI integrated photonic 24 devices still mostly relies on intuitive models and inefficient parameter sweeps, limiting the accessible 25 parameter space, performance, and functionality. Here, we implement a 3D gradient-based inverse-26 27 design model tailored for topology optimization based on the LNOI platform, which not only could efficiently search a large parameter space but also takes into account practical fabrication constraints, 28 including minimum feature sizes and etched sidewall angles. We experimentally demonstrate a spatial-29 mode multiplexer, a waveguide crossing, and a compact waveguide bend, all with low insertion losses, 30 tiny footprints, and excellent agreement between simulation and experimental results. The devices, 31 together with the design methodology, represent a crucial step towards the variety of advanced device 32 33 functionalities needed in future LNOI photonics, and could provide compact and cost-effective solutions for future optical links, quantum technologies and nonlinear optics. 34

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36 Keywords

37 Inverse design, topology optimization, lithium niobate, integrate photonics

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39 Introduction

The lithium niobate-on-insulator (LNOI) platform has seen rapid development⁽¹⁾ in recent years and 40 shows great potential in future advanced photonics systems, owing to the excellent optical properties 41 of lithium niobate (LN), including large nonlinear susceptibilities, a wide optical transparency window, 42 and great stability, as well as strong optical confinement that allows compact and scalable integrated 43 photonic devices and circuits to be built on wafer scales. A wide range of on-chip nanophotonic devices, 44 including high-speed electro-optic modulators⁽²⁻⁴⁾, high-Q micro resonators^(5, 6), broadband frequency 45 comb generators⁽⁷⁻⁹⁾, efficient frequency convertors⁽¹⁰⁻¹³⁾, entangled photon pair generators⁽¹⁴⁾, 46 spectrometer⁽¹⁵⁾, optical isolator⁽¹⁶⁾, ultrafast all-optical switches⁽¹⁷⁾ and frequency shifters⁽¹⁸⁾ have 47 been demonstrated, making LN-based photonic integrated circuits a promising solution for future high-48 speed optical networks, quantum information processing, and nonlinear optics. However, to date the 49 design of most LNOI devices still relies heavily on human intuition augmented by simple analytical 50 models, and can only access limited parameter spaces by scanning a few hand-selected parameters 51 (such as widths, gaps and radii of curvature), either manually or through inefficient parameter sweeps 52 (whose cost scales exponentially with the number of parameters). 53

Inverse-design methods, also as known as topology optimization^(19, 20), on the other hand, have been 54 developed and attracted considerable attention in nanophotonics in the last two decades. Such methods 55 could automatically search for the optimal topological structure of a pre-specified objective in a certain 56 design region, exploring a large parameter space through meticulously developed optimization 57 algorithms. Inverse design can reveal highly non-intuitive device designs with extremely compact sizes 58 (several micrometers in diameter) and unprecedented functionalities. Many inverse-designed 59 functional devices have been demonstrated in recent years, such as photonic crystal structures⁽²¹⁻²³⁾, 60 mode multiplexers and convertors^(24, 25), wavelength multiplexers^(26, 27), meta-surfaces^(28, 29), nonlinear 61 wavelength convertors⁽²⁹⁻³¹⁾, dispersion-engineered microresonators^(32, 33), together with system 62 applications such as massively parallel optical transmitters⁽³⁴⁾, particle accelerators⁽³⁵⁾, and chip-based 63 light detection and ranging (LiDAR) systems^(36, 37). 64

- Thus far, most inverse-designed photonic devices have been demonstrated in silicon (Si) and other 65 CMOS-compatible platforms such as silicon nitride⁽²⁸⁾ because of the mature fabrication technologies 66 available for those materials. Compared with Si photonics, the challenges in the design and 67 implementation of inverse-designed LNOI devices mainly arise from practical fabrication constraints. 68 Due to the difficulty in dry etching LN and electro-optic overlap considerations, typical waveguides 69 in LNOI feature a rib structure with an unetched slab underneath, further reducing the already smaller 70 71 effective index contrast compared with that of Si photonics. Moreover, dry-etched LNOI waveguides usually exhibit a substantial sidewall angle, which needs to be taken into consideration during the 72 optimization process to achieve accurate modeling, and which also limits the minimally achievable 73 74 feature sizes. More recently, inverse-design algorithms that take into account specific geometric constraints⁽³⁸⁾ and sidewall angles⁽³⁹⁾ have led to designs and demonstrations compatible with standard 75 foundry services⁽⁴⁰⁾ and in more exotic material platforms such as diamond⁽⁴¹⁾ and silicon carbide⁽³³⁾. 76 However, the realization of inverse-designed LNOI devices is still missing and could substantially 77 benefit the development of compact LNOI photonic integrated circuits, especially when dealing with 78 complex design problems with multiple objective figures of merit. 79
- 80 Here, we overcome these fabrication and design challenges and demonstrate a series of compact high-
- 81 performance inverse-designed LNOI photonic devices based on the open-source Meep package⁽⁴²⁾.

The design algorithm takes into full consideration of the fabrication constraints in the LNOI platform, 82 including the rib structures, minimum feature sizes and etched sidewalls. We design and fabricate a 83 spatial-mode multiplexer that separates the fundamental transverse-electric (TE₀) mode and second-84 order TE (TE₁) mode, with an insertion loss ~ 1.5 dB and a crosstalk < -15.8 dB, a waveguide crossing 85 with a low loss of 0.48 dB and a crosstalk < -36 dB, and a compact waveguide bend that turns the 86 87 propagation of light by 90° with a radius of curvature of 6 µm and a loss of 0.41 dB. The devices show excellent agreement between theoretical and experimental performances, as well as broad operation 88 bandwidths from 1500 nm to 1600 nm wavelength. 89

90 **Results**

91 **Design framework.** Figure 1(a) shows an overview of our inverse-design strategy specially tailored for the LNOI platform and a representative iteration curve. The inverse design relies on a 92 hybrid time/frequency-domain topology-optimization algorithm⁽⁴³⁾ with adjoint sensitivity analysis 93 and multiple constraint functions. Starting with a homogeneous design region (typically initialized to 94 "gray"-halfway between air and LNOI) and a given objective function, e.g., the transmission 95 coefficients of a particular waveguide mode, the optimization solver efficiently calculates the objective 96 performance and the gradients in each iteration step by two FDTD (finite-difference time domain) 97 98 simulations regardless of the number of pixelated design parameters. Compared with optimization problems in SOI, the slanted sidewalls in the LNOI platform require special optimization algorithm 99 design, whereas the existence of an unetched slab substantially increases the difficulty in achieving 100 good optimization performances due to a lowered effective-index contrast. In this paper we use a 400 101 nm z-cut LN device layer with a 250 nm etch depth and a sidewall angle of 45 degree for all designs 102 (see device fabrication in supplementary information (S.I.)), similar to other devices previously 103 demonstrated in our group $^{(4, 44)}$. 104

The optimization strategy shown in Fig. 1(a) contains three main steps as outlined in Ref.⁽⁴³⁾. In the 105 first step, permittivity in the design region is allowed to vary continuously, while a convolution and 106 projection function is introduced to smoothen the design region and improve the dynamic range. The 107 projection strength β is increased gradually during the iterations to avoid a dramatic change in the 108 design parameters, such that the optimizer focuses on improving the figure of merit while gradually 109 pushing the design towards a more binarized permittivity distribution. In the second step, we 110 incorporate a geometric constraint⁽³⁸⁾ to eliminate features smaller than the indicated minimum length 111 scales, during which the design parameters are further binarized. Finally, we introduce sidewall 112 features by using linearly shifted threshold values for the projection function at different vertical 113 slices⁽³⁹⁾, so that the pattern is "eroded" as a function of the height above the slab surface (Fig. 1(b)). 114 In our simulations, we use a threshold value interval from 0.375 (bottom) to 0.625 (top), within which 115 the length scale roughly changes linearly with the threshold according to previous studies^(45, 46) and our 116 experience. In each step, the optimizer runs for several tens of iterations before convergence, the 117 precise number depending on the specific problem. Importantly, all simulations are performed in 3D, 118 taking fully into account the practical rib/slab thicknesses, crystal anisotropy, and sidewall angles, 119 which are crucial to realize experimentally achievable designs. More mathematical details regarding 120 the inverse design algorithms can be found in the device design section and Fig. S1 in S.I., as well as 121 in previous references^(38, 39, 47). Figure 1(c-e) shows schematic views of the designed structures and 122 their corresponding functions of our mode multiplexer, waveguide crossing and compact waveguide 123 bend, respectively. 124



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Figure 1. Overview of the optimization strategy and device schematic. (a) General optimization strategy and a typical iteration curve of a waveguide crossing optimization problem consisting of three main steps. In the first step, the 129 permittivity in each pixel is allowed to vary continuously; second, we add geometric constraints to binarize the structure 130 and achieve a minimum length scale; finally, slanted sidewall is introduced while the solver reoptimizes the design to 131 "compensate" for the sidewall-induced performance degradation. The cross-section views in step 2 and step 3 are drawn 132 along the diagonal of the design region. (b) Method of adding sidewall features. We introduce linearly shifted threshold 133 values η for the hyperbolic tangent projection function at different heights (left), which leads to eroded ($\eta = 0.625$), 134 normal ($\eta = 0.5$) and dilated ($\eta = 0.375$) structures at the top, center and bottom slices of the rib structure, respectively (right). We design and fabricate (c) a TE_0/TE_1 mode multiplexer, (d) a waveguide crossing, and (e) a compact waveguide 135 136 bend to verify the design strategy, all of which show good performance and excellent fabrication compatibility. 137

Spatial mode multiplexer. First, we consider an LNOI spatial mode multiplexer, an important 138 component for mode-division multiplexing (MDM) technology in future high-volume data 139 transmission systems⁽⁴⁸⁾. In our design, the fundamental TE₀ mode in a 2 μ m wide input waveguide 140 will be coupled into the upper output arm, whereas the TE_1 mode will be converted into TE_0 mode, 141 and output from the bottom arm (Fig. 2(a)). The two output waveguides are both 1 µm wide and are 142 separated by a 2 µm gap. Figure 2(a) shows the final inverse-designed pattern together with simulated 143 field evolution (E_v) for TE₀ and TE₁ input, whereas Figure 2(b) shows the scanning electron 144 microscope (SEM) image of the fabricated mode multiplexer. The footprint of the final design is $12 \times$ 145 12 μ m², orders of magnitude smaller compared with mode multiplexers based on traditional 146 asymmetrical directional couplers (ADCs)⁽⁴⁹⁾ or cascaded Mach-Zehnder interferometers (MZI)⁽⁵⁰⁾. 147 The minimum feature size is set to be $0.2 \mu m$ (at the middle slice of the rib) to balance between the 148 potential degrees of freedom in design and the fabrication constraints. Our simulation results show that, 149

for both TE₀ and TE₁ input scenarios, a single mode multiplexer features an average insertion loss less 150 than 0.9 dB and crosstalk less than -17 dB over a 100-nm wavelength window (1500 nm-1600 nm). In 151 our actual experiments, we fabricate two multiplexers back to back connected by a 50-µm-long, 2-µm-152 wide multimode waveguide, such that all input/output signals are in TE₀ mode and are easier to be 153 precisely calibrated. The simulation result (Fig. 2(c)) for such a cascaded mode multiplexer pair shows 154 155 average insertion losses at the desired outputs (S₃₁ and S₄₂) of less than 2 dB (twice the loss of a single device), and crosstalk (S₄₁ and S₃₂) of less than -16.5 dB. In the experimental test (Fig. 2(d)), we 156 measure average insertion losses of less than 3 dB and crosstalk less than -15.8 dB for the mode-157 multiplexer pair within a wavelength range between 1520 nm and 1600 nm, consistent with the 158 simulation results. The measured insertion loss values are further corroborated by comparing the losses 159 160 of two and four cascaded mode-multiplexers (see performance estimation in S.I.). From the results we estimate that a single mode multiplexer features an insertion loss ~ 1.5 dB and an upper-bound 161 162 crosstalk value of -15.8 dB within the 80-nm wavelength window (see performance estimation for a detailed analysis). The remaining differences between simulation and experimental values could result 163 from excessive scattering losses due to deviations in the actually fabricated device parameters and 164 fabrication imperfections especially on the small features (see origin of excessive insertion loss and 165 166 Fig. S2 in S.I. for more details). We also note that the peripheral structures in our design (e.g. the top and bottom bars in Fig.2(a)) do not substantially contribute to the overall field evolution (Fig. S3 in 167 S.I.), potentially allowing the achievement of similar device performances within even smaller 168 footprints by adopting customized shapes of design region. 169



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Figure 2. Inverse-designed spatial mode multiplexer. (a) The optimized design pattern and simulated field (E_y) evolution of the mode multiplexer with a $12 \times 12 \ \mu\text{m}^2$ footprint. The gray and white areas correspond to LN rib and slab regions, respectively. (b) SEM image of the fabricated mode multiplexer. (c) Simulated transmission coefficients of a back-toback mode multiplexer pair between the four input/output ports as shown in the inset. (d) Experimentally measured transmission coefficients of a fabricated mode multiplexer pair, showing broadband low-loss and low-crosstalk operation consistent with simulation results.

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Waveguide crossing. A waveguide crossing⁽⁵¹⁻⁵⁶⁾ is an essential component for signal routing in 179 large-scale and high-density photonic integrated circuits. Traditional waveguide crossing designs 180 typically rely on heuristic shaped taper or multimode interferometer (MMI) structures combined with 181 exhaustive parameter sweeps^(54, 55). Here we design a compact waveguide crossing using our inverse-182 design algorithm without the need for an initial guess. The design region (Fig. 3(a)) again has a 183 184 footprint of $12 \times 12 \ \mu m^2$, and is set to have mirror symmetry along both horizontal and vertical directions, since the x- and y-crystal orientations are isotropic in our z-cut LNOI wafer. The objective 185 function is designed to maximize the output power in fundamental TE₀ mode. Here the minimum 186 feature size is set to be 0.5 µm as this optimization problem is easier to converge. Figure 3(b) shows 187 the SEM image of the fabricated waveguide crossing. The simulated average insertion loss over a 100-188 nm wavelength band (Fig. 3(c)) is 0.22 dB, with low crosstalk of less than -40 dB. In the experiment, 189 we measure the insertion loss by cascading five and twenty crossing structures and comparing with a 190 191 single waveguide crossing (as indicated in the inset of Fig. 3(d)), showing a low fitted average insertion loss of 0.48 dB per crossing over the tested wavelength range (1500-1600 nm). The crosstalk values 192 measured from the two vertical output waveguides (port 3 and port 4, Fig. 3(d)) are -36 dB and -39 dB, 193 respectively, consistent with the simulation results. Similar to the top and bottom features in the mode 194 multiplexer, we find that the corners of the design region also do not influence much on the device 195 196 function (see Fig. S3(c-d) in S.I).



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Figure 3. Inverse-designed LNOI waveguide crossing. (a) The optimized design pattern and simulated field (E_y) evolution of the waveguide crossing with a 12 × 12 µm² footprint. The gray and white areas correspond to LN rib and slab regions, respectively. (b) SEM image of the fabricated crossing device. (c) Simulated optical transmission (S₂₁) and crosstalk (S₃₁) of the designed waveguide crossing. (d) Experimentally measured device performance, showing a low insertion loss and crosstalk from 1500 nm to 1600 nm. Inset shows cut-back loss measurement results of 5 and 20 cascaded crossing structures.

205 **Compact waveguide bend.** Finally, we demonstrate a compact bending waveguide that rotates 206 the propagation direction of the fundamental TE₀ mode by 90° within a tight bending radius of 6 μ m.

Due to the existence of unetched slab, waveguide bends in LNOI platforms usually require radii of at 207 least 30 µm to limit the radiation loss. A simple circular 90° bend with a radius of 6 µm in our current 208 platform will lead to a high loss of 3.7 dB according to our numerical simulation. Here the design 209 region (Fig. 4(a)) is $10 \times 10 \text{ }\mu\text{m}^2$ which is nearly an order of magnitude smaller than a traditional 30-210 um bend, both widths of input and output waveguides are 1 um, and the minimum length scale is 0.4 211 212 μm. Figure 4(b) shows the SEM image of the fabricated waveguide bend. Physically, this optimized bend design could be roughly interpreted as a sharp bend augmented with a Bragg-mirror-like structure 213 to suppress radiation loss, qualitatively similar to bends designed by topology optimization in other 214 material platforms⁽⁵⁷⁾, but determining the precise details requires the power of inverse design. The 215 simulated average transmission loss (Fig. 4(c), blue) in the desired band (1500 nm~1600 nm) is 0.29 216 dB, which is equivalent to a simple bend with a radius of 30 µm according to our simulation. In the 217 experimental test, we cascaded two, four, and six bends and compare the measured total insertion 218 219 losses with a reference waveguide (Fig. 4(c), red and inset). The average measured insertion loss for a single bend is 0.41 dB over the 100-nm wavelength range. The measured loss is slightly higher than 220 expected, possibly due to an under-etched rib height (~ 230 nm) resulting in more power leakage 221 through the thicker slab. Nevertheless, the measured loss is still more than 8 times lower than that of 222 a simple circular bend with the same radius. 223



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Fig. 4. Inverse-designed LNOI waveguide bend. (a) The optimized design pattern and simulated field (H_z) evolution of the bending waveguide with a 6-µm radius. The gray and white areas correspond to LN rib and slab regions, respectively. (b) SEM image of the fabricated waveguide bend. (c) Simulated and experimentally measured optical transmission of the waveguide bend, showing a low average insertion loss from 1500 nm to 1600 nm. Inset shows cut-back loss measurement results of two, four and six cascaded bends.

Conclusions and Discussions

In summary, we have successfully implemented an inverse-design algorithm which is compatible with 231 232 the practical fabrication constraints in the LNOI platform, including non-vertical sidewalls, relatively large minimum length scales and an unetched-slab structure. Based on this algorithm, we demonstrate, 233 for the first time, a series of inverse-designed LNOI devices with small footprints, low losses, low 234 crosstalk, broad bandwidths, and good consistency with theoretical prediction. Further relaxing the 235 minimum length constraints could lead to designs compatible with stepper photolithography 236 processes⁽⁵⁸⁾ with much better scalability and cost-effectiveness. The same design methodology could 237 be readily applied to achieve LNOI devices with more advanced functions, especially those make use 238 of the electro-optic and/or nonlinear properties of LN, such as fast-tunable switches⁽¹⁷⁾, dispersion-239 engineered comb generators⁽⁸⁾, and efficient nonlinear wavelength convertors⁽¹⁰⁾, and be extended to 240 other material platforms that share similar fabrication constraints, such as yttrium orthovanadate⁽⁵⁹⁾ 241 and titanium dioxide⁽⁶⁰⁾. The devices, together with the design methods, could become important 242

building blocks for future LNOI functional photonic circuits with applications in on-chip optical links, 243 quantum technologies and nonlinear optics. 244

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Supporting Information 246

247 Device fabrication & characterization; Device design; Origin of excessive insertion loss in fabricated devices; Performance estimation of a single mode multiplexer; Simulation after removing edge 248 features 249

250

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