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Time-Varying Compensation Can Yield No Improvement For ℓ_1 -Sensitivity Minimization

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Abstract

It is shown that time-varying compensation gives no improvement over time-invariant compensation for the ℓ_1 -sensitivity minimization problem.

I Introduction

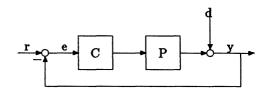


Figure 1: Basic Feedback System

Consider the feedback control system shown above in Figure 1 where P and C are causal linear operators mapping some normed linear space \mathcal{B} , with norm $\|\cdot\|_{\mathcal{B}}$, into itself; we assume that the system is well-posed. A sensitivity minimization problem for this system is a problem in which one aims to pick the compensator, C, to stabilize the closed-loop and minimize $\mu(H_{yd})$ where

 H_{yd} denotes the transfer function from d to y and μ is some given norm. The optimum " μ -sensitivity" is given by

$$\alpha \stackrel{\triangle}{=} \inf_{C \text{ stabilizes } P} \mu \left((I + PC)^{-1} \right) \tag{1}$$

A problem that falls into this framework that has received much attention ([1,2,3,4,5] et. al.) is the H_{∞} - sensitivity problem where $\mathcal{B} = \ell_2$ (with its usual norm) and μ is given by:

$$\mu(H_{yd}) \stackrel{\triangle}{=} \sup_{d \in \ell_2} \frac{\|H_{yd}(d)\|_2}{\|d\|_2}$$

Here the optimum sensitivity is the minimum energy gain from disturbance to output of the closed-loop system. It might be suspected that the achievable minimum energy gain increases when only a restricted type of stabilizing controller — e.g. finite-dimensional and time-invariant (FDLTI)— is considered. However, when P is itself FDLTI, it has been shown ([4,9]) that the optimum H_{∞} -sensitivity cannot be improved by considering time-varying or distributed controllers.

Though H_{∞} -sensitivity minimization has received much attention, it does not directly control *peak* deviation in the output, and hence the ℓ_1 -sensitivity minimization problem was formulated by Vidyasagar in [6]. For this problem, β is chosen to be ℓ_{∞} with its usual norm and μ is selected to be

$$\mu(H_{yd}) \stackrel{\triangle}{=} \sup_{d \in \ell_{\infty}} \frac{\|H_{yd}(d)\|_{\infty}}{\|d\|_{\infty}}$$

as this gives the peak gain from disturbance to output.

Once again, one might suspect that excluding time-varying or distributed controllers from consideration would worsen the achievable optimum ℓ_1 -sensitivity. Dahleh and Pearson ([7]) showed that excluding distributed compensators does not worsen the achievable optimum, and gave a procedure for computing a (rational) compensator, optimal in the class of time-invariant compensators. The question of improvement using time-varying compensation remained open. It might be pointed out that results in the H_{∞} case on time-varying compensation do not directly lead one to believe that a similiar result would be true in the ℓ_1 case. Firstly, the rational optimal solutions to the two problems have vastly different properties ([12]), and secondly the techniques used in H_{∞} proofs are not generally applicable in the ℓ_1 case.

In this paper we show that, for FDLTI plants, the optimum ℓ_1 -sensitivity achievable with time-invariant linear compensation is the same as that achievable with possibly time-varying linear compensation. Hence in fact the ℓ_1 -sensitivity minimization problem shares with the H_{∞} problem the amazing

property that over the class of time-varying distributed compensators, an FDLTI compensator is optimal.

For brevity and clarity, we shall prove the result only in the single-input, single-output (SISO) case. The generalization to the multi-input, multi-output case is straightforward.

II Main Results

Let \mathcal{L} denote the normed ring of all bounded causal linear operators, H, from ℓ_{∞} to itself that possess a pulse response h(l,k), and let \mathcal{L}_{TI} denote the subring of \mathcal{L} consisting of time-invariant operators. For $H \in \mathcal{L}$ it is easy to show that

$$\sup_{x \in \ell_{\infty}} \frac{\|H(x)\|_{\infty}}{\|x\|_{\infty}} = \sup_{k} \sum_{l=0}^{\infty} |h(l,k)| \stackrel{\triangle}{=} \|H\|_{1}$$

We can think of the operators in \mathcal{L} as infinite matrices, \mathcal{X} , that are lower triangular, have each column in ℓ_1 , and the sequence of absolute column sums in ℓ_{∞} . Of course, an operator in \mathcal{L}_{TI} is marked by the fact that its matrix is Toeplitz. We may also identify \mathcal{L}_{TI} with the space ℓ_1 . An isomorphism of \mathcal{L}_{TI} onto ℓ_1 , for instance, takes an infinite Toeplitz matrix to its first column.

The dual of \mathcal{L} , \mathcal{L}^* , is also representable as a space of infinite matrices, \mathcal{R} , upper triangular, with rows in ℓ_{∞} and the sequence of absolute row supremums in ℓ_1 .

These observations are summed up in:

Lemma 1

- 1. The subring \mathcal{L}_{TI} may be identified with ℓ_1 . The dual of ℓ_1 may be identified with ℓ_{∞} .
- 2. Any bounded linear functional, ρ , on \mathcal{L} can be identified with a doubly indexed sequence, $\rho(l,k)$ such that

$$\sum_{k=0}^{\infty} \left(\sup_{l} |\rho(l,k)| \right) = M < \infty$$

the action of ρ on H can be expressed as

$$\rho[H] = \sum_{i=0}^{\infty} \sum_{j=0}^{\infty} \rho(j,i) h(i,j)$$
 (2)

¹Trace class is sufficient.

3. Any sequence in ℓ_{∞} , $\tilde{\rho}(k)$, gives rise to a bounded linear functional, ρ on \mathcal{L} whose doubly indexed sequence is

$$ho(l,k) = \left\{ egin{array}{ll}
ho(k) & \emph{for } l=0; \ 0 & \emph{otherwise}. \end{array}
ight.$$

Moreover, ρ 's action on \mathcal{L}_{TI} agrees with $\rho(k)$'s action on ℓ_1 .

PROOF:

Easy. Left to the reader.

Note that the action (2) can be interpreted via matrices as Trace(RH).

Lemma 2 Let $\frac{N}{D}$ be a stable coprime factorization of P. Let X and Y be stable and satisfy

$$XN + YD = I$$

Then any stabilizing compensator for P in \mathcal{L} has the form

$$C = \frac{(X + DQ)}{(Y - NQ)} \qquad Q \in \mathcal{L}$$
 (3)

With C as given in (3), the closed-loop map from d to y in Figure 1 is given by

$$H_{ud} = YD - NDQ \tag{4}$$

and thus all disturbance to output maps achievable with a stabilizing time-varying compensator are given by (4) with $Q \in \mathcal{L}$; all disturbance to output maps achievable with a time-invariant compensator are given by (4) with $Q \in \mathcal{L}_{\text{TI}}$.

\underline{PROOF} :

This result can be derived from work in [1]. It appears in [10, Theorem 2.7] and [8].

Recalling (1) and using Lemma 2 the optimum ℓ_1 -sensitivity achievable with time-varying compensation is given by

$$\alpha = \inf_{Q \in \mathcal{L}} \| (YD - NDQ) \|_1 \tag{5}$$

whereas the optimum ℓ_1 -sensitivity achievable with time-invariant compensation is given by

$$\tilde{\alpha} = \inf_{Q \in \mathcal{L}_{\text{TI}}} \| (YD - NDQ) \|_1 \tag{6}$$

Our main result is the following.

Theorem 3 $\alpha = \tilde{\alpha}$.

PROOF:

For clarity in the proof, we will consider \mathcal{L}_{TI} to be a normed ring in its own right and will denote elements of $\mathcal{L}_{ extbf{T}I}$ with a tilde. The same operator considered as an element of \mathcal{L} will be denoted without the tilde. Since \mathcal{L}_{TI} is a subring of $\mathcal L$ we have a canonical imbedding (isometric isomorphism) $\phi\colon \mathcal L_{\mathtt{r} \mathtt{i}} o \mathcal L$. We will write

$$ilde{lpha} = \inf_{ ilde{Q} \in \mathcal{L}_{ extbf{TI}}} \| (ilde{H} - ilde{U} ilde{Q}) \|_1$$

and

$$\alpha = \inf_{Q \in \mathcal{L}} \| (H - UQ) \|_1$$

Where \tilde{H}, \tilde{U} are defined appropriately from (6) and H, and U are the images under $\phi(\cdot)$ of \tilde{H} , and \tilde{U} respectively.

Define the sets

$$\begin{split} \tilde{I} & \stackrel{\triangle}{=} & \left\{ \tilde{U}\tilde{Q} \mid \tilde{Q} \in \mathcal{L}_{\text{TI}} \right\} \subseteq \mathcal{L}_{\text{TI}} \\ I & \stackrel{\triangle}{=} & \left\{ UQ \mid Q \in \mathcal{L} \right\} \subseteq \mathcal{L} \\ E & \stackrel{\triangle}{=} & \left\{ \rho \in I^{\perp} \mid |\rho[G]| \le ||G||_{1} \, \forall \, G \in \mathcal{L} \right\} \subseteq \mathcal{L}^{*} \\ \tilde{E} & \stackrel{\triangle}{=} & \left\{ \tilde{\rho} \in \tilde{I}^{\perp} \mid |\tilde{\rho}[\tilde{G}]| \le ||\tilde{G}||_{1} \, \forall \, \tilde{G} \in \mathcal{L}_{\text{TI}} \right\} \subseteq \mathcal{L}^{*}_{\text{TI}} \end{split}$$

An application of Fenchel's Theorem ([11, Theorem 1, p. 119]) then gives

$$\tilde{\alpha} = \max_{\tilde{\rho} \in \tilde{E}} \tilde{\rho}[\tilde{H}] \qquad (7)$$

$$\alpha = \max_{\rho \in E} \rho[H] \qquad (8)$$

$$\alpha = \max_{\alpha \in E} \rho[H] \tag{8}$$

Now, ϕ induces a map, $\phi^*: \mathcal{L}^* \to \mathcal{L}_{\text{TI}}^*$ by

$$\phi^*(\rho)[\tilde{G}] \triangleq \rho[\phi(\tilde{G})] \tag{9}$$

Morover, by the Hahn-Banach Theorem, ϕ^* is onto $\mathcal{L}_{ exttt{ iny T}}^*$ and is norm nonincreasing.

By the results in [7] or [13], we have that \tilde{E} is spanned by m linear functionals representable as elements in ℓ_{∞} by

$$\tilde{\rho}_{\Re} = 1, \Re(z_0^{-1}), \Re(z_0^{-2}), \dots$$

$$\tilde{\rho}_{\Im} = 0, \Im(z_0^{-1}), \Im(z_0^{-2}), \dots$$

where z_0 is a non-minimum phase zero of $ilde{U}$'s z-transform. Now, for each such z_0 there are, by Lemma 1, elements ho_{\Re} and ho_{\Im} in \mathcal{L}^* that are mapped to $\tilde{
ho}_{\Re}$ and $\tilde{\rho}_{\Im}$ under ϕ^* . The action of ϕ^* given in (9) and the observation that $\phi(\tilde{I}) \subset I$, shows ρ_{\Re} and ρ_{\Im} annihilate I and hence are in E.² Thus $\phi^*(E) = \tilde{E}$. Then

$$\alpha = \max_{\rho \in E} \rho[H] = \max_{\rho \in E} \rho[\phi(\tilde{H})]$$

$$= \max_{\rho \in E} \phi^*(\rho)[\tilde{H}]$$

$$= \max_{\tilde{\rho} \in \tilde{E}} \tilde{\rho}[\tilde{H}]$$

$$= \tilde{\alpha}$$
(10)

III Conclusions

We have shown that ℓ_1 -sensitivity cannot be improved by the use of time-varying compensation. This means the ℓ_1 problem and the H_{∞} problem both share the property of having the compensator optimal over the class of all linear compensators be FDLTI when the plant is FDLTI.

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²The matrix interpretation of this is $\operatorname{Trace}(\mathcal{RUQ}) = \operatorname{Trace}((\mathcal{RU})\mathcal{Q}) = \operatorname{Trace}(0\mathcal{Q}) = 0$.

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