

# LECTURE 23

## Last time:

- Broadcast channel
- Capacity region for degraded broadcast channel
- Distributed Source Coding,
- Slepian-Wolf Theorem and Random binning

## Lecture outline

- Using random binning in broadcast channels
- Marton's region
- Fading Channels

## Review

- Degraded broadcast channel

$$R_2 \leq I(U; Y_2)$$

$$R_1 \leq I(X; Y_1|U)$$

- Distributed source coding
  - Use large enough number of bins to distinguish the typical sequences.

$$R_1 > H(U|V)$$

$$R_2 > H(V|U)$$

$$R_1 + R_2 > H(U, V)$$

## Using Random Binning in Channels

Consider a deterministic point-to-point channel  $Y = f(X)$  where  $f(\cdot)$  is deterministic and one-to-one. We try to transmit  $R$  bits per symbol.

- Assign all the possible  $\underline{Y}$  sequences into  $2^{nR}$  bins. The receiver will always decode as the bin number upon receiving  $\underline{Y}$ .
- We want to send  $M = i$ , need to control  $\underline{Y}$  to be in the  $i^{\text{th}}$  bin. As long as there is a typical  $\underline{y}$  in the  $i^{\text{th}}$  bin, we transmit the corresponding  $\underline{x}$ .
- To guarantee there is at least one typical sequence in each bin, the number of bins should be much smaller than  $|A_\epsilon(Y)|$ ,  $2^{n(H(Y)-\epsilon)}$  suffice.

What about a random channel  $P_{Y|X}$ ?

- Introduce an auxiliary random variable  $U$  as the desired output.
- Observing  $Y$ , we can determine  $2^{nI(U;Y)}$  different  $\underline{U}$  sequences.
- Notice we can talk about both the discrete and the continuous cases
- Point-to-point case:  $X = U$ , trivial

## Capacity Region for Deterministic Broadcast Channel

**Theorem** Fix an input distribution, any rate pair  $(R_1, R_2)$  that satisfies

$$\begin{aligned}R_1 &< H(Y_1) \\R_2 &< H(Y_2) \\R_1 + R_2 &< H(Y_1, Y_2)\end{aligned}$$

can be achieved.

- Divide the output space  $\mathcal{Y}_1^n \times \mathcal{Y}_2^n$  into  $2^{nR_1} \times 2^{nR_2}$  bins,
- For an input message  $M_1 \times M_2 \in \{1, \dots, 2^{nR_1}\} \times \{1, \dots, 2^{nR_2}\}$ , find a typical sequence  $\underline{y}_1 \times \underline{y}_2$  in the corresponding bin. This is the desired output
- Transmit  $\underline{x}$  to produce  $\underline{y}_1, \underline{y}_2$ .
- $R_1 < H(Y_1)$  ensures that exists typical  $\underline{y}_1$  per bin
- $R_1 + R_2 < H(Y_1, Y_2)$  ensures that exists a typical  $\underline{y}_1, \underline{y}_2$  per product bin

## General Broadcast Channel

Now if channel has randomness  $P_{Y_1, Y_2|X}$ .

- Introduce auxiliary random variables  $U, V$ .
- For fixed distribution of  $U, V$ , from the channel output, the number of distinguishable sequences  $\underline{u}$  and  $\underline{v}$  are  $2^{nI(U; Y_1)}$  and  $2^{nI(V; Y_2)}$ .

**Theorem**(Marton 75') The broadcast channel capacity region is given by

$$R_1 \leq I(U; Y_1)$$

$$R_2 \leq I(V; Y_2)$$

$$R_1 + R_2 \leq I(U; Y_1) + I(V; Y_2) - I(U; V)$$

for a fixed distribution  $P_{U, V, X}$ .

## Outline of Proof

- generate  $2^{nI(U;Y_1)}$  typical sequence  $\underline{u}$ 's and throw into  $2^{nR_1}$  bins, generate  $2^{n(I(V;Y_2))}$  typical  $\underline{v}$ 's throw into  $2^{nR_2}$  bins.
- upon receiving  $\underline{Y}_1$ ,  $\underline{u}$  can be uniquely determined. similar to  $\underline{v}$ .
- There are  $2^{n(I(U;Y_1)+I(V;Y_2))}$  possible  $(\underline{u}, \underline{v})$  pairs. Each pair being jointly typical with probability  $2^{-nI(U;V)}$ .
- If  $R_1+R_2 \leq I(U;Y_1)+I(V;Y_2)-I(U;V)$ , then there exists a jointly typical pair  $(\underline{u}, \underline{v})$  in each product bin. This is the desired received sequences.
- To transmit that bin number, simply transmit  $\underline{x}$  that is jointly typical with  $(\underline{u}, \underline{v})$ .

# Wireless Channels

**Key difference from AWGN channel** Multi-path Fading

Flat fading model

$$Y_i = H_i X_i + W_i$$

where  $W_i$  is the AWGN with variance  $\sigma_W^2$ .

- $\{H_i\}$  is a random process, for which the marginal distribution is modelled as **unit variance** Rayleigh or Ricean
- The rate at which  $\{H_i\}$  changes over time depends on
  - the speed that receiver moves,
  - the environment
  - the carrier frequency
  - the symbol period



## The wireless challenge

- How do we define a capacity for this random channel
  - the assumptions on the availability of CSI
  - the assumptions on the channel time-variation
- How well the different assumptions apply to practical channels
- How do the optimal signaling change with the channel assumptions
- How do these apply when we have a network

### Example TCP

## First simple example

- Assume  $H$  remains constant
- Assume  $H$  is perfectly known at the receiver
- Assume the transmitter does not know  $H$ , so the input distribution can not depend on  $H$ .

The capacity

$$C(H) = \max_{P_X} I(X; Y)$$

where

$$P_{Y|X} = N(HX, \sigma_W^2)$$

$$\begin{aligned} I(X; Y) &= h(Y) - h(Y|X) \\ &= h(Y) - h(W) \end{aligned}$$

- To maximize  $h(Y)$ , let  $X$  be Gaussian
- Resulting  $Y \sim N(0, |H|^2\sigma_X^2 + \sigma_W^2)$

$$C(H) = \frac{1}{2} \log \left( 1 + \frac{|H|^2\sigma_X^2}{\sigma_W^2} \right)$$

The capacity is a function of  $H$

## Outage Capacity and Ergodic Capacity

- From the transmitter point of view ( $H$  unknown), what is the maximum rate that can be transmitted with 0 probability?
- Shannon capacity for fixed fading channel is 0.
- Considering the distribution of  $H$ ,  $C(H)$  is a random variable.

**Definition**  $a\%$  outage capacity is the data rate that can be supported with  $a\%$ , i.e.,

$$P(C(H) < R) \leq a\%$$

**Definition** The ergodic capacity

$$C = E[C(H)]$$

## Ergodic Capacity

$$C = E \left[ \frac{1}{2} \log \left( 1 + |H|^2 \frac{P}{N} \right) \right]$$

- To achieve the ergodic capacity, we need to code long enough that the statistics of  $H$  start to kick in.
- Do we want channel to be time-varying or static ?
- Interleaving, delay, rate, burstiness trade-off.
- Diversity issues.
- Evaluating the capacity is hard.
- Two asymptotic results
  - as  $P/N \rightarrow 0$ , low SNR

$$C \rightarrow E[|H|^2] \frac{1}{2} \frac{P}{N}$$

the same limit as the AWGN channel

- as  $P/N \rightarrow \infty$ , high SNR

$$C \rightarrow \frac{1}{2} E[\log |H|^2] + \frac{1}{2} \log \frac{P}{N}$$

Different from the AWGN channel by a constant.